

## 4. TEST REPORT

### 4.1 RF Power Measurements

Figure 4-1 shows the test equipment setup for the RF power measurements.

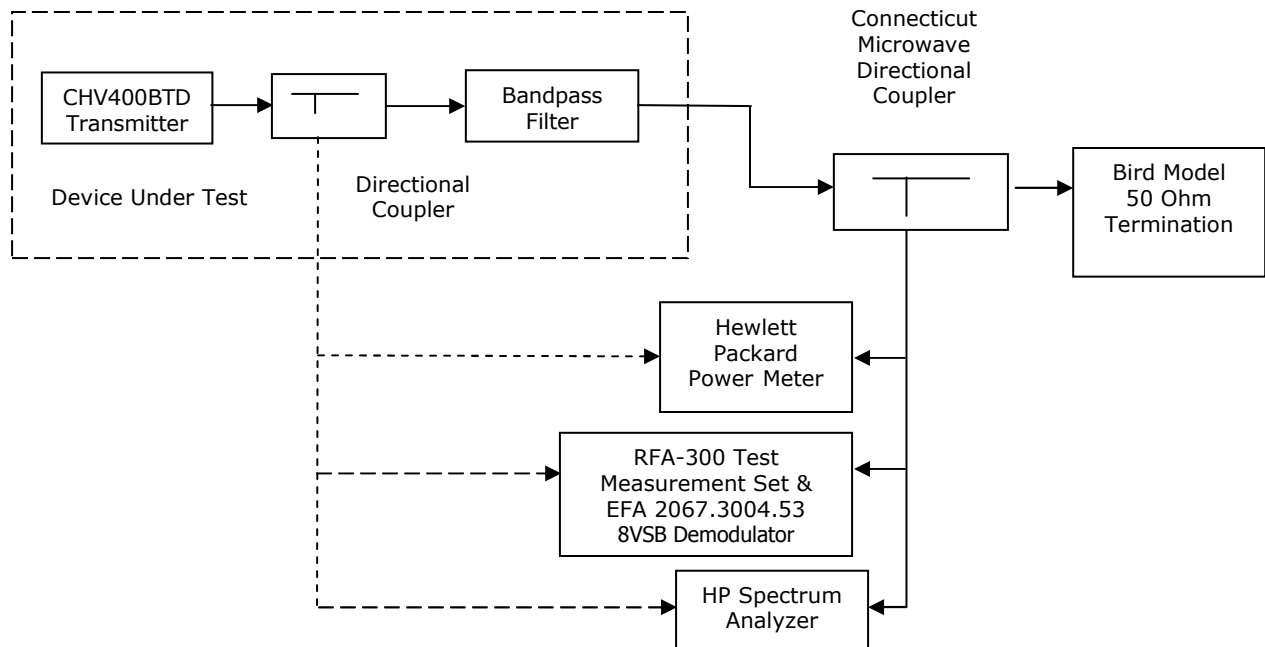


Figure 4-1. Test Equipment Setup for RF Power Measurements

The output power of the CHV400BTD was adjusted to obtain 400 watts average RF output as observed on the power meter. At this power level, the final

Measured Power: +36 dBm  
 Coupling Loss: -20.0 dB  
 Power: +56 dBm  
 Power = 400 Watts

With the power level properly set to 400 watts average, all required tests were performed and recorded in the following sections.

## 4. 2 Modulation Characteristics

The modulator tray incorporates a modulation technique known as 8-Level Vestigial Side-band (8-VSB), which uses a layered digital architecture and a single carrier frequency. A pilot tone is provided, to allow rapid acquisition of the signal by receivers. The 8-VSB system transmits a serial data bit stream at a rate of 19.4 Mbps in a 6 MHz television channel. This type of transmission is far less susceptible to propagation impairments such as multi-path, noise and interference as compared to analog transmissions.

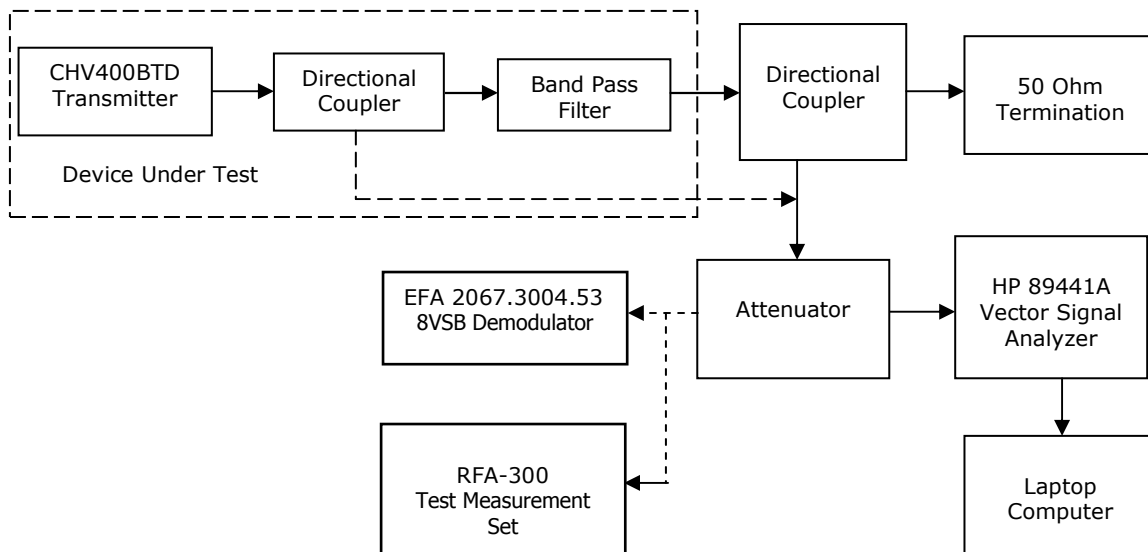


Figure 4-2. Typical Demodulation Test Setup

**4. 3 Signal to Noise Ratio (SNR)**

The Hewlett Packard Vector Signal Analyzer was used to measure that the un-equalized patterns are within the ATSC/FCC SNR limit of 27db.

