AD200 Theory of Operation

The AD200 power sensor is a sensor that detects the force that applies on the pedal of an exercise bike and calculate the power the user is using in his/her exercise. The AD200 worked with the FB613 console.

There is a reed switch in the sensor. It is use to sense the angular speed of the pedal. The sensor also has a strain gauge to detect the force the user applies onto the pedal. The strain gauge signal is amplified by a pre-amp and feeds to the ADC input of the microcontroller. The MCU will calculate the power the user is exercising using the pedal angular speed and force applied on the pedal. The microcontroller in the power sensor activates the nRF24AP2 2.4GHz transceiver IC to send encoded 2.457 GHz RF signal containing the power and angular speed information to the FB613 console and the console displays the power and angular speed reading on its LCD and may store the received power and angular speed information in its internal memory. The RF output of the nRF24AP2 2.4GHz transmitter IC is connected to the antenna via an Antenna matching network. This network is used to match the impedance of the antenna to the nRF240AP2 2.4GHz transmitter IC and suppress unwanted spurious transmission.

The power sensor needs calibration at the factory for correct operation. A static loading is put onto the power pedal. The console will transmit a calibration command to the power sensor via the 2.457MHz to activate the calibration process. When the power sensor receives the calibration command from the console it will read the force applied by the loading and store this information in its internal memory (EEPROM) to calculate the power.

The data signal is transferred through matching network (L100, L101, L102, L103, L104, L105, L106, C111, C112, C113, C114, C105 and C106) fed to the antenna.

A USB port on the PCB Board is used for factory setting only. The application software will never be sold to the end user/market and computer cannot detect the AD200 is connected. In this case this USB connector on the unit that is useless to the end user.



nRF24AP2

nRF24AP2-1CH, nRF24AP2-8CH

Single-chip ANTTM ultra-low power wireless network solution

Product Specification v1.0

Key Features

- · Second generation single chip ANT solution
- nRF24AP2-1CH supports 1 ANT (logic) channel – ideal for sensors
- nRF24AP2-8CH supports up to 8 ANT (logic) channels – ideal for hubs
- World wide 2.4 GHz ISM band operation
- Fully embedded, enhanced ANT protocol stack
- True ultra-low power operation
- Typically years of battery lifetime on a coin cell
- Built-in device search and pairing
- Built-in timing and power management
- · Built-in interference handling
- · Configurable channel period 5.2 ms 2 seconds
- Broadcast, Acknowledged and Burst communication modes
- Burst data rate up to 20 kbps
- Simple to complex network topologies:
 Peer-to-peer, star, tree and practical mesh
- Supports both public and private (managed) networks
- Support for ANT+ device profile implementations enabling multivendor interoperability
- Fully interoperable with nRF24AP1 and Dynastream ANT chipset / module based products
- Simple asynchronous/ synchronous host interface
- Single 1.9 3.6V power supply
- RoHS compliant 5x5 mm 32-pin QFN package
- Low cost external 16 MHz crystal
- Optional on-chip 32.768 kHz crystal oscillator

Applications

- Sport
- Wellness
- · Home health monitoring
- · Home/industrial automation
- Environmental sensor networks
- Active RFID
- · Logistics/goods tracking
- · Audience-response systems



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All application information is advisory and does not form part of the specification.

Limiting values

Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the specifications are not implied. Exposure to limiting values for extended periods may affect device reliability.

Life support applications

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Nordic Semiconductor ASA customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Nordic Semiconductor ASA for any damages resulting from such improper use or sale.

Datasheet status	
Objective product specification	This product specification contains target specifications for product development.
Preliminary product specification	This product specification contains preliminary data; supplementary data may be published from Nordic Semiconductor ASA later.
Product specification	This product specification contains final product specifications. Nordic Semiconductor ASA reserves the right to make changes at any time without notice in order to improve design and supply the best possible product.

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RoHS statement

nRF24AP2 where explicitly stated in this product specification meets the requirements of Directive 2002/95/EC of the European Parliament and of the Council on the Restriction of Hazardous Substances (RoHS). Complete hazardous substance reports as well as material composition reports for all active Nordic products can be found on our web site www.nordicsemi.com.

Revision History

Date Version		Description		
September 2009	1.0	Product specification		



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1 Introduction

nRF24AP2 is a member of Nordic Semiconductor's low-cost, high-performance family of 2.4 GHz ISM RF Transceivers with the ANT protocol stack embedded. nRF24AP2 offers the market's most efficient, single chip, transceiver solution for ultra low power (ULP) networks, through the integration of the extremely power efficient ANT protocol stack, the world leading Nordic Semiconductor 2.4 GHz RF technology as well as critical low-power oscillator and timing features.

This document covers the two products:

- nRF24AP2-1CH
- nRF24AP2-8CH

1.1 Prerequisites

In order to fully understand the product specification, a good knowledge of electronics and software engineering is necessary. Please also refer to the document *ANT Message Protocol and Usage* when reading this product specification. You can download the document from Nordic's web site www.nordicsemi.no or www.thisisant.com.

1.2 Writing conventions

This product specification follows a set of typographic rules to ensure that the document is consistent and easy to read. The following writing conventions are used:

- Commands, bit state conditions, and register names are written in Courier New.
- Pin names and pin signal conditions are written in Courier New bold.
- Cross references are <u>underlined and highlighted in blue</u>.



2 Product overview

ANT is a demonstrably superior Wireless Sensor Network (WSN) RF protocol for almost all practical ultralow power networking applications – from simple point-to-point links to complex networks. Embedded in nRF24AP2, it is paired up with Nordic Semiconductor's market leading 2.4 GHz radio technology. The combination gives you a high performance-, ultra-low-power network connectivity to applications, and requires minimal resources in the application's microcontroller. Less than 1 kB of code space, and an Asynchronous or Synchronous serial interface are all it takes to enable ANT connectivity in your application.

The two nRF24AP2 variants meet the specific requirements of end nodes and central nodes in a network. nRF24AP2-1CH offers one logic communication channel (ANT channel) for end nodes like sensors to connect to data collectors. nRF24AP2-8CH can manage up to eight ANT channels to collect data from multiple sensors.

<u>Figure 1. on page 7</u> shows a network in which a network node with nRF24AP2-8CH embedded, communicates with up to eight nodes with nRF24AP2-1CH devices embedded. An example might be a sports watch collecting data from several sensors (like heart rate-, speed and distance sensors). The 8-channel node can of course also set up ANT channels with other central nodes (gym equipment, for instance). These central nodes are in turn connected to more sensors. Black nodes indicate the use of nRF24AP2-8CH, white nodes indicate the use of nRF24AP2-1CH.

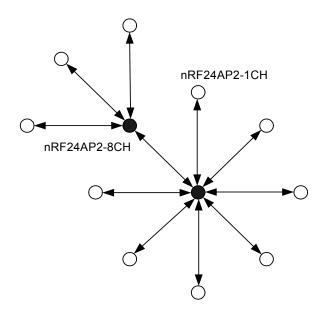


Figure 1. Simple setup with nRF24AP2

See Figure 10. on page 21 for more complex ANT-network topologies.

2.1 Features

Features of the 1-channel nRF24AP2-1CH and 8-channel nRF24AP2-8CH include:

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- Ultra low power 2.4 GHz transceiver
 - ► World wide 2.4 GHz ISM band operation
 - ▶ Based on nRF24L01+
 - GFSK modulation
 - ▶ 1 Mbps on-air data rate
 - ▶ 1 MHz frequency resolution
 - ▶ 78 RF channels
 - ▶ -85 dBm sensitivity
 - ▶ Up to 0 dBm output power
- ANT protocol stack
 - Full implementation of the physical, data link, network- and transport OSI layers
 - ▶ Packet-based communication 8 byte payload per packet
 - Optimized for ultra-low power operation
- ANT channels
 - Logic communication channel between ANT nodes
 - ▶ nRF24AP2-1CH supports 1 channel – ideal for sensors
 - ▶ nRF24AP2-8CH supports up to 8 channels –ideal for hubs
 - ▶ Built-in timing and power management
 - ▶ Built-in interference handling
 - ► Configurable channel period 5.2 ms - 2 seconds
 - ▶ Broadcast, acknowledged and burst communication modes
 - ▶ Burst data rate up to 20 kbps
- Device search and pairing
 - ▶ Wild-card searches
 - ► Proximity searches
 - ▶ Specific searches
 - Automatic link establishment if correct device is found
 - ▶ Automatic re-link attempt if link is lost
 - ▶ Configurable search timeout

- Network topologies
 - ► Point-to- point and star networks using independent ANT channels
 - Shared networks: Polled data collection (N:1) by using ANT shared channel option
 - Broadcast networks: Mass distribution of data (1:N)
- Network management / ANT+
 - Supports public and private (managed) networks
 - Support for ANT+ system implementations enabling multivendor interoperability
- ANT core stack enhancements
 - Background scanning channel
 - ▶ Continuous scanning mode
 - ► High density node support
 - Improved channel search
 - ► Channel ID management
 - Improved transmission power control on a per channel basis
 - ▶ Frequency agility
 - ▶ Proximity search
- Power Management
 - ► Fully controlled by ANT protocol stack
 - ► On-chip voltage regulator
 - ▶ Single DC supply operation
 - ▶ 1.9 to 3.6V supply range
- Ultra low power operation
 - ► Up to 50% lower average compared to nRF24AP1
 - Up to 40% lower peak current compared to nRF24AP1
 - ▶ 17 µA average current consumption at 1 Hz broadcast
 - ▶ 59 µA average current consumption at 4 Hz broadcast
- On-chip oscillators and clock inputs
 - 16 MHz crystal oscillator supporting lowcost crystals
 - ▶ 16 MHz clock input
 - Ultra low power 32.768 kHz crystal oscillator
 - ▶ 32.768 kHz clock input
- Host interface
 - Supports asynchronous and synchronous modes
 - ▶ 5-pins for asynchronous
 - ▶ 6-pins for synchronous



2.2 Block diagram

nRF24AP2 is composed of five main blocks as shown in <u>Figure 2</u>. on <u>page 9</u>. The blocks indicate the interface, power management, the ANT protocol engine, on-chip oscillators and the RF transceiver.

