

picoNode Module Datasheet and Integration Specification

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Revision History

Revision	Release Date	Change Description
0.1	May 10, 2017	Initial release
0.2	Jan. 15, 2018	Update Tables 1 and 4
0.3	Jan. 31, 2018	Add FCC Warnings in Sec. 10

1 Overview

This document provides a brief overview of guidelines allowing an integrator to design a Host product that utilizes the picoNode and ensures that the system meets all of its technical objectives and requirements.

1.1 picoNode

The picoNode is a small form factor wireless network module that easily integrates with a microcontroller or applications processor using a Serial Peripheral Interface (SPI). The top side of the printed circuit board (PCB) is enclosed with a radio frequency (RF) shield. The picoNode is an LGA-style module designed to be soldered directly onto a host board via SMT processes. For details, see Appendix B: PCB Land Pattern and Keep-outs.

Table 1 picoNode Specifications

picoNode	
Dimensions (per unit)	26 mm x 16 mm Thickness: 2.4mm
Weight (per unit)	1.5 Grams
Transmit Power	21.5 dBm (typical)
RX Sensitivity	-133 dBm (nominal)

For more mechanical details about the picoNode, refer to the mechanical drawing in oD Appendix D picoNode Mechanical Drawing.

The following figure shows how a picoNode interfaces with a Host application, running on an applications processor.



Figure 1 Typical Application Diagram

2 DC and AC Characteristics

2.1 Absolute Maximum Ratings

Operating outside of these ranges may damage the unit.

The picoNode is MSL 3-rated and should be handled as an MSL 3 device per IPC/JEDEC J-STD-033 (latest revision).

Table 2 Absolute Maximum Ratings

Parameter	Min	Max	Unit
Storage Temperature	-40	85	°C
Operating Temperature	-40	85	°C
Vbatt Input Voltage	2.5	6.0	V
3.3V Supply	3.1	3.5	V
Digital Interface Signals, 3.3V nominal	3.0	3.6	V

2.2 Recommended Operating Conditions

Table 3 Operating Conditions

Parameter	Min	Max	Unit
Input voltage, VBATT	2.5	5.5	V
3.3V Input	3.2	3.4	V
Ambient Temperature, Ta	-40	85	°C

The following characteristics apply across the -40 $^{\circ}$ C to +85 $^{\circ}$ C temperature range unless otherwise noted.

Table 4 Operating Characteristics

Description	Min	Тур	Max	Units			
DC Characteristics	DC Characteristics						
Voltage – Vbatt	2.5	3.3	5.5	Volt			
Off Current – Note 1		0.1		μA			
Deep Sleep Current - Note 1		19		μΑ			
Idle Current – Note 1		20		mA			
Receive Current – Note 1		100		mA			
picoNode: Transmit Current — Note 2		330		mA			

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Description	Min	Тур	Max	Units	
Digital					
VOL – Voltage Output, Low (4mA sink)	0		0.4	V	
VOH – Voltage Output High (4mA source)	2.4		3.3	V	
SPI Clock – Note 11	0.1		8.6	MHz	
Environmental					
Operating Temperature	-40		+85	°C	
Storage Temp	-40		+85	°C	
Humidity – non-condensing	5		95	%	
Ramp Temperature (maximum rate at which operating temperature should change)			30	°C/Hr.	
MTBF (picoNode)				MHrs	
Receiver					
Receiver Sensitivity – Note 3		-133		dBm	
Receiver Image Reject	25	33		dB	
Noise Figure		4.9		dB	
Input IP3 (high LNA gain mode)				dBm	
Maximum RF input level for specification compliance				dBm	
General RF Characteristics					
Frequency Range – Note 4	2402		~2482	MHz	
Channel Spacing	N/A	1.99	N/A	MHz	
Transmitter					
Maximum RF Conducted Power –Note 5					
FCC/IC markets:		21.5	23.65	dBm	
ETSI markets:	8.5	9.5	10	dBm	
Carrier Rejection	-35	-51		dBc	
Signal Modulation		DSSS- DBPSK			
Signal Bandwidth		1.0		MHz	
BT Factor		0.3			
Peak to Average Ratio		2.3		dB	
Spectral bandwidth at maximum RF power:					
-6dB BW		0.96		MHz	
		1.75		MHZ	
ACPR – Note 6			-30	dBc	
Harmonics – Note 7			-43	dBm	
Transmit Power Level Accuracy – Note 8			±1.5	dB	