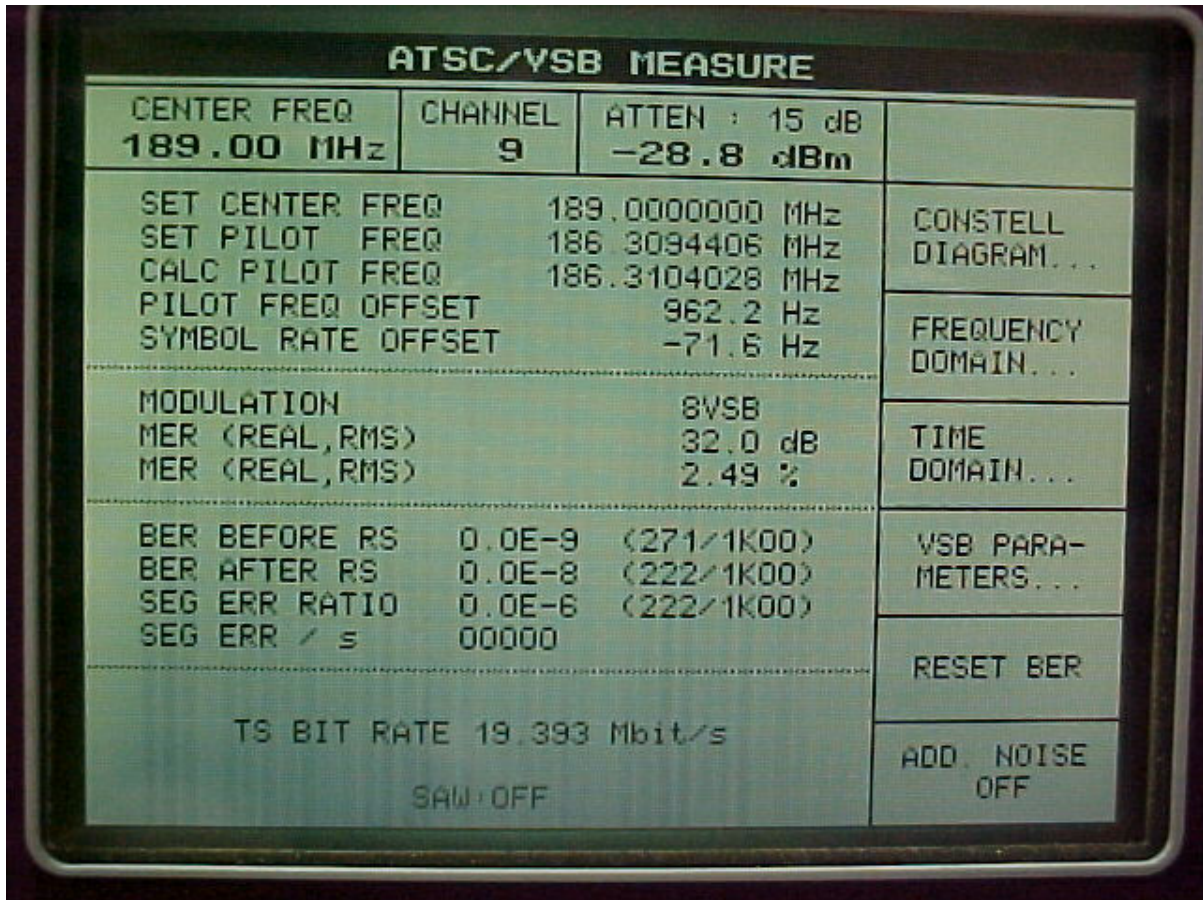


Figure 4-3. SNR

#### 4. 4 Frequency Response

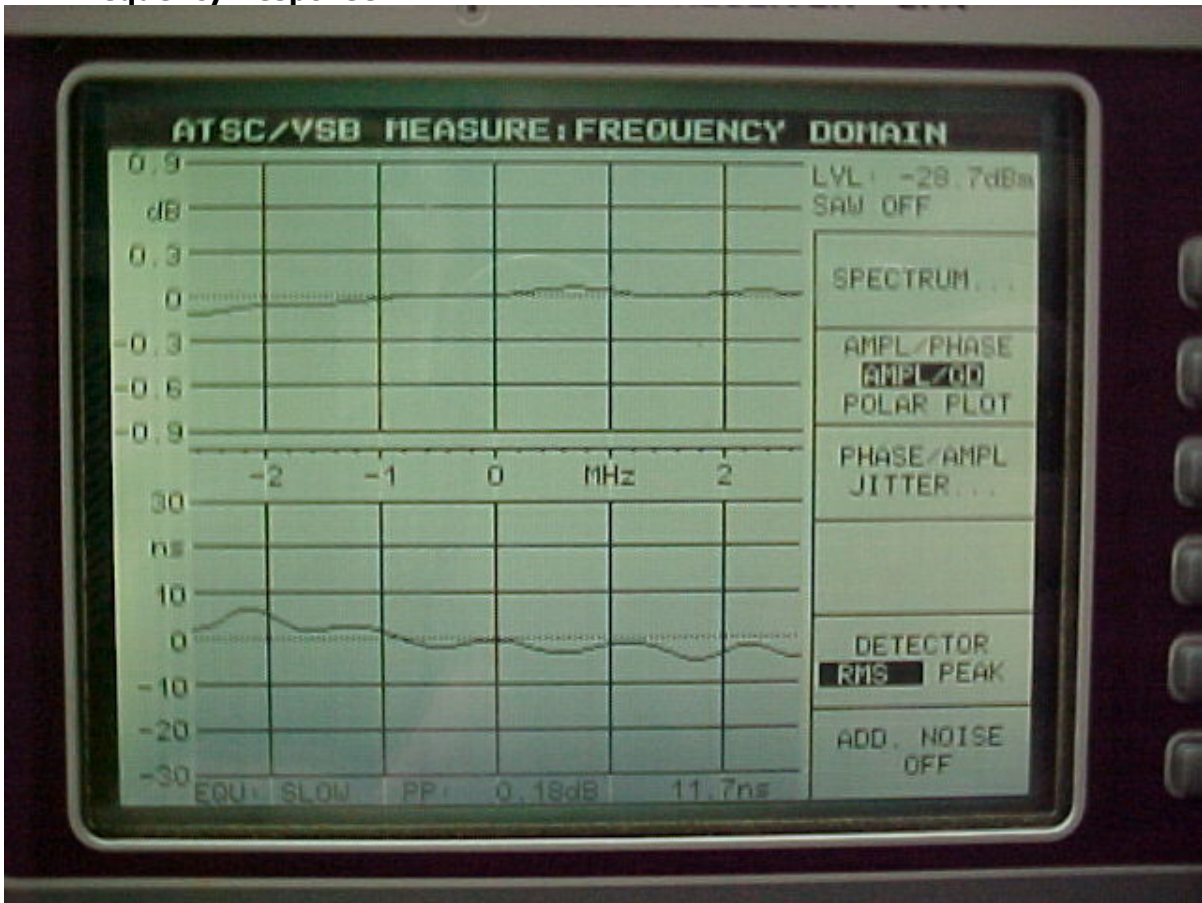


Figure 4-4. Frequency Response Plot



#### 4.5 Occupied Bandwidth

Using the test setup in Figure 4-2, with the