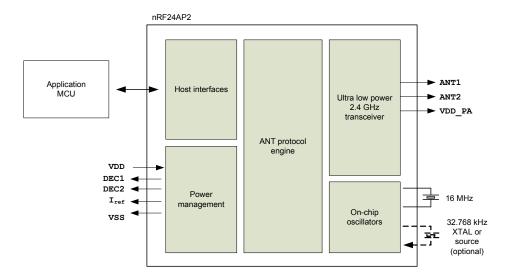


Figure 2. Block diagram of nRF24AP2 solution

To find more information about each block in the diagram, see <u>Table 1.</u>:

Name	Reference
RF Transceiver	Chapter 3 on page 13
ANT protocol engine	Chapter 4 on page 15
Host interfaces	Chapter 5 on page 24
On-chip oscillators	Chapter 6 on page 35
Power management	Chapter 8 on page 40

Table 1. Block diagram cross references



2.3 Pin Assignments

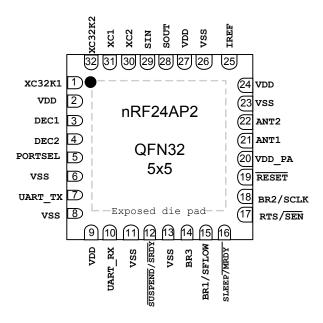


Figure 3. nRF24AP2 pin assignment (top view) for a QFN32 5x5 mm package



2.4 Pin Functions

2	XC32K1	A salas issuut				
2		Analog input				
2			optionally a synthesized or external 32.768 kHz clock			
2			can be used as described in <u>chapter 6 on page 35</u>			
	VDD	Power	Power Supply (1.9-3.6V DC)			
3	DEC1	Power	Power supply outputs for de-coupling purposes			
4		D	(100nF)			
4	DEC2	Power	Power supply outputs for de-coupling purposes (33nF)			
5	PORTSEL	Digital input	Port Select			
	FORISEL	Digital Iliput	Asynchronous serial interface: Tie to VSS			
			Synchronous serial interface: Tie to VDD			
6	VSS	Power	Ground (0V)			
7	UART TX	Output	Asynchronous mode: Transmit data signal			
	_	·	Synchronous mode: Not connected			
8	vss	Power	Ground (0V)			
9	VDD	Power	Power Supply (1.9-3.6V DC)			
10	UART_RX	Digital input	Asynchronous mode: Receive data signal			
			Synchronous mode: Tie to VDD			
11	VSS	Power	Ground (0V)			
12		Digital input	Asynchronous mode: Suspend control			
	SUSPEND/SRDY		Synchronous mode: Serial port ready			
13	vss	Power	Ground (0V)			
14	BR3		Asynchronous mode: Baud rate selection			
45	DD1 /077 077		Synchronous mode: Tie to VSS			
15	BR1/SFLOW		Asynchronous mode: Suspend Control			
			Synchronous mode: Bit or Byte flow control select (Bit:			
16			Tie to VDD, Byte: Tie to VSS) Asynchronous mode: Sleep mode enable			
10	SLEEP/MRDY	•	·			
17	SHEEF/MKD1		Synchronous mode: Message ready indication Asynchronous mode: Request to send			
1,	RTS/SEN		Synchronous mode: Serial enable signal			
18	BR2 / SCLK		Asynchronous mode: Baud rate selection			
	,	-	Synchronous mode: Clock output signal			
19			Reset, active low. Internal pull up. Leave unconnected			
	RESET	5 1	if not used.			
20	VDD_PA	Power output	Power supply output (+1.8V) for on-chip RF Power			
	_		amplifier			
21	ANT1	RF	Differential antenna connection (TX and RX)			
22	ANT2	RF	Differential antenna connection (TX and RX)			
23	vss	Power	Ground (0V)			
24	VDD	Power	Power Supply (1.9-3.6V DC)			
25	IREF	Analog output	Device reference current output. To be connected to			
26	VSS	Power	reference resistor on PCB. Ground (0V)			
26 27	VSS	Power	Power Supply (1.9-3.6V DC)			
28	SOUT		Asynchronous mode: Not connected			
20	5001	•	Synchronous mode: Not connected			
29	SIN		Asynchronous mode: Tie to VDD			
	OIN.		Synchronous mode: Data input			
30	XC2		Crystal connection for 16 MHz crystal oscillator			



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Pin	Pin name	Pin functions	Description
31	XC1	Analog Input	Crystal connection for 16 MHz crystal oscillator
32	XC32K2		Crystal connection for 32.768 kHz crystal oscillator, optionally a synthesized or external 32.768 kHz clock can be used as described in chapter 6 on page 35
Exposed die pad	VSS	Power	Connects the die pad to vss

Table 2. nRF24AP2 pin functions



3 RF Transceiver

All transceiver operations are controlled solely by the ANT protocol stack. Configuration of the ANT protocol stack occurs through a serial interface by issuing ANT commands to nRF24AP2.

3.1 Features

Features of the RF transceiver include:

- General
 - ▶ Worldwide 2.4 GHz ISM band operation
 - ▶ Common antenna interface in transmit and receive
 - ▶ GFSK modulation
 - ▶ 1 Mbps on air data rate
- Transmitter
 - ▶ Programmable output power: 0, -6, -12 or -18 dBm
- Receiver
 - ▶ Integrated channel filters
 - ▶ -85 dBm sensitivity
- · RF Synthesizer
 - ▶ Fully integrated synthesizer
 - ▶ 1 MHz frequency programming resolution
 - ▶ 78 RF channels in the 2.4 GHz ISM band
 - ▶ Accepts low cost ± 50 ppm 16 MHz crystal
 - ▶ 1 MHz non-overlapping channel spacing



3.2 Block diagram

Figure 4. on page 14 shows a block diagram of the RF transceiver in nRF24AP2.

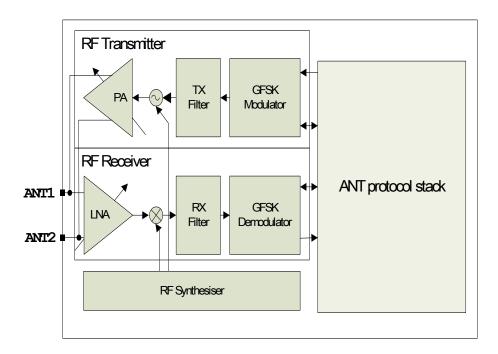


Figure 4. Internal circuitry of RF transceiver relative to ANT



4 ANT overview

The ANT protocol has been engineered for simplicity and efficiency. In operation, this results in ultra-low power consumption, maximized battery life, a minimal burden on system resources, simpler network designs and lower implementation costs.

4.1 Block diagram

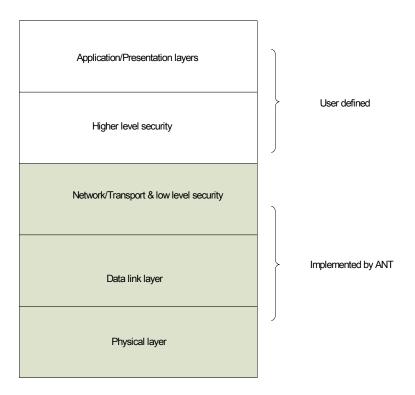


Figure 5. OSI layer model of ANT protocol stack

ANT provides carefree handling of the Physical, Data Link, Network, and Transport OSI layers. Please see <u>Figure 5</u>. on <u>page 15</u>. In addition, it incorporates key, low-level security features that form the foundation for user-defined, sophisticated, network-security implementations. ANT ensures adequate user control while considerably easing the computational burden, by providing a simple yet effective wireless networking solution.

4.2 Functional description

A brief overview of the ANT concept is presented here for convenience. A complete description of the ANT protocol is found in the ANT Message Protocol and Usage document available at www.nordicsemi.no or www.thisisant.com.



4.2.1 ANT nodes

All ANT networks are built up of nodes. See the ANT node represented in Figure 6. on page 16. A node can be anything from a simple sensor to a complex, collection unit like a watch or computer. Common to all nodes is that they contain an ANT engine (nRF24AP2) handling all connectivity to other nodes and a host processor handling the application features. nRF24AP2 interfaces to the host processor through a serial interface, and all configuration and control are done using a simple command library.

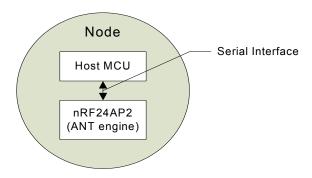


Figure 6. The ANT node

4.2.2 ANT channels

nRF24AP2 can establish one (1CH) or more (8CH) logic channels, called ANT channels, to other ANT nodes.

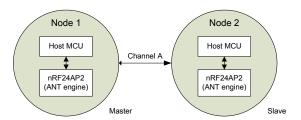


Figure 7. ANT nodes and the channel between them

The simplest ANT channel is called an independent channel and consists of two nodes, one acting as master, the other as slave for this channel. For each ANT channel opened, nRF24AP2 will set up and manage a synchronous wireless link, exchanging data packets with other ANT nodes at preset time intervals called the channel period as illustrated in Figure 8. on Page 17. The master controls the timing of a channel, that is to say, it will always initiate communication between the nodes. The slave locks on to the timing set by the master, receives the transmissions from the master and can then (if configured so) send acknowledge and/or data (if any) back to the master.



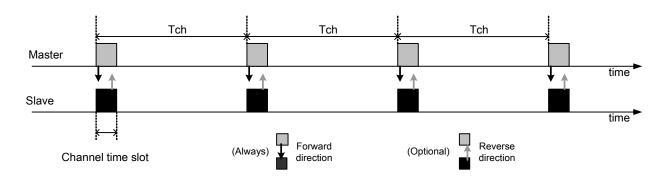


Figure 8. Channel communication showing forward and reverse directions. Not to scale.

At each time slot an ANT channel can transfer user data (8 bytes) both ways as simple broadcasts, broadcast with acknowledgement from the receiver, or transfer data as bursts (this will extend the time slot used) to accommodate transfer of larger blocks of user data. The total available payload bandwidth (20 kbps) in an ANT node is shared between active ANT channels through a Time Division Multiple Access (TDMA) scheme. If a channel timeslot comes up, but there is no new data from the master, it will still send the last packet to keep the timing of the channel and enable the slave to send data back if needed.

Each ANT channel available in the nRF24AP2 can be configured to be a simple unidirectional (broadcast) or bi-directional independent channel; or as a more complex, shared channel where a master interfaces to multiple slaves (1:N topologies). Please see the *ANT Message Protocol and Usage* document for further details on shared ANT channels.

4.2.3 ANT channel configuration

Unique to ANT is that the setup of each ANT channel is independent from all the other ANT channels in the network, including other channels in the same node. This means that one ANT node can act as master on one ANT channel while being a slave to another. Since there is no overall 'network master' present in ANT networks, ANT allows you to configure and run each ANT channel solely based on the needs of the nodes on that channel. Search- and pairing algorithms in ANT let you easily set up and shut down ANT channels in an ad-hoc fashion. This gives you ultimate flexibility in adjusting ANT channel parameters like data rate and latency versus power consumption. Moreover, you only make the network as complex as it needs to be at any given time. In order for two ANT nodes to set up an ANT channel, they must share a common channel configuration and channel ID. The necessary configuration parameters are summarized in Table 3. on page 18.



Parameter	Parameter Comment				
Channel configuration					
Channel period	Time interval between data exchanges on this				
	channel (5.2 ms - 2 s)				
RF frequencies	Which of the 78 available RF frequencies are used				
	by this channel				
Channel type	Bi-directional slave, bi-directional master, shared				
	bi-directional slave, Slave Receive only				
Network type	Decides if this ANT channel is going to be gener-				
	ally accessible (public) for all ANT nodes, having a				
	channel with the same Channel configuration and				
	ID, or if it shall limit its connectivity to devices				
	belonging to a managed or private network				
	Channel ID				
Transmission type	1 byte – Identifying characteristics of the transmis-				
	sion, can for instance contain codes on how pay-				
	load is to be interpreted				
Device type	1 byte - ID to identify the device type of the chan-				
	nel master (Ex: heartrate belt, temperature sensor				
	etc.)				
Device number	2 byte - Unique ID for this channel.				

Table 3. ANT channel ID

The channel configuration parameters are static, system parameters that must match in the master and slave, and the channel ID is included in all transmissions identifying the two nodes for each other. For indepth details on each parameter please refer to *ANT Message Protocol and Usage*, but note the following details.

Network

In addition to setting the content of the channel ID, which is the primary ID of an ANT node, ANT nodes can limit their connectivity to a selection of other ANT nodes by defining a network for each ANT channel. The limited access to certain networks is managed through unique network keys

The defined ANT networks are:

- Public networks: Open ANT networks, no limitation on connectivity. All ANT nodes sharing the same channel configuration (by design or by accident) will be able to connect. This is the default setting in nRF24AP2.
- 2. Managed networks: These are ANT networks managed by special interest groups or alliances. An example is the ANT+ alliance for sport and wellness products. To join the ANT+ alliance, please visit www.thisisant.com. By joining the ANT+ alliance and complying with the ANT+ device profiles set by the alliance, you achieve two goals:
 - ▶ Limited connectivity: Only other ANT+ compliant devices can connect on this channel.
 - ▶ Interoperability: Your node can connect to ANT+ compliant products from other vendors.
- Private networks: Your own protected networks, no other devices will be able to connect to your ANT nodes unless you share the network key. Please note that this requires purchase of a unique network key from ANT, see www.thisisant.com.