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Description	Min	Тур	Max	Units
Transmitter Spurious Outputs – Note 9 30MHz to 2400MHz: 2482MHz to 8000MHz:			< -43 < -43	dBm dBm
VSWR Tolerance Maximum VSWR for spec compliance – Note 10: Maximum VSWR for stability.			1.5:1 9:1	

NOTES:

- 1. Tested at 3.3V input, +25C. Please note the following:
 - a. There are power differences between the Voltage/Current numbers in this table.
 - b. The Table 4 refers to a maximal current draw that the Host system should be designed to accommodate.
- 2. Measured at:
 - □ picoNode: +21.5 dBm TX output (Typ=50 Ω), 3.3 V, range includes VSWR ≤ 1.5:1 (Po not compensated).
- 3. Sensitivity at maximum DL spreading factor of 11 (2048) with 10% FER.
- 4. The upper frequency range is market dependent:
 - c. FCC/ISED: CH₃8; 2475.63 MHz.
 - d. ETSI: CH40; 2475.63 Hz.
 - e. Japan: CH41; 2481.60 MHz.
- 5. Maximum TX RF power:

picoNode: This is limited by FCC/IC grant to 21.5 dBm in these markets. Transmit power is configured during network join time to meet country-specific deployment and regulatory requirements. The configurable range is 0 - 21.5 dBm in 1 dB integer increments. For non-integer Power such as 21.5dBm, the Node's MAX_TX Power must be set to 22dBm to force a maximum calibrated value.

- 6. Spec and test method comes from FCC 15.247(d); Band Edge Emissions, 2 MHz offset.
- 7. At any TX power level, VSWR ≤ 3:1. Harmonics fall into FCC restricted bands.
- 8. Estimated sum of all contributors with VSWR ≤ 1.5:1. Normal link mode.
- 9. At any TX power level, VSWR ≤ 3:1. Applies to spurious, not ACPR or harmonics. Generally the largest spurious output outside the 2.40-2.48GHz band is at 2/3LO and 4/3LO.
- 10. Maximum VSWR for spec compliance applies at 25°C only. Slightly degraded ACPR/mask and power variation can be expected at temperature extremes.
- 11. The SPI clock has a maximum rate of 26 MHz/3 and a minimum of 100 kHz. There is no physical limitation on the minimum clock rate but the 100 kHz is deemed "marginal" and is not absolute. Depending on the data traffic model and level of debug traffic, 100 kHz may cause a backup of SPI traffic, which then causes buffer overflow conditions. The application must be validated to ensure that the SPI clock is sufficient to support required traffic.

<u>3 Electrical Interface</u>

This chapter describes the electrical interface of the picoNode and how the Host processor controls for the picoNode.

Table 5 picoNode Pin Descriptions

Pin #	Pin Name	Signal Direction Relative to picoNode	Signal Type	Comment
2	WAKE	Output, 1.8V	CMOS_O*	This is a 1.8V output signal that reflects the status of the Node's power state. When WAKE is "high" the Node is active in Idle, RX, or TX states. When WAKE is "low", the Node's 3.3V is internally gated OFF and the Node is in its lowest power state. This signal is to be used for "battery operating modes", as describe in Figure 6 Lithium Battery Example. Connect the WAKE pin directly to the enable pin of a voltage regulator or FET switch controlling 3.3V
1,3,4,5,7,10,12, 14,18,24,26,28, 29,30,31,32,34	Ground	Power	Power	Ground return. Should be low RF impedance to a solid ground plane of the Host
23	3V3	Power	Power	The 3.3V can be continuously supplied(line powered) or only when the WAKE pin is asserted "high"(battery powered). This power domain is high power (internal CPU, Transceiver, and RF PA) and should be decoupled with a low ESR, high capacitance Capacitor.
15	VBATT	Power	Power	Input power to the picoNode. This power domain is low current but is used 100% of the time to supply internal Supervisory domains.