transmitter operating at maximum power, a photograph of the transmitter occupied bandwidth spectrum was taken and is shown in Figure 4-7A and 4-7B.

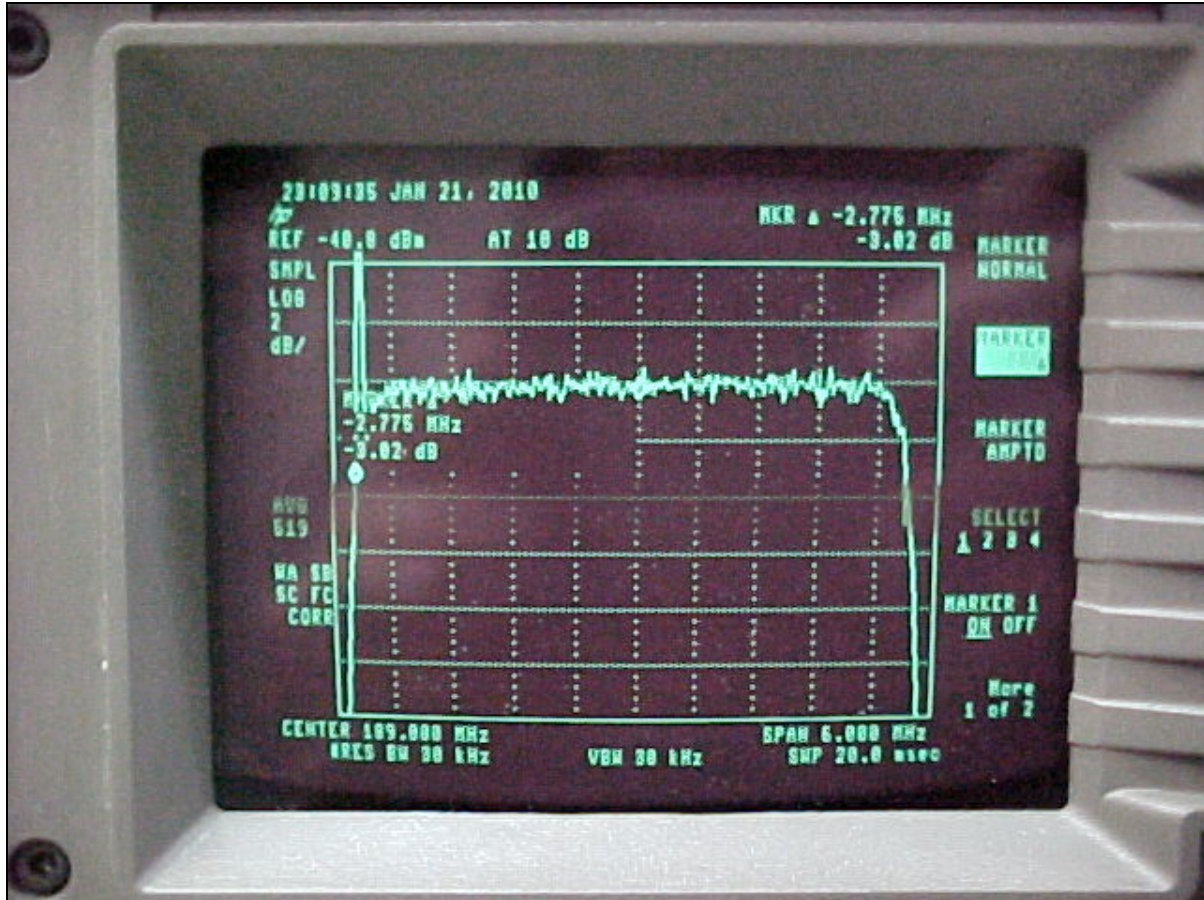


Figure 4-7A. Channel Occupied Bandwidth (lower channel edge)