Since the network parameter can be chosen independently for each ANT channel, one ANT node (1 nRF24AP2-8CH) can have up to eight ANT channels, operating on different networks at the same time.



Note: The network parameter has no impact on the network topologies you can build. It is merely a tool to protect your ANT network and prevent accidental or deliberate access from other ANT nodes.

Channel ID, search and pairing

The primary parameters which two ANT nodes use to identify each other make up the **channel ID**. Once an ANT channel is established, the **channel ID parameters** must of course match; but they don't have to be known by both nodes (pre-configured) to be able to establish an ANT channel.

When an nRF24AP2 configured as a master (set in channel type) opens an ANT channel, it will broadcast its entire **channel ID**. Hence you must make a decision and configure all three **channel ID** parameters before opening an ANT channel as a master.

On the other hand, in a slave you can configure nRF24AP2 to search for and connect with both known and unknown masters. To connect with a known master you must configure the **Transmission type**, **Device type** and **Device number** in nRF24AP2 before opening the ANT channel.

You can also configure the nRF24AP2 to conduct wild-card searches on one or more of the three parameters in the **channel ID** to enable it to pair up with unknown masters. You can for instance set only the **Device type** of the masters you want to link up with, and set wild cards on the **Transmission type** and **Device number**. If a new master with a matching **Device type** is found, the slave device will connect and store the unknown parts of the **channel ID**. The new parts of the **channel ID** can then be stored in the host MCU to enable specific searches for this master later.

4.2.4 Proximity search

When using the basic search and pairing algorithm a slave will automatically identify and connect to the first master it finds matching the search criteria. In areas where you either have a high density of similar master nodes or high density of independent ANT networks, there is always the chance that multiple masters are found within the coverage area and you run the risk that it is not the master you want to connect to that is found first. The proximity search feature in ANT designates 'bins' of proximity from 1 (closest) to 10 furthest as shown in Figure 9. on page 19.

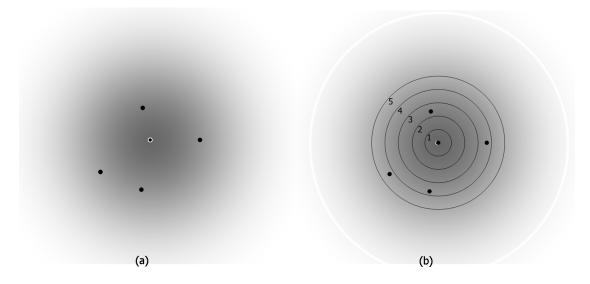


Figure 9. Standard search (a), Proximity search (b), showing bins 1-5 (of maximum 10)

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This 'binning' enables you to further control your search by for instance only accept the master that is closest (only accept masters that fall in bin 1-2), this makes it easy for a user to pair up network nodes and prevents accidental connection to nodes possibly belonging to another network close by.

4.2.5 Continuous scanning mode

Continuous scanning mode allows for fully asynchronous communication between an ANT node using continuous scanning mode, and any other ANT node using a standard master channel. This has two main advantages over only using standard ANT channels. The first is that the latency to initiate communication with the scanning node is reduced to zero and every message sent by a master channel in proximity will be received by the scanning device. Secondly, the requirement to maintain communication for the purpose of synchronization while in proximity is removed. This means that it is possible for nodes to come and go very quickly or to turn off for long periods of time in between communication events which saves power on the transmitting node.

The disadvantage of continuous scanning mode is that it consumes much more power than standard ANT channels (~17 mA) and will therefore only typically be used on devices that are plugged in and not mobile such as a PC (USB dongle). Another disadvantage is that a node in scanning mode can no longer be configured to have discoverable master channels because scanning mode disables standard ANT channel functionality. It is worth noting that two ANT nodes in scanning mode cannot communicate with one another because neither will be able to spontaneously generate communication.

Standard ANT channels are recommended over scanning channels, even in dynamic systems where devices are coming and going. This is because scanning channels are not recommended for mobile networks which is the primary use case for ANT. Scanning channels will typically be used in statically located networks where the scanning channel node is plugged in and not mobile.

4.2.6 ANT network topologies

By combining ANT channels with different features depending on local needs, you can build anything from very simple peer-to-peer links and star networks to complex networks as shown in <u>Figure 10</u>. on page 21.



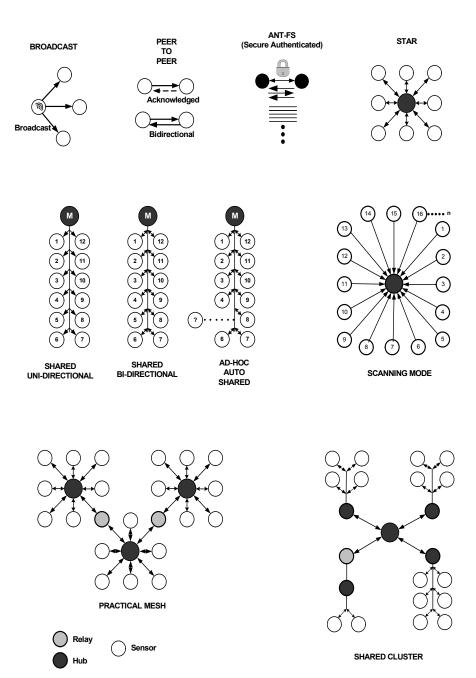


Figure 10. Network topology examples supported by ANT

4.2.7 ANT message protocol

All the configuration and control of the various ANT node and channel parameters in nRF24AP2 are handled by the host microcontroller over a simple serial interface by using the command library. See the document *ANT Message Protocol and Usage* for further details on the command library.



Class	Туре	Commands in ANT command library	Reply	From
Config.	Unassign Channel	ANT_UnassignChannel()	Yes	Host
messages	Assign Channel	ANT_AssignChannel()	Yes	Host
Channel ID		ANT_SetChannelld()	Yes	Host
	Channel Period	ANT_SetChannelPeriod()	Yes	Host
	Search Timeout	ANT_SetChannelSearchTimeout()	Yes	Host
	Channel RF Frequency	ANT_SetChannelRFFreq()	Yes	Host
	Set Network	ANT_SetNetworkKey()	Yes	Host
	Transmit Power	ANT_SetTransmitPower()	Yes	Host
	ID List Add	ANT AddChannelID() ^a	Yes	Host
	ID List Config	ANT ConfigList() ^a	Yes	Host
	Channel Transmit Power	ANT_SetLowPriorityChannelSearch Timeout()	Yes	Host
	Low Priority Search Tim- eout		Yes	Host
	Enable Ext RX Mesgs	ANT_RxExtMesgsEnable()	Yes	Host
	Enable LED	ANT EnableLED()	Yes	Host
	Crystal Enable	ANT_CrystalEnable()	Yes	Host
	Frequency Agility	ANT_ConfigFrequencyAgility()	Yes	Host
	Proximity Search	ANT_SetProximitySearch()	Yes	Host
Notifications	Startup Message	→ ResponseFunc(-, 0x6F)	-	ANT
Control	SystemReset	ANT_ResetSystem()	No	Host
Messages	Open Channel	ANT_OpenChannel()	Yes	Host
	Close Channel	ANT_CloseChannel()	Yes	Host
	Open Rx Scan Mode	ANT OpenRxScanMode() ^a	Yes	Host
	Request Message	ANT RequestMessage()	Yes	Host
	Sleep Message	ANT_SleepMessage()	No	Host
Data Messages	Broadcast Data	ANT_SendBroadcastData()	No	Host/ANT
		→ ChannelEventFunc(Chan,EV)		
	Acknowledge Data	ANT_SendAcknowledgedData()	No	Host/ANT
		→ ChannelEventFunc(Chan, EV)		
	Burst Transfer Data	ANT_SendBurstTransferPacket()	No	Host/ANT
		→ ChannelEventFunc(Chan, EV)		
Channel Event	•	→ ChannelEventFunc(Chan,	-	ANT
Messages	Event	MessageCode) or		
		→ ResponseFunc(Chan, MsgID)		
Requested	Channel Status	→ ResponseFunc(Chan, 0x52)	-	ANT
Response	Channel ID	→ ResponseFunc(Chan, 0x51)	-	ANT
Messages	ANT Version	→ ResponseFunc(Chan, 0x51)	-	ANT
	Capabilities	→ ResponseFunc(-, 0x3E)	-	ANT
Test Mode	CW Init	ANT InitCWTestMode()	Yes	Host
	CW Test	ANT SetCWTestMode()	Yes	Host
Ext Data mes-	Extended Broadcast	ANT SendExtBroadcastData() ^b	No	Host
sages	Data	→ ChannelEventFunc(Chan, EV)		
	Extended Ack. Data	ANT SendExtAcknowledgedData() ^b → ChannelEventFunc(Chan, EV)	No	Host
	Extended Burst Data	ANT SendExtBurstTransferPacket() ^b → ChannelEventFunc(Chan, EV)	No	Host

a. This is only supported by the nRF24AP2-8CH.

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b. nRF24AP2 does not send these ChannelEventFunctions() to the host. nRF24AP2 will send extended messages by appending the additional bytes to standard broadcast, acknowledged and burst data.

Table 4. ANT message summary supported by nRF24AP2



5 Host interface

The host microcontroller can configure and control all of the nRF24AP2 features through a simple serial interface. Three interface options are available, enabling both high and low end microcontrollers to be used.

5.1 Features

Serial interfaces supported by nRF24AP2:

- Asynchronous (UART)
 - ▶ Interface requires 5 pins to host microcontroller
 - ▶ Configurable baud rate from 4800 to 57600 baud
- Synchronous
 - ▶ Bit or byte flow
 - ▶ Interface requires 6 pins to host microcontroller

5.2 Asynchronous serial interface

The Host MCU and nRF24AP2 may communicate using the asynchronous mode of the serial interface. Asynchronous mode is selected by the PORTSEL input being tied low.

5.2.1 Block diagram

The asynchronous serial interface between nRF24AP2 and the Host MCU is shown in <u>Figure 11. on page 24.</u>

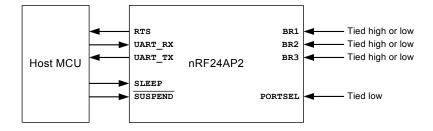


Figure 11. Asynchronous mode connections

The **UART** communication is for one start bit, one stop bit, 8 bits of data and no parity. Data is sent and received LSBit first.

5.2.2 Baud rate

The baud rate of the asynchronous communication between the Host and ANT is controlled by the speed select signals BR1, BR2 and BR3. <u>Table 5. on page 25</u> below shows the relationship between the states of the speed select signals and the corresponding baud rates.



BR3	BR2	BR1	Baud rate
0	0	0	4800
0	1	0	19200
0	0	1	38400
0	1	1	50000
1	0	0	1200
1	1	0	2400
1	0	1	9600
1	1	1	57600

Table 5. Relationship between states of speed-select signals and corresponding baud rates

Note: The baud rate may have a significant impact on system current consumption. Refer to section 8.2 on page 47 for application-specific current consumption figures.