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Pin #	Pin Name	Signal Direction Relative to picoNode	Signal Type	Comment
8	SRQ	Output	СМОS_О	SPI Slave Request. SRQ must be connected to a pin that can wake the application processor from sleep, for battery powered applications.
9	SRDY	Output	CMOS_O	SPI Slave Ready
21	SCLK	Input	CMOS_I	SPI Clock
19	MISO	Output	CMOS_O	SPI Master Input Slave Output
22	CS	Input	CMOS_I	SPI Chip Select(Note other slaves are prohibited on the SPI interface, but this pin must be controlled by the Host Common Library). It CANNOT be tied low on the PCB.
20	MOSI	Input	CMOS_I	SPI Master Output Slave Input
11	MRQ	Input	CMOS_I	SPI Master Request
13	ON_OFF	Input	CMOS_A	 This is used to turn ON/OFF the Internal Power supplies of the picoNode. It is controlled by the Host Common Library. Low: Node consumes <1uA High: Node is active and will run through a wide range of power
17	TOUT	Output	CMOS_O	states. TOUT is a normally low signal that pulses high in response to specific Network Timing Events. It allows an application to trigger a measurement with sub-1ms accuracy.
6	RF_TXENA	Output	CMOS_O	This signal is used to indicate status of the Power Amplifier for the picoNode: Low = OFF High = Enabled (Transmitting) The rise edge can be used to trigger a Host CPU's ADC read of VBATT (battery voltage while under maximum load).

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Pin #	Pin Name	Signal Direction Relative to picoNode	Signal Type	Comment
25, 27	RF1, RF2	RF RX/TX	50 Ohm	These are the RF ports for the picoNode. They are 50 Ohm port, DC coupled. RF1 is required but both are desired for antenna diversity. Single port or dual antenna port can be configured in the provisioning process.
33	RF_SHDN	Output	CMOS_O	 This pin indicates the status of the RF Transceiver for the picoNode: Low = Shutdown High = Active It can be used for WIFI/BT coexistence, and to reduce power supply current during low power states (see 3.1.3)
16	TIME_QUAL	Output	CMOS_O	This pin is reserved and should be left as Do Not Connect.

NOTES:

- 1. The VDD of the internal logic of the picoNode is 3.3Volt.
- 2. The Host is the SPI Master and the picoNode is the SPI Slave.
- CMOS_I: The Node input voltages are 3.3V CMOS levels. VIH = 2.0V (minimum) and VIL = 0.8V (maximum).
- 4. CMOS_O: The Node output voltages are 3.3V CMOS levels (4mA). VOH = 2.4V (minimum) and VOL = 0.4V (maximum).
- 5. SPI inputs to the node (SCLK, MOSI, CS) must be tri-stated or driven low when the node may be sleeping (MRQ and SRQ are both low). See section 5: SPI Interface and Sequences for more details.

3.1 Signal Descriptions

3.1.1 GND

Ground is the Host CPU's ground to enable a common reference between CPU and Node.

3.1.2 VBATT

This supplies a low current 2.5 V – 5.5 V for the Node's internal supervisory circuitry. This pin should be decoupled with a 0.1 μ F capacitor on the Host processor board.

3.1.3 3V3

This pin drives the CPU, Transceiver, RF PA section of the module. It can consume up to 800 mW. Allow for bypassing with a 47 μ F low ESR cap (bulk) and a 0.1 μ F ceramic cap for optimal performance. Depending on the Host design, there are some nuances that are important regarding this signal:

- 3.3 V can be supplied continuously or only when the WAKE signal is asserted "high.", for battery powered applications.
- The Node runs through various operating states when 3.3V is supplied.
 - □ If the Node internally is in a state that requires no RF, the 3.3 V can be "noisy" (+/-100 mV ripple). The RF state is defined by the RF_SHDN pin. This allows the Host's 3.3 V regulator to work in low quiescent (power save) modes.
 - □ If the Node internally is active and does require RF, the 3.3 V must be "clean" (+/-20mV ripple). This forces the Host's 3.3V supply into a high precision mode and forcing a high quiescent current of that regulator.
 - □ If the Node is operated in a battery mode, when 3V3 isn't always enabled, the 3V3 supply must power up and be stable within 2mS of the WAKE signal going "high"

This switching of "noisy" and "clean" becomes clear (and important) when working with battery operated devices and optimal low power drain.

3.1.4 ON_OFF

This input signal controls the power-on of the LDO circuitry for the picoNode. This signal is controlled by the Host Common Library, compiled onto the user's apps processor. For reference only: It must be shut off prior to starting the picoNode power-up sequence as defined in section 5.4: Startup (Power On) Sequence. After the picoNode powers up, this signal is to remain logic high during normal operational modes. This pin dually serves a power on/off function as well as a Node Reset function.