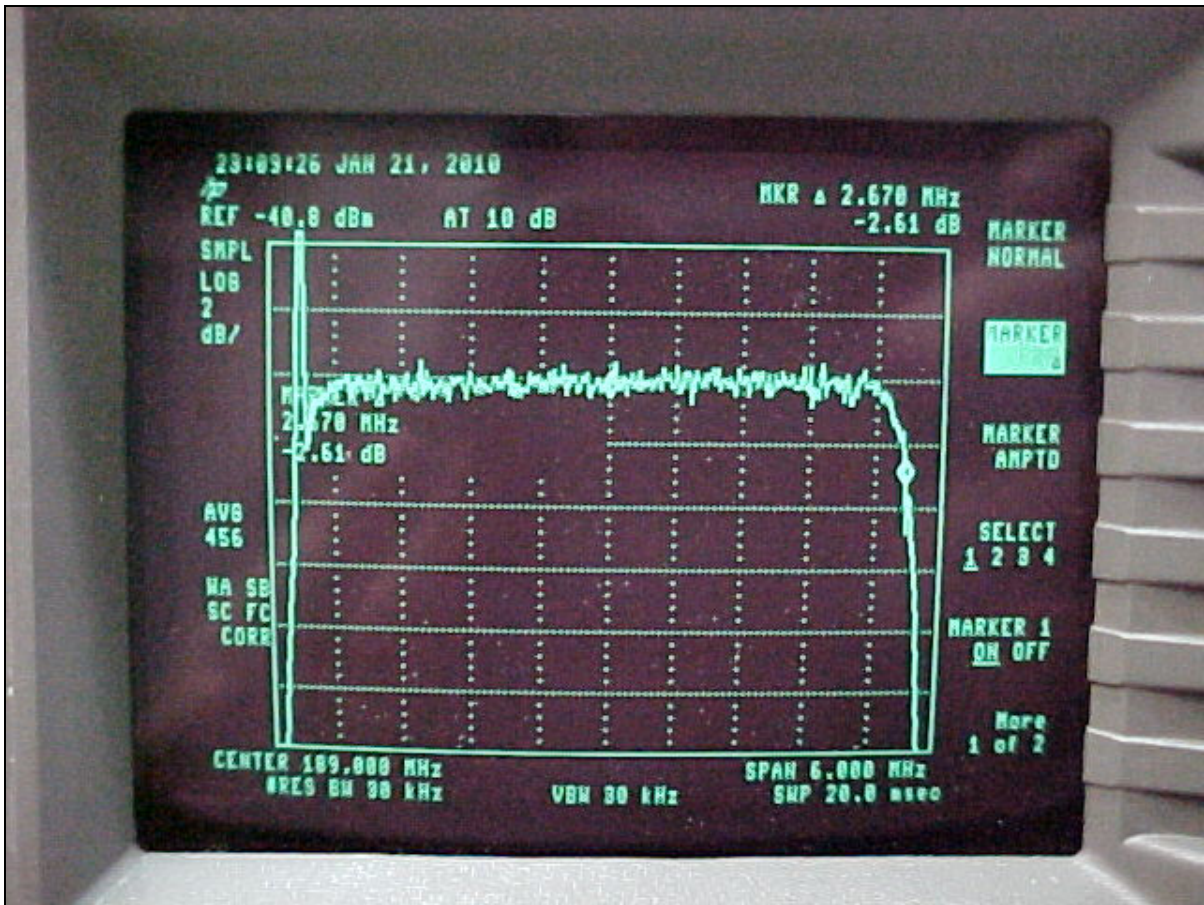


Figure 4-7B. Channel Occupied Bandwidth (Upper channel edge)



#### 4.6 Conducted Spurious Emissions

The Hewlett Packard Spectrum Analyzer was used to measure harmonics before the mask filter. The levels were then added to the filter attenuation at any harmonics that were detected in order to obtain the final harmonic levels after the mask filter.

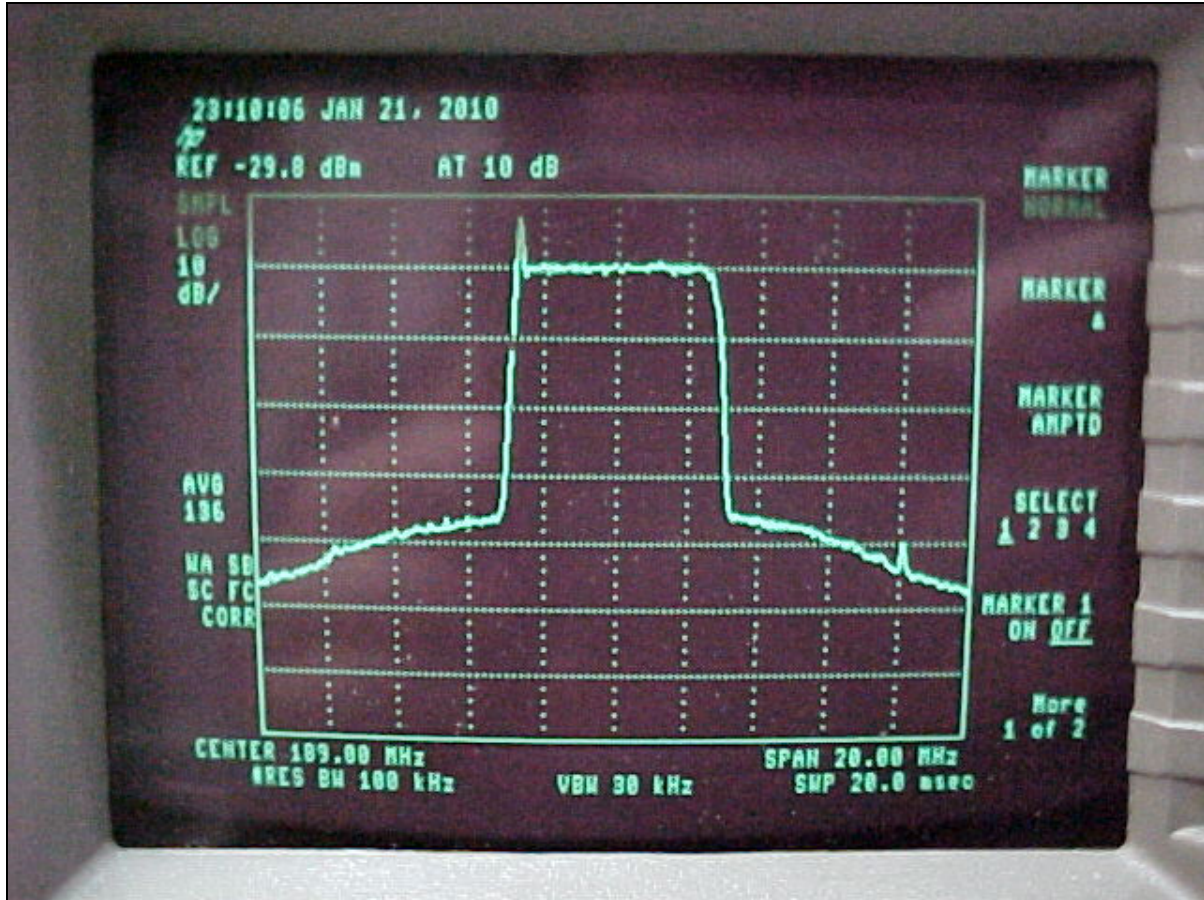


Figure 4-8A. Reference (Pre-Channel Filter)