5.2.3 Asynchronous Port Control (RTS)

When nRF24AP2 is configured in asynchronous mode, a full duplex asynchronous serial port is provided with flow control for data transmission from the Host to ANT. The flow control is performed by the RTS signal, which conforms to standard hardware flow control CMOS signal levels. The signal may therefore be attached to a PC serial port (with use of an RS-232 level shifter), or to any other RS-232 device. The RTS signal is de-asserted for approximately 50 µs after each correctly formatted message has been received. This RTS signal duration is independent of the baud rate. Incorrect messages or partial messages are not acknowledged.

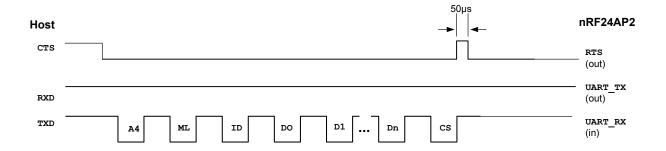


Figure 12. RTS signal following a serial Host -> nRF24AP2 transfer

When nRF24AP2 raises the RTS signal high, the Host MCU may not send any more data until the RTS signal is lowered again. There is no flow control for data being transmitted from nRF24AP2 to the Host controller, and therefore the Host controller must be able to receive data at any time. RTS is toggled following a reset.

The RTS signal is raised by nRF24AP2 after the last byte of a message has been received, and nRF24AP2 will therefore lose any bytes that were sent, or in the process of being sent, before the RTS signal is acted upon by the Host MCU, and the transmission is halted. To avoid this problem, either the messages need to be spaced apart by the Host MCU or 0-pad bytes need to be added to the end of each message being transmitted to handle whatever byte pipeline is in place. For example, when considering computer communication, two 0-bytes must be appended to every message, since computers interpret CTS at the driverrather than the hardware level. nRF24AP2 will discard 0-pad bytes received. This issue usually occurs only when using burst transfers from the Host to nRF24AP2 and high data rates are expected.



5.2.4 Sleep enable (SLEEP)

The **SLEEP** input signal allows nRF24AP2 to sleep when the serial port is not required. The signal is essential for conserving power when using the asynchrnous serial interface. This control mechanism is illustrated below in <u>Figure 13</u>. on page 26.

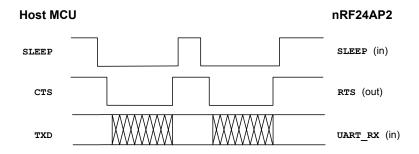


Figure 13. nRF24AP2 sleep control

If the **SLEEP** signal is not used, then it must be tied low. In this configuration, the nRF24AP2 will never sleep and will always be ready to receive data. The **SUSPEND** functionality cannot be used if the **SLEEP** signal is not used.

The **SLEEP** and **RTS** signals only affect the data being transferred from the Host MCU to nRF24AP2. nRF24AP2 will send data to the Host, when available, regardless of the state of these two signals.

5.2.5 Suspend mode control (SUSPEND)

When using the asynchronous serial interface, you also have a SUSPEND signal available. The assertion of the SUSPEND signal will cause nRF24AP2 to terminate all RF and serial port activity and power down. This will happen immediately, regardless of the state of the nRF24AP2 system. This signal provides support for use in USB applications, where USB devices are required to quickly enter a low-power state through hardware control.

Entering and exiting from the suspend mode require the use of the SLEEP signal, in addition to the SUSPEND signal. The assertion of SUSPEND is only recognized if SLEEP is also asserted at the time. Deassertion of the SLEEP signal is the only method for exiting from suspend mode, as shown in Figure 12. on



page 25. Following exit, all previous transactions and configurations will be lost – nRF24AP2 will be in its power-up state.

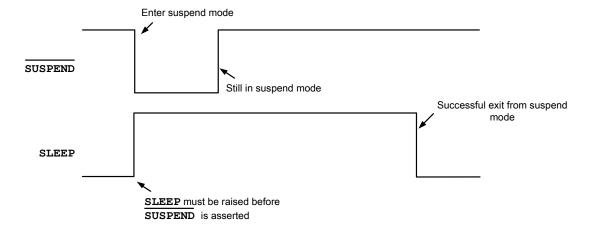


Figure 14. SUSPEND signal use

5.3 Synchronous serial interface

This section explains in detail the synchronous serial interface between nRF24AP2 and a Host MCU. This mode is selected by connecting the **PORTSEL** input high.

When operating in synchronous mode careful attention to reset behavior is required to prevent inadvertent deadlock conditions between nRF24AP2 and the Host MCU.

In synchronous mode, nRF24AP2 uses a half-duplex synchronous master serial interface with message flow control. The Host must be configured as a synchronous slave. The interface is meant to accommodate either a hardware synchronous slave port or a simple I/O control on the Host MCU. The Host MCU retains full control of the message flow and can halt incoming messages as required.

5.3.1 Block diagram

The synchronous serial interface between nRF24AP2 and the Host MCU is shown in <u>Figure 15. on page 27.</u> The **PORTSEL** signal should be connected to logic high for synchronous serial mode.

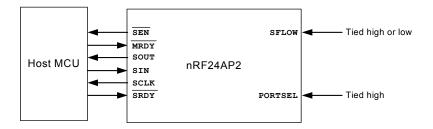


Figure 15. Synchronous mode connections



5.3.2 Flow Control Select (SFLOW)

The Flow Control Select signal is used to configure the synchronous serial port for either Byte or Bit flow control.

SFLOW	Flow control
0	Byte flow control
1	Bit flow control

Please note that Byte flow control assumes that the Host contains synchronous communication hardware which can be configured for synchronous slave communication. Bit flow control can be used by all microcontrollers. It is especially useful for microcontrollers that offer no HW serial interface, and which require the serial interface to be emulated in software on the Host MCU. The differences between byte and bit flow control are detailed in the remaining sections of this chapter.

5.3.3 Synchronous interface handshaking

A basic description of the communications mechanism follows.

- The synchronous serial port provided by nRF24AP2 is a half duplex synchronous master.
- Two handshake signals (SEN, MRDY) are used to set up communication.
- Being a master, the nRF24AP2 will forward all incoming radio messages to the host as they become available.
- The host must request the use of the serial port and get acknowledge from nRF24AP2 before a transaction can take place.
- SRDY enables flow control in both directions.
- The first byte in each message is always sent from the nRF24AP2 and indicates the direction of this message.

The steps needed to initiate synchrnous message transfers in both directions are shown in <u>Figure 16. on page 28.</u>

5.3.3.1 Synchronization

In order for the Host MCU to guarantee synchronization with nRF24AP2 in startup conditions, a reset sequence must be applied to nRF24AP2. This only applies to synchronous mode communication.

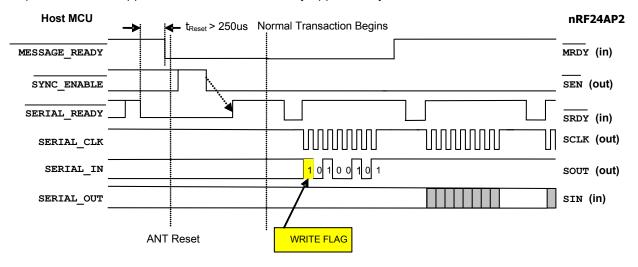


Figure 16. Synchronization with nRF24AP2 upon startup



5.3.3.2 Power up/power down

nRF24AP2 will <u>automatically</u> place itself into idle mode when all radio channels are closed and there is no activity on the <u>MRDY</u> input signal. The Host MCU should ensure these conditions during times that the nRF24AP2 radio is not required in order to maximize product battery life. Upon every power up, the host must apply the Synchronous Reset sequence.

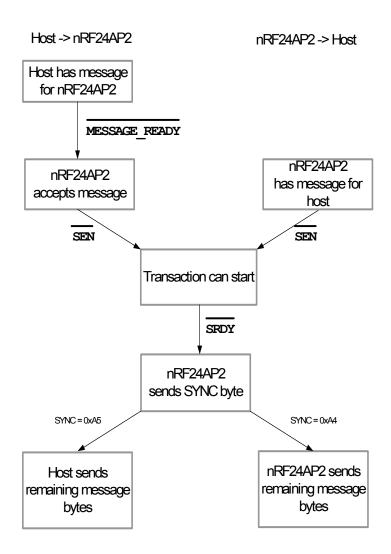


Figure 17. Synchronous serial communication

<u>Figure 17. on page 29</u> and timing diagrams in <u>Figure 18. on page 30</u> and <u>Figure 19. on page 31</u> illustrate the basic, message transaction sequence:

For a message from Host->nRF24AP2:

The Host will assert the MRDY signal indicating it has a message for the nRF24AP2.

For messages in either direction:

nRF24AP2 will assert <u>SEN</u> to indicate the start of a message transfer.



- 2. After SEN has been asserted, the Host will assert SRDY to indicate it is ready for communication.
- 3. After SEN and SRDY are both asserted, nRF24AP2 always transmits the first (for example SYNC) byte. This is output from SOUT, and clocked with SCLK (see chapter 8 on page 40 for details of clock frequency). The LSB of the SYNC byte indicates the direction of the remaining message bytes (0 : Message Receive, nRF24AP2 → Host; 1: Message Transmit, Host → nRF24AP2).
- 4. If the **SYNC** byte indicates a message receive (nRF24AP2->Host), the additional message bytes will be transmitted the same way as the **SYNC** byte.
- 5. If the **SYNC** byte indicates a message transmit (Host->nRF24AP2), the Host must output its data to nRF24AP2 **SIN** at the clock rate provided by nRF24AP2 **SCLK**.

Data is transmitted least-significant-bit (LSB) first.

5.3.4 Synchronous messaging with byte flow control

Byte flow-control mode is used when a synchronous hardware serial port is available.

The Host MCU flow-control signal SRDY must be toggled for each byte and can either be implemented with a software controlled IO line, or in some cases may be controlled by the Host's hardware serial port. Data bits change state on the falling edge of SCLK and are read on the rising edge of SCLK. This is true for transactions in either direction.

The first byte in the transaction sequence is always sent from nRF24AP2 to the Host MCU. The first bit of the first byte dictates the direction for the remaining bytes in the transaction.

<u>Figure 18. on page 30</u> and <u>Figure 19. on page 31</u> show transactions between the Host and nRF24AP2 in byte synchronous mode.

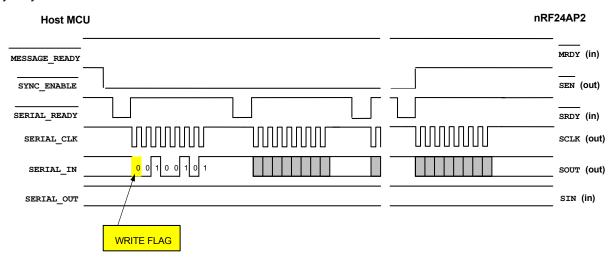


Figure 18. nRF24AP2 → host transaction

The nRF24AP2 asserts **SEN** and waits for the host to assert **SRDY**. Once both **SEN** and **SRDY** have been asserted, nRF24AP2 will send the SYNC byte from **SOUT**.

For hardware <u>SRDY</u>, this signal will be de-asserted on the first <u>SCLK</u> transition, if a software controlled IO line is used for <u>SRDY</u>, it only needs to stay asserted for 2.5 µs minimum before the host can de-assert it again.. The LSB of the SYNC byte will notify the host of <u>the message</u> direction (that is to say, nRF24AP2 - > host), and once ready, the host will once <u>again</u> assert <u>SRDY</u> to receive the next message byte from nRF24AP2. After the last message byte, <u>SRDY</u> must remain de-asserted until the next message

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transaction is requested.

The process for nRF24AP2 to host transactions with software **SRDY** (Figure 18.) is very similar as for hardware **SRDY**. The sole difference is that the host can just pulse **SRDY** and not have to wait until the first SCLK transition.