3.1.5 MRQ

The MRQ (Master Request) is the Host's normal way of waking the picoNode to initiate SPI communications. Logic "High" forces the picoNode awake. This signal is controlled by the Host Common Library, compiled onto the user's application processor.

3.1.6 SRDY

SRDY (Slave Ready) is an indication from the picoNode that it has fully booted its internal Firmware image, initialized its Hardware and Interfaces, and is ready for communication (arbitration) with the Host. Logic "High" indicates the picoNode is ready for communications. This signal is controlled/handled by the Host Common Library, compiled onto the user's application processor.

3.1.7 SRQ

The SRQ (Slave Request) signal is an indication from the picoNode that it wants the Host's attention. When SRQ is asserted "High," the Host must read the Status registers of picoNode. If SRQ is "High," SRDY will also be "High." This signal is controlled/handled by the Host Common Library, compiled onto the user's application processor.

For battery powered applications, SRQ must be connected to a pin that can wake the application processor from sleep.

3.1.8 SPI System

The SPI system is the generic term used for all SPI signals (MOSI, MISO, CS, SCLK) to be set up for SPI communications to occur between Host and picoNode. The picoNode SPI is the Slave in the Master/Slave communications and is defined in section 5.2: SPI Mode and Timing.

IMPORTANT NOTES:

- 1. Other SPI slaves are not allowed to share the SPI signals.
- 2. CS must be controlled by the Host Common Library API to guarantee correct sequencing. Specifically, the user must ensure that the SPI CS is active (low) for the whole duration of a message transfer, with no gaps.

3.1.9 TOUT

This signal is a Time Synchronizing signal that pulses high upon specific network timing events.

3.1.10 RF_TXENA

This signal indicates when the device is transmitting. When transmitting, it is recommended that the Host processor use this opportunity as a trigger to read the system "Vbatt" power line to show battery voltage under maximum load.

3.1.11 RF_SHDN

This Node signal indicates status of the RF Transceiver of the picoNode. If low, the transceiver sleeps (no RX and no TX). This output of the module (3.3 V) indicates when the RF transceiver is on or off. When RF_SHDN is high, the RF is "ON" (RX or TX). In the RF "ON" mode, the module needs a "clean" 3.3 V (low ripple).

3.1.12 RF1 and RF2

These are the RF ports (RX and TX) of the Node. They are DC-coupled, 50 Ohm and require special Host routing of PCB. RF1 is the primary antenna and is always required. RF2 is a secondary antenna that the Node can use for Antenna Diversity. A single or dual antenna (diversity) system can be configured during the Provisioning process. For best results ensure the load termination (antenna) has a VSWR of 1.5:1 or better (return loss < -10 dB).

3.1.13 WAKE

The WAKE signal is generated by the module and is 1.8V. It signals that it now requires a 3.3V source. Generally, for a Powered Node (Figure 5 Powered Example) the WAKE is not required since 3.3V already exists. In the battery example (Figure 6 Lithium Battery Example) the WAKE turns on the Host's main supply to regulate the battery to the required 3.3V.

If WAKE is used to enable the Host's 3V3 supply, the 3V3 supply must be on and stable within 2mS of WAKE going high.

3.1.14 TIME_QUAL

This signal is reserved for Compal.

4 Applications

This chapter describes two major methods of implementing the picoNode: Powered and battery operated modes.

1. Powered Operating Mode

"Powered" refers to the Host board and its power supply. Specifically, it refers to a good solid, consistent power supply – such as an AC powered source. In this type of usage case, the Host can potentially supply a good clean 3.3 V rail (pin 4) without regard to extreme detail of optimizing power consumption.

2. Battery Operated Mode

This mode refers to a battery operated device in which extreme care and attention is applied to reducing overall power consumption. All details and all modes of power consumption are considered and optimized.

4.1 Powered Operating Mode

In this scenario, the Host and Node's power consumption is not really a consideration. In this case, the Host can take an input Voltage and convert to a 100% duty 3.3 V supply. This would be applied to:

- Host CPU
- VBATT of picoNode
- 3V3 of picoNode

In this case, the node will internally turn the 3.3 V rail ON/OFF as it is required. The Node attempts to gate this internal supply OFF to conserve power, when it can. The HW status of the 3.3V supply is the WAKE signal (3.3V = "ON" if WAKE is High). When the 3.3V is internally gated off, the Node consumes virtually no power on that supply.