Figure 4-8C Second Harmonic (Pre-Filter)



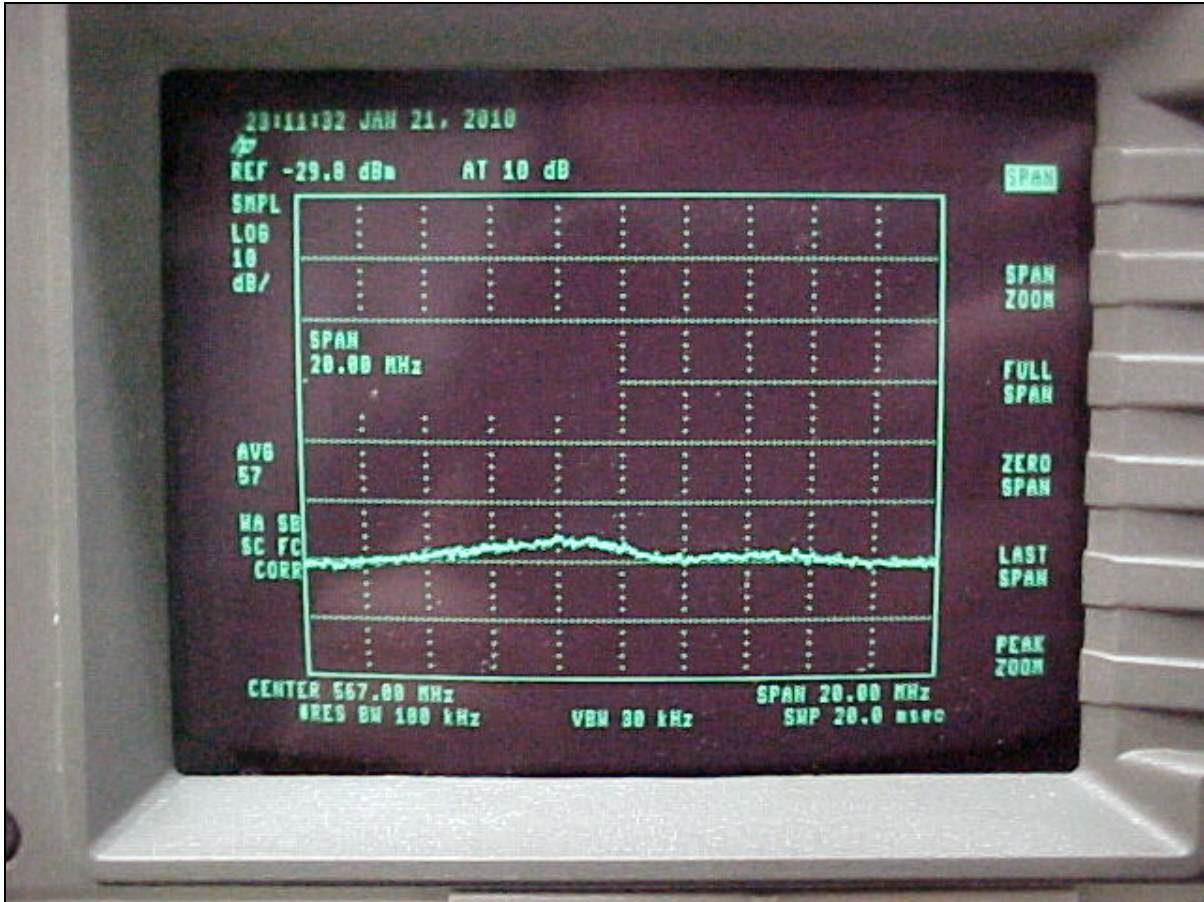


Figure 4-8D Third Harmonic (Pre-Filter)