Host MCU nRF24AP2 MESSAGE_READY MRDY (in) SYNC ENABLE SEN (out) SERIAL READY SRDY (in) SERIAL CLK SCLK (out) 0 1 0 0 1 0 1 SERIAL_IN SOUT (out) SERIAL_OUT SIN (in) READ FLAG

Figure 19. Host transaction → nRF24AP2 transaction

For host to nRF24AP2 transactions with hardware SRDY (See Figure 19. on page 31) the process is very similar. The main difference is that the host first asserts MRDY to inform nRF24AP2 that it wished to send a message. nRF24AP2 will respond by asserting SEN and then waiting for the host to assert SRDY. Once both SEN and SRDY have been asserted, nRF24AP2 will the send the SYNC byte. For hardware SRDY, this signal will be de-asserted on the first SCLK transition. The first bit of the SYNC byte will notify the host of the message direction (meaning host-> nRF24AP2), and the host will once again assert SRDY and then send the next message byte to nRF24AP2 on host SOUT at the rate of SCLK. Again, the hardware SRDY will de-assert on the first SCLK transition and re-assert after each byte until the entire message has been transferred. After the last message byte, SRDY will remain de-asserted until the next message transaction is requested.

The process for host to nRF24AP2 transactions with software SRDY (<u>Figure 19. on page 31</u>) is very similar as for hardware SRDY. The only difference is that the host can pulse SRDY and does not have to wait until the first SCLK transition.



5.3.5 Synchronous timing with byte flow control

Synchronous mode with byte flow is compatible with a Host microcontroller, hardware SPI slave, configured as mode 3 and polarity 1. In <u>Figure 20. on page 32</u> signals to the left indicate pins on the Host MCU. Signals on the right-hand side indicate pins on nRF24AP2. Shaded areas indicate "don't care" values.

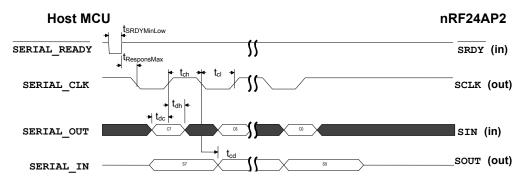


Figure 20. Synchronous byte flow timing

Symbol	Parameter (condition)	Notes	Min	Тур	Max	Units
SCLK _{frequency}	Synchronous clock frequency (byte			500		kHz
, ,	mode)					
t _{dc}	Data to SCK Setup (byte mode)		100			ns
t _{dh}	SCK to Data Hold (byte mode)		20			ns
t _{cd}	SCK to Data Valid (byte mode)				60	ns
t _{cl}	SCK Low Time (byte mode)		900	1000		ns
t _{ch}	SCK High Time (byte mode)		900	1000		ns
t _{SRDY} MinLow			2.5			μs
	Minimum SRDY low time					
t _{Reset}			250			μs
	Synchronous reset. SRDY falling edge					
	to MRDY falling edge					
t _{POR}	Power on reset time (supply rise time	а			2.0	ms
	not included)					
t _{SoftReset}	Software reset (synchronous reset	а			1.5	ms
	suspend reset and reset command)					
t _{ResponseMax}	Time the nRF24AP2 will take to				1.0	ms
l istp shooman	respond to input signal					

a. Defines the time before the Host MCU can start to configure the nRF24AP2 after a reset.

Table 6. Synchronous serial timing

5.3.6 Synchronous messaging with bit flow control

If no hardware serial port is available on the Host MCU, nRF24AP2 can still be controlled using bit flow control. Using this method, the serial lines are implemented with software controlled I/O lines. All of the signaling at the message transaction level remains the same as above. However, instead of pulsing after every byte, SRDY is pulsed for each bit of the message as shown below in Figure 20. on page 32.



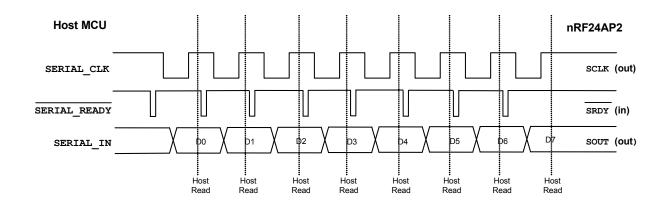


Figure 21. nRF24AP2 → host transaction

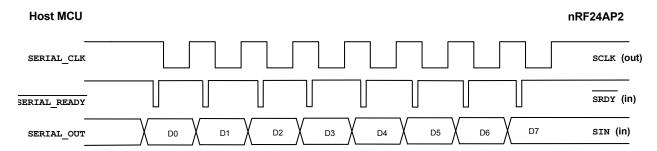


Figure 22. Host → nRF24AP2 transaction

It is important to note that the Host MCU will do all bit processing on the rising edge of the SCLK signal, with the exception being when the byte is being transmitted from the Host MCU to nRF24AP2, where the first data bit will need to be asserted prior to the first clock edge. The final rising edge of the byte transaction will be the event to drive byte processing.

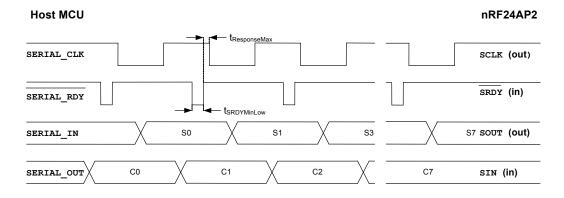


Figure 23. Synchronous bit flow timing



5.3.7 Serial enable control

The **SEN** signal will be asserted by nRF24AP2 prior to all message transmissions. It can therefore be used as a serial port enable signal, which is useful in cases where the Host serial port requires hardware activation.

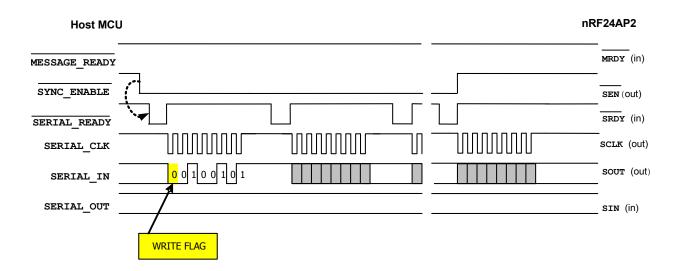


Figure 24. Serial enable control using nRF24AP2

5.3.8 Synchronization

In order for the Host MCU to guarantee synchronization with nRF24AP2 in startup conditions, a reset sequence must be applied to nRF24AP2. This only applies to synchronous mode communication.

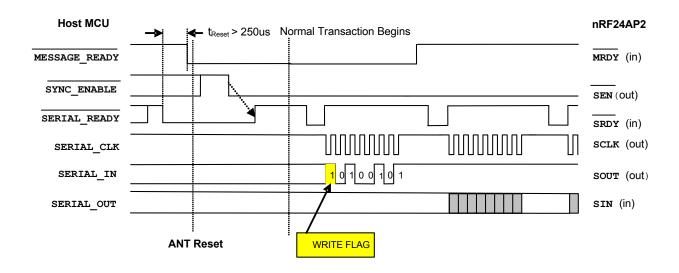


Figure 25. Synchronization with nRF24AP2 upon startup



6 On-chip oscillators

In order to provide the necessary clocks for the ANT protocol stack, nRF24AP2 contains one high frequency oscillator used by the RF transceiver and two optional low frequency oscillators for ANT protocol timing. The mandatory, high frequency clock source must be a 16 MHz crystal oscillator. The low frequency clock source can be generated by a 32.768 kHz crystal oscillator or synthesized 32.768 kHz from the 16 MHz crystal oscillator clock. External 16 MHz and 32.768 kHz clocks may also be used instead of the on-chip oscillators of nRF24AP2. For ultra low-power applications, we recommend you use the 32.768 kHz crystal oscillator or provide a 32.768 kHz clock signal, to achieve the lowest possible current consumption.

6.1 Features

- Low-power, amplitude regulated 16 MHz crystal oscillator
- Ultra low-power amplitude regulated 32.768 kHz crystal oscillator
- Low power, synthesized 32.768 kHz clock from the 16 MHz crystal oscillator

6.2 Block diagrams

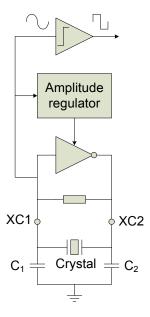


Figure 26. Block diagram of 16 MHz crystal oscillator

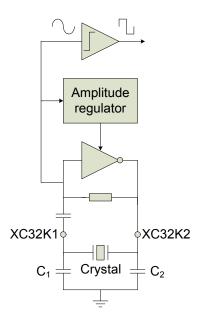


Figure 27. Block diagram of 32.768 kHz crystal oscillator

6.3 Functional description

6.3.1 16 MHz crystal oscillator

The 16 MHz crystal oscillator is designed to be used with an AT-cut quartz crystal in parallel resonant mode. To achieve correct oscillation frequency it is very important that the load capacitance matches the specification in the crystal datasheet. The load capacitance is the total capacitance seen by the crystal across its terminals:

$$C_{LOAD} = \frac{C_{1}^{'} \cdot C_{2}^{'}}{C_{1}^{'} + C_{2}^{'}}$$

$$C_{1}^{'} = C_{1} + C_{PCB1} + C_{PIN}$$

$$C_{2}^{'} = C_{2} + C_{PCB2} + C_{PIN}$$

 C_1 and C_2 are ceramic SMD capacitors connected between each crystal terminal and VSS, C_{PCB1} and C_{PCB2} are stray capacitances on the PCB, while C_{PIN} is the input capacitance on the **xc1** and **xc2** pins of nRF24AP2 (typically 1pF). C_1 and C_2 should be of the same value, or as close as possible.

To ensure a functional radio link the frequency accuracy must be \pm 50 ppm or better. The initial tolerance of the crystal, drift over temperature, aging and frequency pulling due to incorrect load capacitance must all be taken into account. For reliable operation the crystal load capacitance, shunt capacitance, equivalent series resistance (ESR) and drive level must comply with the specifications in <u>Table 9. on page 42</u>. It is recommended to use a crystal with lower than maximum ESR if the load capacitance and/or shunt capacitance is high. This will give faster start-up and lower current consumption.

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The start-up time is typically about 1 ms for a crystal with 9pF load capacitance and an ESR specification of $60~\Omega$ max. This value is valid for crystals in a 3.2x2.5 mm can. If you use the smallest crystal cans (like 2.0x2.5 mm), pay particular attention to the start-up time of the crystal. These crystals have a longer start up than crystals in larger cans. To make sure the start-up time is <1.24 ms use a crystal for load capacitance of 6pF. A low load capacitance will reduce both start-up time and current consumption. For more details regarding how to measure the start up of a specific crystal, please see the nAN24-13 application note. This application note describes measurements on the nRF24LE1, which has an equal crystal oscillator. The start-up time must be measured to <1.5 ms in this setup since it includes a debounce time of $256\mu s$.

6.3.2 External 16 MHz clock

nRF24AP2 may be used with an external 16 MHz clock applied to the **xc1** pin. The input signal must be analog, coming from the crystal oscillator of a microcontroller, for example. An input amplitude of 0.8V peak-to-peak or higher is recommended to achieve low current consumption and a good signal-to-noise ratio. The DC level is not important as long as the applied signal never rises above VDD or drops below VSS. The **xc1** pin will load the microcontroller's crystal with approximately 1pF in addition to PCB routing. **xc2** shall not be connected.

Note: A frequency accuracy of ±50 ppm or better is required to get a functional radio link.