However, the Host CPU is always powered by 3.3 V – the operating level of the Host-picoNode SPI interface. The picoNode requires a full 3.3 V during SPI transfers. For a wired example of the Powered circuitry, refer to Figure 5 Powered Example.

4.2 Battery Operated Mode

Battery operating mode offers a much more challenging design constraint. The goal is to reduce current consumption where possible. To this conservation goal, see Figure 6 Lithium Battery Example.

The assumption in this design goal is to use a long life battery such as Lithium battery/cell. These primary cells offer:

- Extreme low self-discharge
- Long life

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- Wide operating temperatures
 - □ High capacity
 - □ 2.5 V 3.6 V direct voltage over the operating temperature and discharge characteristics of the cell

In this case, assume that:

- The VBATT pin of the Node is connected directly to the cell (2.5 V 3.6 V).
- The 3V3 signal of the Node is connected directly to the 3.3 V regulator. The regulator is turned ON/OFF depending on the state of the WAKE pin of the Node.
- The CPU uses an indirect path through a dual Schottky diode to permit seamless switching/transition from battery voltage to 3.3 V.
- Most modern CPUs used for this type of application have the following characteristics:
 - **D** Operate at 1.8 V 3.6 V
 - **\Box** Low sleep current (1 2 μ A with full SRAM retention for fast wakeup)
 - □ Pins that use interrupts to wake the processor from deep sleep

The following sections describe the major operating states and power supply modes.

4.2.1 Sleep Mode Assumptions

- 1. The Node is powered "on" (ON_OFF = high).
- 2. The Node is in sleep mode when in the WAKE state (WAKE = low).
- 3. The Node is consuming minimal current through its VBATT pin.
- 4. 3V3 is off and is not required by the Host or Node.
- 5. The Host CPU can run off the "natural voltage" of the cell only drawing microAmps of current.

4.2.2 Active Mode

When the Node awakens (WAKE = high) and the 3.3 V power supply is turned "on":

- The Node uses the supplied 3.3 V as required. The 3V3 must be on and stable within 2mS of WAKE being asserted high.
- The voltage of the CPU ramps to the higher of the following: cell voltage or 3.3 V supply.
- The 3.3 V on the Host and the Node permits the Node to fully function on the RPMA network and also allows Host-Node SPI communications.

A Texas Instruments TPS63000 3.3 V buck/boost regulator (or equivalent) is recommended for picoNode applications. This has some unique benefits:

- When in high efficiency mode (PS/SYNC = low), the regulator outputs a loose 3.3 V with +/-100 mV ripple. In this mode, its quiescent current is only 50 µA.
- When in low efficiency (but "clean" output) mode (PS/SYNC = high), the regulator is within +/-20 mV ripple but consumes a 0.6 mA quiescent current.

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These operating modes align well with the Node's operating modes as defined by the RF_SHDN pin:

- When tight regulation (low ripple) is required by the RF of the Node:
 - □ RF_SHDN = high
 - □ PS/SYNC = high
- When the Node does not need tight regulation and is in a moderately low power mode, RF_SHDN is low (and PS/SYNC is low).

Other design considerations for optimal low power are as follows:

- 1. Reduce capacitance in the 3.3V domain. Extra capacitors require charging each time the 3.3 V is turned "on".
- 2. Ensure all software and CPU configurations set CPU pins to low power states.
- 3. Use high value resistors such as 1 M or greater for pull-ups/pull-downs where possible.

5 SPI Interface and Sequences

5.1 SPI System Interface Overview

The SPI slave interface is currently the only supported interface for Host-to-Node communication. The SPI System interface is controlled/handled by the Host Common Library, compiled onto the user's apps processor. The following section (all of section 5) is informational only.

NOTE: The picoNode must be the only SPI slave on the bus.

The SPI slave interface provides communication with an external Host through a 7-wire interface. The Host is the SPI master and the picoNode is the SPI slave. In addition to the four standard SPI signals, three additional signals are used to complement the SPI bus: MRQ, SRQ, and SRDY. The additional signals are included to support picoNode state transitions and bi-directional message traffic.

The SPI signals include four that are controlled by the master and three that are controlled by the slave.