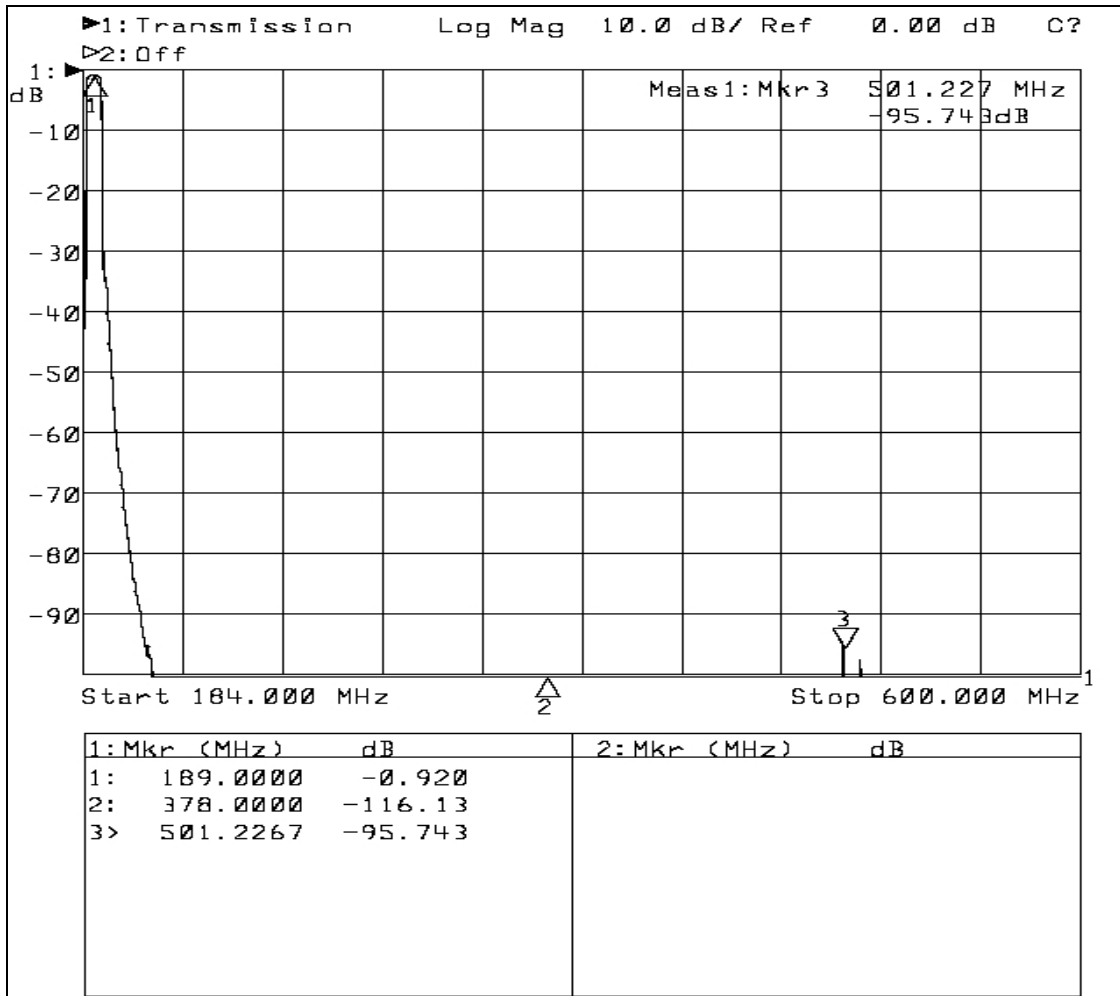


Figure 4-8D.Mask and Harmonics Filter Response

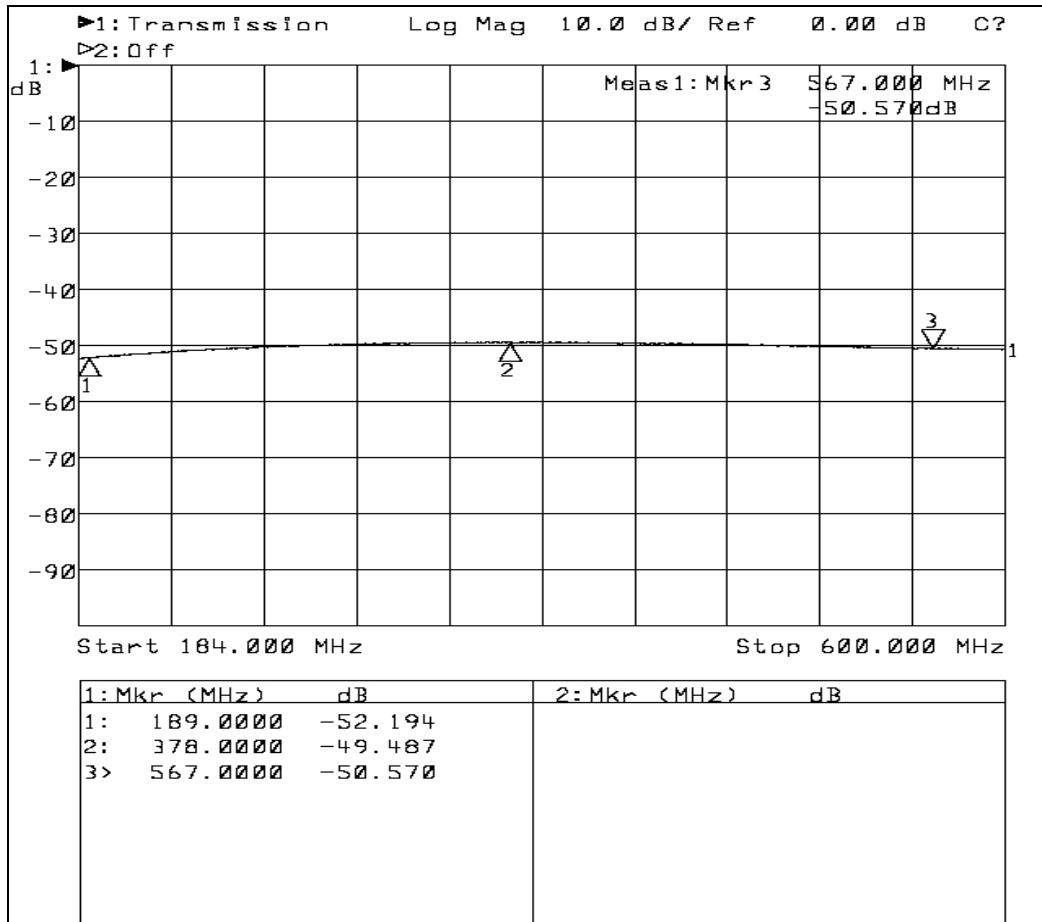


Figure 4-E. Directional Coupler Response



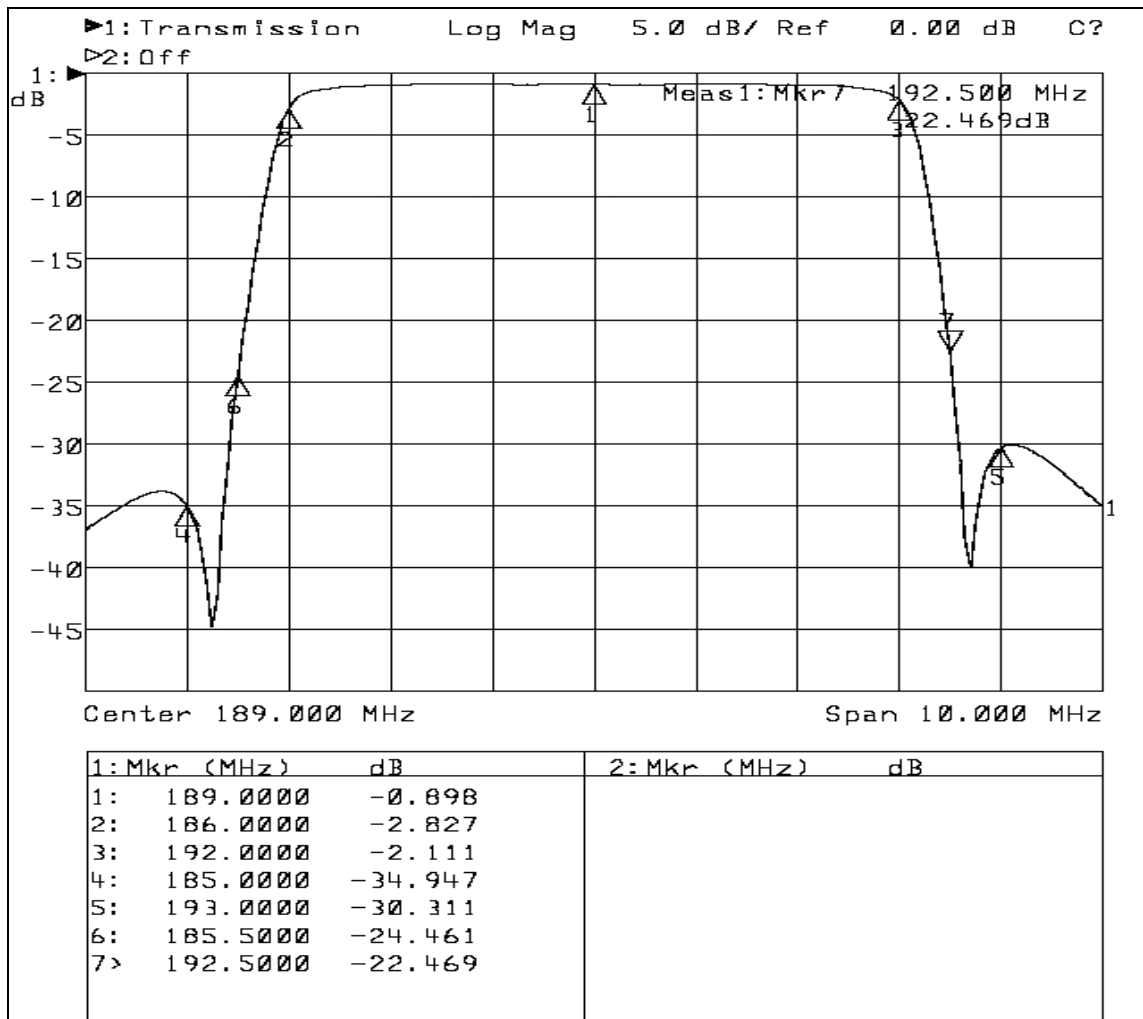


Figure 4-8F. Filter response

Harmonic (MHz)	Level Relative to fundamental	Filter Response (dB)	Coupler Response (dB)	Bandwidth correction (dB)	Total (dB)
Fundamental (189)	ref	--	--	---	ref
Second (378)	-34	-95	-3	10	-142
Third (567)	-46	-95	-2	10	-153

Table 4-1. Post Filter Harmonic Levels

The FCC rules require the energy to be measured in 500 kHz, and compared to the total channel power. The channel power is spread out evenly over an occupied bandwidth of 5.4 MHz. The total channel power is  $10 \cdot \log(5.4/0.5) = 10.3$  dB.

Frequency	Relative Level (dB)	FCC Requirement (dB)
Fundamental	0 dB (reference)	---
Second Harmonic	-142	-76
Third Harmonic	-153	--76

Part 74.794 of the Rules states:

**(ii) Stringent mask.** In the first 500kHz from the channel edges, emissions must be attenuated no less than 47 dB. More than 3 MHz from the channel edges, emissions must be attenuated no less than 76 dB. At any frequency between 0.5 and 3 MHz from the channel edges, emissions must be attenuated no less than the value determined by the following formula:  $A(\text{dB}) = 47 + 11.5 (Df-0.5)$

Frequency relative to Center Freq (MHz)	Transmitter IMD (dB)	Filter Response (dB)	Bandwidth correction (dB)	Total (dB)	FCC Requirement (dB)