6.3.3 32.768 kHz crystal oscillator

The crystal must be connected between port pins xc32k2 and xc32k1. To achieve correct oscillation frequency it is important that the load capacitance matches the specification in the crystal datasheet. The load capacitance is the total capacitance seen by the crystal across its terminals:

$$C_{LOAD} = \frac{C_{1}^{'} \cdot C_{2}^{'}}{C_{1}^{'} + C_{2}^{'}}$$

$$C_{1}^{'} = C_{1} + C_{PCB1} + C_{PIN}$$

$$C_{2}^{'} = C_{2} + C_{PCR2} + C_{PIN}$$

 C_1 and C_2 are ceramic SMD capacitors connected between each crystal terminal and vss, C_{PCB1} and C_{PCB2} are stray capacitances on the PCB, while C_{PIN} is the input capacitance on the vc32vc2 and vc32vc2 pins of nRF24AP2. vc32vc2 and vc32vc2

Note: A frequency accuracy of ± 50 ppm or better is required to get reliable ANT functionality. The ANT_CrystalEnable() must be executed in order to enable external, crystal oscillator.

6.3.4 Synthesized 32.768 kHz clock

The low frequency clock can also be synthesized from the 16 MHz crystal oscillator clock. This saves the cost of a crystal but increases average power consumption. The synthesized clock is enabled by connecting xc32k1 to VSS and leaving xc32k2 unconnected.



6.3.5 External 32.768 kHz clock

nRF24AP2 may be used with an external 32.768 kHz clock applied to the xc32x1 port pin. The external clock must be a rail-to-rail digital signal. xc32x2 must not be connected.

Note: A frequency accuracy of ±50 ppm or better is required to get reliable ANT functionality. The ANT_CrystalEnable() must be executed in order to enable external, clock.



7 Operating conditions

Symbol	Parameter	Notes	Min.	Тур.	Max.	Units
VDD	Supply voltage		1.9	3.0	3.6	V
t _{R_VDD}	Supply rise time (0V to 1.9V)	а	1 µs		50 ms	
T _A	Operating temperature		-40		+85	°C

a. The power-on reset circuitry may not function properly for rise times outside the specified interval.

Table 7. Operating conditions



8 Electrical specifications

This section contains electrical and timing specifications.

Conditions: VDD = 3.0V, $T_A = -40$ °C to +85°C (unless otherwise noted)

Symbol	Parameter (condition)	Notes	Min.	Тур.	Max.	Units
V _{IH}	Input high voltage		0.7·VDD		VDD	V
V _{IL}	Input low voltage		VSS		0.3·VDD	V
V _{OH}	Output high voltage (I _{OH} =0.5mA)		VDD-0.3		VDD	V
V _{OL}	Output low voltage (I _{OH} =0.5mA)		VSS		0.3	V

Table 8. Digital inputs/outputs



General RF conditions f_OP Operating frequency a 2400 2403-2480 PLL_{res} PLL Programming resolution 1 f_XTAL Crystal frequency 16 Δf Frequency deviation ±160 R_GFSK Air data rate b 1000 F_CHANNEL Non-overlapping channel spacing c 1 Current consumption	2483.5	MHz MHz MHz kHz kbps MHz
PLL _{res} PLL Programming resolution 1 f _{XTAL} Crystal frequency 16 Δf Frequency deviation ±160 R _{GFSK} Air data rate b 1000 F _{CHANNEL} Non-overlapping channel spacing c 1 Current consumption I _{DeepSleep} Deep Sleep Command 0.5 I _{Idle} No active channels — No communications 2.0 I _{PeakRX} Peak RX Current d 17 I _{PeakRX} Peak RX Current at 0 dBm e 15 I _{PeakTX-6} Peak TX Current at -6 dBm e 13 I _{PeakTX-12} Peak TX Current at -12 dBm e 11 I _{PeakTX-18} Peak TX Current at -18 dBm e 11 Transmitter operation P _{RF} Maximum output power f 0	2483.5	MHz MHz kHz kbps MHz
f _{XTAL} Crystal frequency 16 Δf Frequency deviation ±160 R _{GFSK} Air data rate b 1000 F _{CHANNEL} Non-overlapping channel spacing c 1 Current consumption I _{DeepSleep} Deep Sleep Command 0.5 I _{Idle} No active channels — No communications 2.0 I _{PeakRX} Peak RX Current d 17 I _{PeakRX} Peak RX Current at 0 dBm e 15 I _{PeakTX-6} Peak TX Current at -6 dBm e 13 I _{PeakTX-12} Peak TX Current at -12 dBm e 11 I _{PeakTX-18} Peak TX Current at -18 dBm e 11 Transmitter operation P _{RF} Maximum output power f 0		MHz kHz kbps MHz
f _{XTAL} Crystal frequency 16 Δf Frequency deviation ±160 R _{GFSK} Air data rate b 1000 F _{CHANNEL} Non-overlapping channel spacing c 1 Current consumption I _{DeepSleep} Deep Sleep Command 0.5 I _{Idle} No active channels — No communications 2.0 I _{PeakRX} Peak RX Current d 17 I _{PeakRX} Peak RX Current at 0 dBm e 15 I _{PeakTX-6} Peak TX Current at -6 dBm e 13 I _{PeakTX-12} Peak TX Current at -12 dBm e 11 I _{PeakTX-18} Peak TX Current at -18 dBm e 11 Transmitter operation P _{RF} Maximum output power f 0		kHz kbps MHz µA µA mA mA mA
RGFSK Air data rate b 1000 FCHANNEL Non-overlapping channel spacing c 1 Current consumption IDeepSleep Deep Sleep Command 0.5 IIdle No active channels — No communications 2.0 IPeakRX Peak RX Current d 17 IPeakTX Peak RX Current at 0 dBm e 15 IPeakTX-6 Peak TX Current at -6 dBm e 13 IPeakTX-12 Peak TX Current at -12 dBm e 11 IPeakTX-18 Peak TX Current at -18 dBm e 11 Transmitter operation PRF Maximum output power f 0		kbps MHz μA μA mA mA mA
F_CHANNEL Non-overlapping channel spacing C		MHz μΑ μΑ mA mA mA mA
Current consumption IDeepSleep Deep Sleep Command 0.5 IIdle No active channels — No communications 2.0 IPeakRX Peak RX Current d 17 IPeakTX Peak TX Current at 0 dBm e 15 IPeakTX-6 Peak TX Current at -6 dBm e 13 IPeakTX-12 Peak TX Current at -12 dBm e 11 IPeakTX-18 Peak TX Current at -18 dBm e 11 Transmitter operation PRF Maximum output power f 0		μΑ μΑ mA mA mA
I DeepSleep Deep Sleep Command 0.5 I I No active channels — No communications 2.0 I I PeakRX Peak RX Current d 17 I PeakTX Peak TX Current at 0 dBm e 15 I I PeakTX-6 Peak TX Current at -6 dBm e 13 I I PeakTX-12 Peak TX Current at -12 dBm e 11 I I PeakTX-18 Peak TX Current at -18 dBm e 11 Transmitter operation Transmitter operation PRF Maximum output power f 0		mA mA mA mA
I No active channels — No communications 2.0 I I No active channels — No communications 2.0 I I PeakRX		mA mA mA mA
tions d 17 I _{PeakRX} Peak RX Current d 17 I _{PeakTX} Peak TX Current at 0 dBm e 15 I _{PeakTX-6} Peak TX Current at -6 dBm e 13 I _{PeakTX-12} Peak TX Current at -12 dBm e 11 I _{PeakTX-18} Peak TX Current at -18 dBm e 11 Transmitter operation P _{RF} Maximum output power f 0		mA mA mA
I _{PeakRX}		mA mA mA
I_{PeakTX}		mA mA mA
I _{PeakTX-6}		mA mA
I _{PeakTX-12}		mA
I _{PeakTX-18} Peak TX Current at -18 dBm e 11 Transmitter operation P _{RF} Maximum output power f 0		
Transmitter operation P _{RF} Maximum output power f 0		11117
P _{RF} Maximum output power f 0		, \
	+4	dBm
FREC IN power controllaringe I I I I I I I I	20	dB
14.0	±4	dB
PRFCR RF power accuracy PBW1 20dB bandwidth for modulated carrier 950	1100	kHz
DW1	-20	dBc
P _{RF1.1} 1 st Adjacent Channel Transmit Power 1 MHz	-20	ubc
P _{RF2.1} 2 nd Adjacent Channel Transmit Power	-40	dBc
2 MHz		
Receiver operation		
RX _{MAX} Maximum received signal at < 0.1% 0 BER		dBm
RX _{SENS} Sensitivity (0.1% BER) -85		dBm
RX selectivity according to ETSI EN 300 440-1 V1.3.1 (2001-09) page 27		
C/I _{CO} C/I co-channel 9		dBc
C/I _{1ST} 1 st ACS, C/I 1 MHz		dBc
C/I _{2ND} 2 nd ACS, C/I 2 MHz -20		dBc
C/I _{3RD} 3 rd ACS, C/I 3 MHz -30		dBc
C/I_{Nth} N^{th} ACS, C/I $f_i > 6$ MHz		dBc
C/I_{Nth} N^{th} ACS, C/I $f_i > 25$ MHz		dBc
RX selectivity with nRF24AP2 equal modulation on interfering signal (Pin = -67d	Bm for w	anted
signal)		-
C/I _{CO} C/I co-channel 12		dBc
C/I _{1ST} 1 st ACS, C/I 1 MHz		dBc
C/I _{2ND} 2 nd ACS, C/I 2 MHz -21		dBc
C/I _{3RD} 3 rd ACS, C/I 3 MHz -30		dBc
C/I_{Nth} N^{th} ACS, C/I $f_i > 6$ MHz		dBc
C/I_{Nth} N^{th} ACS, C/I $f_i > 25$ MHz		dBc

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Symbol	Parameter (condition)	Notes	Min.	Тур.	Max.	Units
RX intermod 2004, page 4	ulation performance in line with Blue 2	tooth spe	cification	version 2.0	, 4 th Nove	ember
P_IM(3)	Input power of IM interferers at 3 and 6 MHz distance from wanted signal	g		-36		dBm
P_IM(4)	Input power of IM interferers at 4 and 8 MHz distance from wanted signal	g		-36		dBm
P_IM(5)	Input power of IM interferers at 5 and 10 MHz distance from wanted signal	g		-36		dBm

- a. Usable band is determined by local regulations.
- b. Data rate in each burst on-air.
- c. The minimum channel spacing is 1 MHz.
- d. Time of Maximum Current consumption in RX is typical 500 µs and maximum 1 ms.
- e. Time of maximum TX Only Current is typical 300 μs and maximum 350 μs .
- f. Antenna load impedance = 15 Ω + j88 Ω .
- g. Wanted signal level at Pin=64 dBm. Two interferers with equal input power are used. The interferer closest in frequency is unmodulated, the other interferer is modulated equal to the wanted signal. The input power of interferers where the sensitivity equals BER=0.1% is presented.

Table 9. Electrical characteristics

Symbol	Parameter (condition)	Notes	Min.	Тур.	Max.	Unit
	16 MHz crystal					
f _{NOM}	Nominal frequency (parallel resonant)			16.000		MHz
f _{TOL}	Frequency tolerance	а			±50	ppm
C _L	Load capacitance			9	16	pF
C ₀	Shunt capacitance			3	7	pF
ESR	Equivalent series resistance			50	100	Ω
P_{D}	Drive level				100	μW
T _{START}	Required 16 MHz oscillator startup time	b			1.24	ms
	32.768 kHz crystal					
f _{TOL}	Frequency tolerance				±50	ppm
f _{NOM}	Crystal frequency (parallel resonant)			32.768		kHz
C _L	Load capacitance			9	12.5	pF
C ₀	Shunt capacitance			1	2	pF
ESR	Equivalent series resistance			50	80	kΩ
P_{D}	Drive level				1	μW

a. Includes initial accuracy, stability over temperature, aging and frequency pulling due to incorrect load capacitance.