Slave-controlled Signals (picoNode)

Master-controlled Signals (Host)

MOSI

MISO

SRQ

SRDY

- SCLK
- CS
- MRQ

When MRQ and SRQ are low, the remaining Master controlled signals (MOSI, SCLK, and CS) must be held low or tri-stated. This is to prevent these signals from back-driving the picoNode (Slave) that may be in deep sleep. When either MRQ or SRQ assert high, the Master should set each of the three signals appropriately according to their standard usage. No pull-up resistors should ever be applied to any signals on the picoNode since it often needs to fall into a Deep Sleep mode (all internal regulators turned off).





5.2 SPI Mode and Timing

Figure 2 SPI Timing, CPOL = 0, CPHA = 0

5.3 Host Initialization

What is described here is the initialization of the Host, its operating software, and the control sequences used to drive the picoNode.

Due to specific clock and memory requirements, the picoNode must go through specific Initialization and Wake sequences.

NOTE: Some CPUs have internal pull-up resistors that are active after Power On Reset. Through CMOS leakage, the Host CPU can supply voltages to the picoNode I/O bus prior to the Host CPU fully initializing and disabling the pull-up resistors. It must be noted that during the brief initialization period, the ON_OFF signal must be "low." Activating the ON_OFF signal with other picoNode signals being pulled "high" can cause CMOS latchup within the picoNode.

5.4 Startup (Power On) Sequence

During, and immediately after Power On Reset (POR), the Host has no control of its I/O power states. For instance, some CPUs have GPIO that tri-state or act as inputs during power up. Other CPU brands have programmable pull-ups on its I/O and need the Host CPU to disable those pull-ups for the Host's GPIO to work correctly with the picoNode. This setup and configuration of GPIO takes a finite time during the Host boot process. This is detailed in the following figure.

Whereas the power-up sequence is described here, it is recommended the Integrator not attempt this entire startup sequence without assistance. Compal offers a formal and controlled library to help with this startup and communication interface.

The initial sequencing of the 3.3V and Vbatt rails are:

• Both 3.3V and Vbatt van be applied simultaneously, or

• The Vbatt can rise first if 3.3V follows within 1mS, but





Figure 3 picoNode Power-up Timing Sequence

The timing sequence shown in Figure above is described below. **NOTE:** The timing shown in the figure is not to scale.

- t0 → t1 This phase is where the Host's System power has been applied and the Host Software must power up and initialize the GPIO interfaces for the picoNode to the required states defined at time t1. The t1 state becomes "TRUE" indicating the Host sets all the GPIO to a known and controlled state (Low).
- t2 At t2 state, all the output signals to the picoNode are set low and the Power On Signal is set high. This turns on the internal LDO regulators of the picoNode to initiate a power up sequence. The time between t1-t2 is approximately 100 ms, or longer. Assert ON_OFF "high" starts an internal clock (32 K) and releases an internal reset pin which may take up to 3 seconds to stabilize.
- t3 t3 is when the Host releases the picoNode from its Reset state. This time allows the 32 kHz of the picoNode to turn on and stabilize. The time between t2-t3 is 1 second, or longer.
- t4 signals the start of the Host wanting to initiate communications (arbitration) with the picoNode. The Host raises MRQ to turn on various circuitries. The time between t3-t4 is 10 ms, or longer.
- t5 After the assertion of t4, the picoNode begins its "wake sequence." The picoNode must boot, initialize its operating system and hardware and when it is ready for

communications it raises its SRDY signal back to the Host. At this point, communications (Arbitration) can begin.

 t6 At this point the picoNode signals its readiness by asserting the SRDY pin. The Host can now begin communications with the picoNode.

5.5 Wake Sequence

The picoNode will often go into sleep mode even though ON_OFF is "high" (active). To begin Node-Host communications, the host must ensure the picoNode is awake. The picoNode can be awakened in two manners:

- MRQ assertion from the Host. The Host desires communications with the picoNode and awakens the picoNode by asserting the MRQ line. This is a Synchronous Wake Sequence.
- The picoNode can "self-awaken" due to network events. In this case, a timer internal to the picoNode "pops" and triggers the picoNode to "wake." When the picoNode is awake it asserts its SRDY as a matter of course to indicate to the Host (if it needs to) that it can start communicating with the picoNode while it is awake. This is an Asynchronous Wake Sequence.

5.5.1 Wake Sequence (Synchronous)

The following sequence demonstrates the timing required of the Host to awaken the picoNode from a sleep state.