-3	-37	-2	-10	-49	-47
-3.5	-37	-24	-10	-71	-47
-4.0	-37	<-30	-10	-77	-64.5
+3.0	-37	-2	-10	-49	-47
+3.5	-37	-22	-10	-69	-47
+4.0	-37	<-35	-10	-82	-64.5

#### 4.7 Radiated Emissions

Using the test setup shown in Figure 4-1, with the transmitter operating at full power, the spectrum analyzer was moved 3 meters from the transmitter and connected to a dipole antenna cut to the 189 MHz. This antenna was oriented to maximize the received level and the data was recorded. The antenna was then cut to the local oscillator frequency and the second and third through harmonic frequencies of the transmitter, and all of the signals received, were maximized by antenna orientation and their absolute levels were recorded.

With these various antennas, the only measurable level observed was at 189 MHz. This level is shown below in Table 4-2 and an analysis of the relative field and strength is provided in the following paragraphs.

FREQUENCY	MEASURED LEVEL (INTO 50 Ω)
189 MHz	-45 dBm

Table 4-3. Measurable Levels Observed in Frequency Spectrum

The spectrum analyzer had a maximum sensitivity of -110 dBm during these tests.



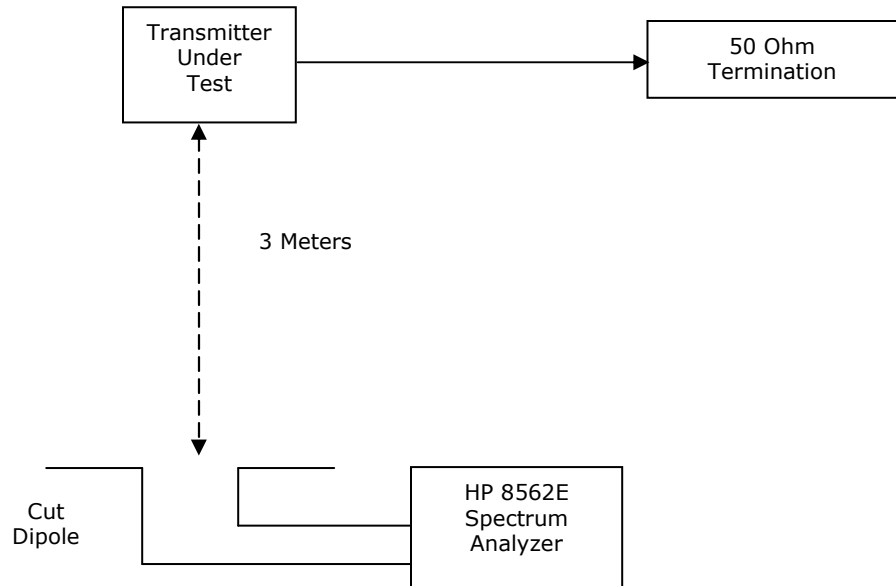


Figure 4-9. Test Setup for Measuring Radiated Emissions

One level was compared to the following reference level.