Table 10. External circuitry specification

b. Crystal oscillator start up time must not exceed 1.24 ms. Please see section 6.3.1 on page 36.



8.1 Application-specific current consumption

The power nRF24AP2 consumes depends on the configuration of nRF24AP2, in specific what you use in the way of serial interface, channel period, master-slave operation and broadcast-, acknowledge- or burst data.

Conditions: VDD = 3.0V, TA = +25°C

The next table shows peak and average current consumption for typical applications and interfaces.

Symbol	Parameter (condition)	Notes	Min.	Тур.	Max.	Units
I _{DeepSleep}	Deep Sleep Command			0.5		μA
I _{Idle}	No active channels—no communications			2.0		μA
I _{Suspend}	Asynchronous suspend activated			2.0		μA
I _{Base_32kXO}	Base active current (32.768 kHz crys-			3.0		μA
-base_32kAO	tal oscillator or 32.768 kHz external clock source)					P
I _{Base_32kSynt}	Base active current (synthesized 32.768 kHz from 16 MHz)			87		μA
I _{Search}	Search current			2.9		mA
					1	
I _{Msq_Rx_ByteSync}	Average current per Rx message in byte sync mode			18		μA
I _{Msq_Rx_BitSync}	Average current per Rx message in bit sync mode			25		μA
I _{Msg_Rx_57600}	Average current per Rx message in async mode at 57600 baud					μA
I _{Msg_Rx_50000}	Average current per Rx message in async mode at 50000 baud			μA		
I _{Msg_Rx_38400}	1 5 1 0 5			μA		
I _{Msg_Rx_19200}	Average current per Rx message in async mode at 19200 baud	rerage current per Rx message in 31			μA	
I _{Msg_Rx_9600}	Average current per Rx message in async mode at 9600 baud			48		μA
I _{Msg_Rx_4800}	Average current per Rx message in async mode at 4800 baud			83		μA
						1 4
IMsg_TxAck _ByteSync	Average current per Acknowledged Tx message in byte sync mode			30		μA
I _{Msg_TxAck} _BitSync	Average current per Acknowledged Tx message in bit sync mode			42		μA
I _{Msg_TxAck} _57600	Average current per Acknowledged Tx message at 57600 baud			44		μA
I _{Msg_TxAck} 50000	Average current per Acknowledged Tx message at 50000 baud			41		μA
I _{Msg_TxAck} _38400	Average current per Acknowledged Tx message at 38400 baud			43		μA
I _{Msg_TxAck} 19200	Average current per Acknowledged Tx message at 19200 baud			56		μA
I _{Msg_TxAck}	Average current per Acknowledged Tx message at 9600 baud			78		μΑ

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Symbol	Parameter (condition)	Notes	Min.	Тур.	Max.	Units
I _{Msg_TxAck}	Average current/Acknowledged Tx			132		μA
_4800	message at 4800 baud					
	[A	T	T	04		
I _{Msg_RxAck}	Average current/Acknowledged Rx			21		μA
_ByteSync	message in byte sync mode Average current/Acknowledged Rx			35		μA
Msg_RxAck	message in bit sync mode			35		μΑ
_BitSync	Average current/Acknowledged Rx			25		μA
I _{Msg_RxAck} _57600	message at 57600 baud					μ, τ
I _{Msg_RxAck}	Average current/Acknowledged Rx			24		μA
_50000	message at 50000 baud					
I _{Msg_RxAck}	Average current/Acknowledged Rx			27		μA
_38400	message at 38400 baud					
I _{Msg_RxAck}	Average current/Acknowledged Rx			34		μA
_19200	message at 19200 baud					
I _{Msg_RxAck}	Average current/Acknowledged Rx			53		μA
_9600	message at 9600 baud Average current/Acknowledged Rx			86		
I _{Msg_RxAck}	message at 4800 baud			80		μA
_4800	message at 4000 badd					
Mag Ty Dyta Cyma	Average current/Tx-only message in	а		14		μA
I _{Msg_Tx_ByteSync}	byte sync mode					μ., .
I _{Msg_Tx_BitSync}	Average current/Tx-only message in	а		21		μA
Wiog_TX_BitOyrio	bit sync mode					
I _{Msg_Tx_57600}	Average current/Tx-only message in	а		25		μA
0	async mode at 57600 baud					
I _{Msg_Tx_50000}	Average current/Tx-only message in	а		20		μA
	async mode at 50000 baud					
I _{Msg_Tx_38400}	Average current/Tx-only message in	а		24		μA
	async mode at 38400 baud					
I _{Msg_Tx_19200}	Average current/Tx-only message in	а		31		μA
	async mode at 19200 baud					
I _{Msg_Tx_9600}	Average current/Tx-only message in	а		63		μA
	async mode at 9600 baud			108		
^I Msg_Tx_4800	Average current/Tx-only message in async mode at 4800 baud	а		100		μA
	async mode at 4000 badd					
I _{Msg_TR_ByteSync}	Average current/Tx message in byte			24		μA
Wisg_TK_byte3yiic	sync mode					
I _{Msg_TR_BitSync}	Average current/Tx message in bit			33		μA
9	sync mode					
I _{Msg_TR_57600}	Average current/Tx message in			36		μA
_	async mode at 57600 baud					
I _{Msg_TR_50000}	Average current/Tx message in			33		μA
1	async mode at 50000 baud			25		
I _{Msg_TR_38400}	Average current/Tx message in async mode at 38400 baud			35		μA
	Average current/Tx message in			42		μA
^I Msg_TR_19200	async mode at 19200 baud			74		μΛ
I _{Msg_TR_9600}	Average current/Tx message in			71		μA
- way_1 K_9000	async mode at 9600 baud					P., .
	ı •	i	i	i	1	

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Symbol	Parameter (condition)	Notes	Min.	Тур.	Max.	Units
I _{Msg_TR_4800}	Average current/Tx message in async mode at 4800 baud			120		μA
I _{Ave}	Broadcast Tx at 0.5 Hz in byte sync mode	b		12		μA
I _{Ave}	mode			48		μA
I _{Ave} Broadcast Rx at 0.5 Hz in byte sync mode		b		9.0		μA
I _{Ave}	Broadcast Rx at 2 Hz in byte sync mode	b		36		μA
I _{Ave}	Acknowledged TX at 0.5 Hz in byte sync mode	b		15		μA
I _{Ave}	-			60		μA
I _{Ave}	Acknowledged RX at 0.5 Hz in byte sync mode	b		10		μA
I _{Ave}	Acknowledged RX at 2 Hz in byte sync mode	b		42		μA
I _{Ave}	Burst continuous at 20 kbps in byte sync mode			4.8		mA
I _{Ave}	Burst continuous at 7.5 kbps in bit sync mode			4.0		mA
I _{Ave}	Burst continuous at 20 kbps in async mode at 57600 baud			5.9		mA
I _{Ave}			4.9		mA	
I _{Ave}	Burst continuous at 13.8 kbps in async mode at 38400 baud			4.7		mA
I _{Ave}	Burst continuous at 8.4 kbps in async mode at 19200 baud			4.2		mA
I _{PeakRX}	Peak RX current	С		17		mA
I _{PeakTX}	Peak TX current at 0 dBm	d		15		mA

a. Transmit only operation provides no ANT channel management across the air channel and is not recommended for normal operation.

Table 11. Average current consumption for typical applications and interfaces

b. Does not include base current. See $I_{\mbox{Ave}}$ examples below.

c. Time of Maximum Current consumption in RX is typical 500 μs and maximum 1 ms.

d. Time of maximum TX Only Current is typical 300 μs and maximum 350 μs .



8.2 Current calculations examples

By using the values in <u>Table 11. on page 46</u> together with the formulas presented in this section, you can calculate the current consumption for a specific application setup. Channel period is defined as the number of data packets received or transmitted each second.

1. Master channel with Broadcast data at 0.5 Hz with a byte synchronous serial interface using a 32.768 kHz external clock source.

$$I_{Ave}$$
 = ($I_{Msg_Tx_ByteSync}$ * Message_Rate) + I_{Base_32kXO}
= (14 μA/message * 0.5 message) + 3 μA
= 10 μA

2. Receive channel with Acknowledged data at 2 Hz with an asynchronous serial interface at 57600 baud using a 32.768 KHz external clock source.

$$I_{Ave}$$
 = ($I_{Msg_RxAck_57600}^*$ Message_Rate) + I_{Base_32kXO}
= (24 μ A/message * 2 messages) + 3 μ A
= 51 μ A

3. Transmit channel at 2 Hz with an asynchronous serial interface at 50000 baud using the internal clock source .

$$I_{Ave} = (I_{Msg_TR_50000} * Message_Rate) + I_{Base_32kSynt}$$

= (33 µA/message * 2 messages) + 87 µA
= 153 µA



9 Absolute maximum ratings

Maximum ratings are the extreme limits to which nRF24AP2 can be exposed without permanently damaging it. Exposure to absolute maximum ratings for prolonged periods of time may affect device reliability.

The device is not guaranteed to operate properly at the maximum ratings.

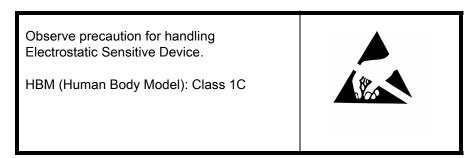
Operating conditions	Minimum	Maximum	Units				
	Supply voltage	es					
VDD	-0.3	+3.6	V				
VSS		0	V				
I/O pin voltage							
V_{IO}	-0.3	VDD +0.3,	V				
		max 3.6					
Temperatures							
Operating temperature	-40	+85	°C				
Storage temperature ^a	-40	+85	°C				

 a. The device can withstand up to 125°C for short periods without damage. Recommended long-time storage temperature <65°C.

Table 12. Absolute maximum ratings

Note: Stress exceeding one or more of the limiting values may cause permanent damage to the device.

Attention!





10 Mechanical specification

nRF24AP2 is packaged in the following QFN-package:

• QFN32 5 x 5 x 0.85 mm, 0.5 mm pitch.

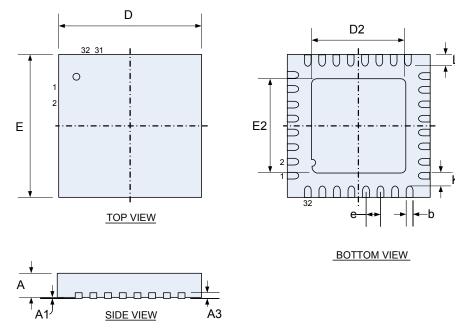


Figure 28. QFN32 pin 5x5mm

Package	Α	A1	A3	b	D, E	D2, E2	е	K	L	
QFN32	0.80	0.00		0.18	4.9	3.20		0.20	0.35	Min
	0.85	0.02	0.20	0.25	5.0	3.30	0.5		0.40	Тур
	0.90	0.05		0.30	5.1	3.40			0.45	Max

Table 13. QFN32 dimensions in mm



11 Application example

This chapter presents schematics, layouts and BOM for each of the serial interface options that are available.

To ensure optimal performance it is essential that you follow the references both on schematics and layout closely. Especially in the antenna matching circuitry (components between device pins ANT1, ANT2, VDD_PA and the antenna) any changes of the layout can change the behavior, resulting in degradation in RF performance or need for changes in component values. All the reference circuits are designed for use with a 50 Ohm single end antenna.

11.1 PCB guidelines

A well designed PCB is necessary to achieve good RF performance. A poor layout can lead to loss of performance or functionality. A fully qualified RF-layout for the nRF24AP2 and its surrounding components, including matching networks, can be downloaded from www.nordicsemi.no.