If all of the power of the transmitter was radiated by an isotropic radiator, the power density at 3 meters would be:

$$P = P_t / 4\pi R^2 = 400 / 4\pi \cdot (3)^2 = 3.5 \text{ w/m}^2$$

Using a dipole-transmitting antenna increases this by 1.64 to:

$$1.64 \times 3.5 = 26 \text{ w/m}^2$$

If a dipole-receive antenna of area  $1.64 \times \lambda^2 / 4\pi$  is used to receive the signal, the received level would be: +27 dBm

The receive level at -45 dBm was therefore at -72 dB relative to this level.

The cabinet radiation was also checked with the receive dipole antenna cut to 189.00 MHz, within very close proximity to the trays of the transmitters, and the received level that was recorded at no time exceeded a power density in excess of -10 dBm:

This level is far less than the current or proposed standard for safe radiation levels.

## 4.8 Frequency Stability

The transmitter is designed to operate using either an internal or external 10 MHz precise reference oscillator. The frequency stability of this reference determines the frequency stability of the transmitter.

The frequency determining variables of the transmitter may be defined as follows:

$F_{LO1}$  = Desired local oscillator 1 frequency  
 $F_{LO2}$  = Desired local oscillator 2 frequency  
 $F_{IF}$  = Desired IF oscillator frequency  
 $F_R$  = Desired external reference oscillator frequency  
 $F_{RF}$  = Desired RF output frequency  
 $E_{LO1}$  = Local oscillator 1 frequency offset error  
 $E_{LO2}$  = Local oscillator 2 frequency offset error  
 $E_{IF}$  = IF oscillator frequency offset error  
 $E_R$  = External reference oscillator frequency offset error  
 $E_{RF}$  = RF output frequency error

The PLL circuitry maintains a constant ratio between the external reference frequency and the output frequency of the oscillator. This ratio is defined below for both the LO and IF oscillators.

$G_{LO1} = F_{LO1} / F_R$   
 $G_{LO2} = F_{LO2} / F_R$   
 $G_{IF} = F_{IF} / F_R$

Any change in the external 10 MHz reference will effect a corresponding change in the output frequency such that the above ratios are maintained.

$G_{LO1} = (F_{LO1} + E_{LO1}) / (F_R + E_R) = F_{LO1} / F_R$   
 $G_{LO2} = (F_{LO2} + E_{LO2}) / (F_R + E_R) = F_{LO2} / F_R$   
 $G_{IF} = (F_{IF} + E_{IF}) / (F_R + E_R) = F_{IF} / F_R$

Solving for the change in output frequency yields:

$E_{LO1} = E_R * (F_{LO1} / F_R) = E_R * G_{LO1}$   
 $E_{LO2} = E_R * (F_{LO2} / F_R) = E_R * G_{LO2}$   
 $E_{IF} = E_R * (F_{IF} / F_R) = E_R * G_{IF}$

The desired RF carrier frequency is equal to the LO2 frequency minus the LO1 Frequency and the IF frequency:

$F_{RF} = F_{LO2} - F_{IF} - F_{LO1}$

The actual RF frequency, including any error introduced by the external reference, may be defined as follows:

$F_{RF} + E_{RF} = (F_{LO2} + E_{LO2}) - (F_{IF} + E_{IF}) - (F_{LO1} + E_{LO1})$   
 $F_{RF} + E_{RF} = (F_{LO2} - F_{LO1} - F_{IF}) - (E_{LO2} - E_{LO1} - E_{IF})$   
 $F_{RF} + E_{RF} = F_{RF} + (E_{LO2} - E_{LO1} - E_{IF})$

Calculating for the error of the carrier yields:

$$E_{RF} = (E_{LO2} - E_{LO1} - E_{IF})$$

$$E_{RF} = E_R (G_{LO2} - G_{LO1} - G_{IF})$$

$$E_{RF} = E_R / F_R * (F_{LO2} - F_{LO1} - F_{IF})$$

$$E_{RF} = E_R / F_R * F_{RF}$$

Therefore, the error of the RF carrier is a function of the external 10 MHz reference error.

The frequency stability requirements are described in 27.54:

*The frequency shall be sufficient to ensure that the fundamental emissions stay within the authorized bands of operation.*

The 10 MHz oscillator was tested over temperature and voltage below:

TEMP. (°C)	10 MHz Error
-30	3.0 Hz
-20	2.1 Hz
-10	1.2 Hz
0	0.7 Hz
+10	0.5 Hz
+20	0.1 Hz
+30	0.2 Hz
+40	0.3 Hz
+50	0.8 Hz

Table 4-4.

Line Voltage	Reference Oscillator Error
95V	0.0 Hz
115V	0.0 Hz
135V	0.0 Hz

Table 4-5.

The frequency error at final frequency is  $3.0 * \text{channel frequency} / 10 \text{ MHz}$ . The highest frequency for this application is 213 MHz, giving a frequency error of  $3 * 213 / 10 = 64 \text{ Hz}$ .



#### 4.9 Test Equipment

The test equipment that was used to analyze the Axcera CHV400BTD system is listed in Table 4-6.

<b>MODEL</b>	<b>MANUFACTURER</b>	<b>DESCRIPTION</b>	<b>SERIAL #</b>
435B	Hewlett Packard	Power Meter	2732UO9080
8482B	Hewlett Packard	Power Sensor	N/A
8892-300	Bird	50 $\Omega$ Termination	2867
EFA 2067.3004.53	Rohde & Schwarz	8VSB Demodulator	100137
3020A	Narda	Directional Coupler	40660
89441A	Hewlett-Packard	Vector Signal Analyzer	3416A01547
8562E	Hewlett-Packard	Spectrum Analyzer	03760
8714B	Hewlett-Packard	Network Analyzer	U535490409
RFA300	Tektronix	Test Measurement Set	B010226

*Table 4-6. Test Equipment*