A PCB with a minimum of two layers including a ground plane is recommended for optimum performance. The nRF24AP2 DC supply voltage should be decoupled as close as possible to the VDD pins with high performance RF capacitors. See the schematics layout in Chapter 11 on page 50 for recommended decoupling capacitor values. The nRF24AP2 supply voltage should be filtered and routed separately from the supply voltages of any digital circuitry.

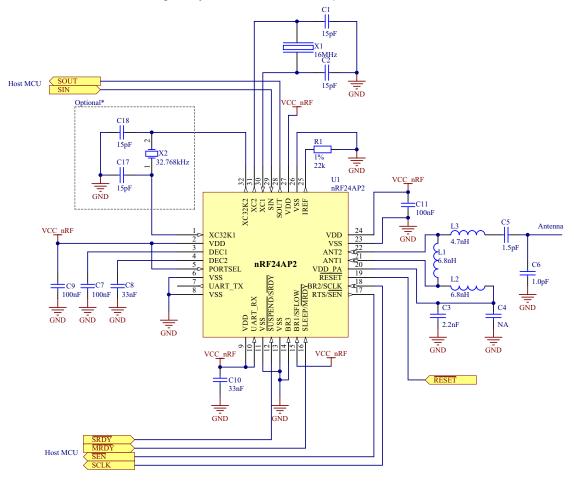
Long power supply lines on the PCB should be avoided. All device grounds, VDD connections and VDD bypass capacitors must be connected as close as possible to the nRF24AP2 IC. For a PCB with a topside RF ground plane, the VSS pins should be connected directly to the ground plane. For a PCB with a bottom ground plane, the best technique is to have via holes as close as possible to the VSS pads. A minimum of one via hole should be used for each VSS pin.

Full swing digital data or control signals should not be routed close to the crystal or the power supply lines.



11.2 Synchronous (bit) mode schematics

Figure 29. on page 51 shows that all interface signals are connected directly to I/O pins on the microcontroller. SCLK and SEN need to be on interrupt-capable I/O pins on the microcontroller. RESET allows optional control of the RESET signal by a microcontroller I/O pin.

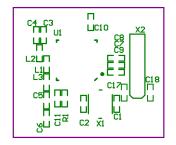


^{*)} For low-frequency oscillator options, refer to chapter 6 on page 35.

Figure 29. Synchronous (bit) mode schematics

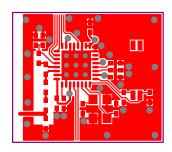


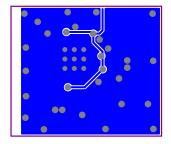
11.2.1 Layout



No components in bottom layer

Top silk screen



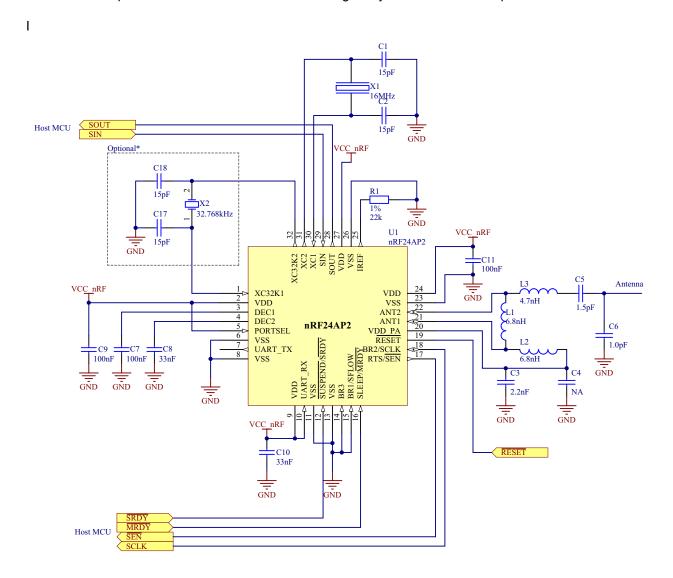


Top view Bottom view



11.3 Synchronous (byte) mode schematics

Figure 30. on page 53 shows how the pins SOUT and SIN are connected directly to hardware USART (SPI) of the microcontroller. SCLK and SEN need to be on interrupt-capable I/O pins on the microcontroller. RESET allows optional control of the module RESET signal by a microcontroll I/O pin.

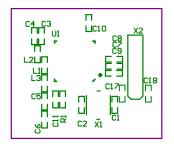


^{*)} For low-frequency oscillator options, refer to chapter 6 on page 35.

Figure 30. Synchronous (byte) mode schematics

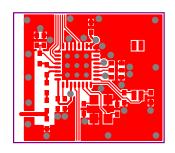


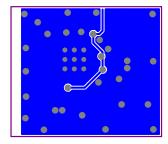
11.3.1 Layout



No components in bottom layer

Top silk screen



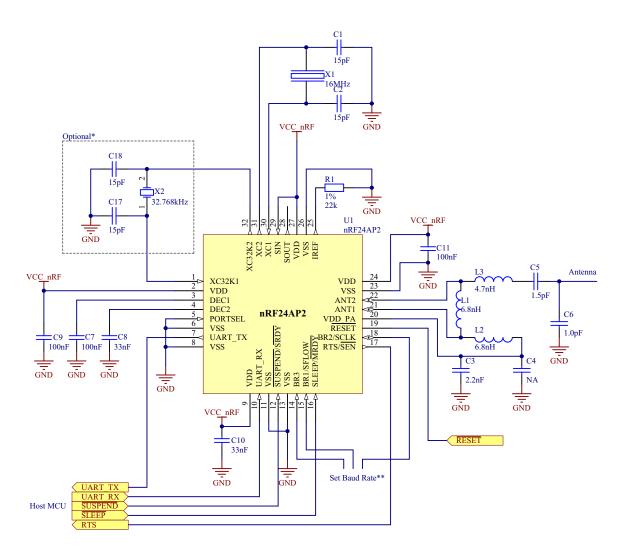


Top view Bottom view



11.4 Asynchronous mode schematics

Figure 31. on page 55 shows that pins UART_TX and UART_RX are directly connected to hardware USART (UART) of the microcontroller. The illustrated baud rate selection pins (BR1, BR2, and BR3) define UART baud rate. The Baud rate selection pins may be connected directly to the logic level of interest. RESET allows optional control of the RESET signal by a microcontroller I/O pin.



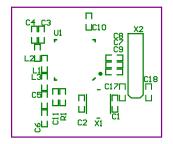
^{*)} For low-frequency oscillation options, refer to chapter 6 on page 35.

Figure 31. Asynchronous mode schematics

^{**)} Refer to Table 5 on page 25 for selectable baud rates.

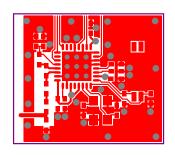


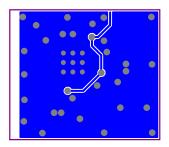
11.4.1 Layout



No components in bottom layer

Top silk screen





Top view

Bottom view

11.5 Bill Of Materials (BOM)

Please refer to the next table to see Bill of Materials for the schematics and layout presented in this document.

Designator	Value	Footprint	Comment
C1, C2, C17,	15pF	0402	NP0 ± 2%
C18			
C3	2.2nF	0402	X7R ± 10%
C4	Not mounted	0402	
C5	1.5pF	0402	NP0 ± 0.1pF
C6	1.0pF	0402	NP0 ± 0.1pF
C7, C9, C11	100nF	0402	X7R ± 10%
C8, C10	33nF	0402	X7R ± 10%
L1, L2	6.8nH	0402	High frequency chip inductor ± 5%
L3	4.7nH	0402	High frequency chip inductor ± 5%
R1	22 kΩ	0402	1%
U1	nRF24AP2	QFN32	QFN32 5x5 mm package
X1	16 MHz	3.2 x 2.4	SMD-3225, 16 MHz, CL=9pF, ±50 ppm
X2	32.768 kHz	7.0 x 1.5	32.768 kHz CL=9pF, ±50 ppm
PCB substrate	FR4 laminate	17.5 x 16	2-layer, 1.6 mm thickness



12 Ordering information

12.1 Package marking

Z	R	F		В	Χ
2	4	Α	Ρ	2	Ζ
Υ	Υ	W	W	L	L

12.1.1 Abbreviations

Abbreviation	Definition
24AP2	Product number
В	Build Code, that is, unique code for production sites, package type and test platform
X	"X" grade, that is, Engineering Samples (optional)
WW	Two-digit week number
LL	Two-letter wafer-lot number code
Z	Product identifier: O = nRF24AP2-1CH, E = nRF24AP2-8CH
YY	Two-digit year number

Table 14. Abbreviations

12.2 Product options

12.2.1 RF silicon

Ordering code	Product	Package	Container	MOQ
nRF24AP2-1CHQ32-T	nRF24AP2-1CH	5x5mm 32-pin QFN	Tray	490
	Single chip ANT solution with 1 ANT channel			
nRF24AP2-1CHQ32-R7	nRF24AP2-1CH	5x5mm 32-pin QFN	Tape-and-reel	1500
	Single chip ANT solution with 1 ANT channel			
nRF24AP2-1CHQ32-R	nRF24AP2-1CH	5x5mm 32-pin QFN	Tape-and-reel	4000
	Single chip ANT solution with 1 ANT channel			
nRF24AP2-1CHQ32-S	nRF24AP2-1CH	5x5mm 32-pin QFN	Sample box	5
	Single chip ANT solution with 1 ANT channel			
nRF24AP2-8CHQ32-T	nRF24AP2-8CH	5x5mm 32-pin QFN	Tray	490
	Single chip ANT solution with 8 ANT channels			
nRF24AP2-8CHQ32-R7	nRF24AP2-8CH	5x5mm 32-pin QFN	Tape-and-reel	1500
	Single chip ANT solution with 8 ANT channels			
nRF24AP2-8CHQ32-R	nRF24AP2-8CH	5x5mm 32-pin QFN	Tape-and-reel	4000
	Single chip ANT solution with 8 ANT channels			
nRF24AP2-8CHQ32-S	nRF24AP2-8CH	5x5mm 32-pin QFN	Sample box	5
	Single chip ANT solution with 8 ANT channels			

Table 15. nRF24AP2 RF silicon options



12.2.2 Development tools

Type Number	Description
nRF24AP1-DK3	ANT Development Kit ^a
nRF24AP2-UPGRADE	ANT Upgrade Kit with two nRF24AP2 modules
	(requires an nRF24AP1-DK3 kit)

a. The ANT Development Kit does not contain modules with nRF24AP2.

Table 16. nRF24AP2 solution options



13 Glossary

Term	Description
ANT TM	Wireless Personal Network protocol from Dynastream Innovations Inc.
ANT+	ANT+ represents the interoperability function that can be added to the base ANT pro-
	tocol. ANT+ refers to device profiles with specific data formats, channel parameters
	and a network key.
BER	Bit Error Rate
GFSK	Gaussian Frequency-Shift Keying
Independent	Each ANT channel is set up with its own channel configuration and this configuration
channel	is independent of other ANT channels (meaning each ANT channel can be set up
	with its own unique, channel type and channel period with unique frequency and net-
	work-key, without considering any other node in the network.)
ISM	Industrial-Scientific-Medical
MCU	MicroController Unit
MOQ	Minimum Order Quantity
OSI	Open Systems Interconnection
PA	Power Amplifier
PCB	Printed Circuit Board
QFN	Quad Flat package. No leads
RFID	Radio Frequency Identification
TDMA	Time Division Multiple Access

Table 17. Glossary