Assumptions:

- The picoNode has been previously Powered On and Arbitrated.
- The power (VBATT) has remained stable and the picoNode has not been Reset (Reset is set to tri-state/float).



Figure 4 Host-Initiated picoNode Wake Sequence – SRDY Low (Synchronous)

The timing sequence shown in Figure 4 Host-Initiated picoNode Wake Sequence – SRDY Low (Synchronous). **NOTE:** The timing shown in the figure is not to scale.

- **t0** The Host desires to wake the picoNode and asserts MRQ high.
- t0 → t1 After MRQ has gone High, the Host's SPI system and other I/O can be enabled. Asserting the MRQ has enabled the internal I/O power supply of the picoNode and the Host's SPI can be enabled 4 ms after the rise of MRQ.
- t1 → t2 After the initial assertion of MRQ, the picoNode has to internally power up and initialize its systems. When it is ready to communicate it will assert its SRDY line to signal it is now ready for SPI interaction. From MRQ assertion until the picoNode is ready, takes about 80 ms.
- t2 The picoNode is now ready to fully communicate with the Host.

5.5.2 Wake Sequence (Asynchronous)

In this scenario, the picoNode is already awake due to a networking event (SRDY is already High) and the Host wants to communicate with the picoNode while it is awake. The Host asserts MRQ to ensure that the picoNode stays awake during its communication cycle.



NOTE: The timing shown in the figure is not to scale.

Figure 5 Host-Initiated picoNode Wake Sequence – SRDY High (Asynchronous)

5.6 Host-Driven Reset Sequence

If the picoNode fails to communicate (or similar), it may be necessary to "Reset" the picoNode. The following figure shows the proper sequence to reset the device.

NOTE 1: Resetting the device causes it to go through an RPMA Cold Acquisition process to reacquire the network.



NOTE 2: The timing shown in the figure is not to scale.

Figure 6 Host-Driven Reset Sequence

5.7 Host MRQ Release/picoNode Allowed to Sleep Sequence

If the Host determines there are no more messages or SPI transactions required, it nominally deasserts the MRQ to allow the picoNode to fall back to Deep Sleep (lowest power mode). The figure below shows how this is sequenced by the Host/picoNode. A small delay in de-asserting SRDY is enforced to prevent quick toggling (waking) of the picoNode.

NOTE: The timing shown in the figure is not to scale.





6 Power States

Compal can provide partners with tools to accurately predict battery life for their particular application. We recommend that partners use these tools. This section is provided for reference only.

The picoNode has a number of states it runs through during its various operating modes.

General comments:

- The picoNode accepts a wide input voltage range (2.5 V 5.5 V on VBATT).
- The picoNode has low drop out (LDO) regulators that will operate 100% of the time the picoNode is powered (ON_OFF signal set to high).
- The picoNode requires either a fixed or switched 3.3 V supply on its 3V3 pin.

The picoNode always tries to minimize its power consumption but is largely driven by network operating states and modes of operation. This document does not describe all of the modes in detail but, in general, there are two main operating modes for the picoNode:

1. Continuous Mode

In this mode, the picoNode is ON (awake) at least 50% of the time (100% of its RX cycle). The picoNode starts up, searches for the network, locks on, and Joins. In this mode, the picoNode is nominally in RX or TX mode (radio is ON and in a high power consumption state), or in an Idle state where the clocks and CPU are ON but the radio is OFF (moderately low power mode). The continuous mode is usually for applications where the Host and picoNode are AC-powered and system current consumption is not an issue.

2. Slotted Mode

This mode has the picoNode falling into a Deep Sleep state—the lowest power state of the picoNode. In this mode, the picoNode is mostly powered down except for a couple of low power LDO Regulators. The picoNode can sleep for hours at a time if the network is configured to allow this.

The power states are described in the following sections.

6.1 Operating States

This section describes the various operating states within the operational modes.

6.1.1 Power Off State

When the picoNode is totally non-functional, the Host can set the ON_OFF signal Low to deactivate the circuitry of the picoNode. This should NOT be confused with Deep Sleep states where the picoNode mostly sleeps yet maintains key network timers to wake up synchronously with network activity. If awakened from the Power Off state, the picoNode must go through a very power-hungry search/acquisition algorithm to re-acquire the RPMA Network.

6.1.2 Deep Sleep State

The picoNode shuts off all its power regulators except a couple low quiescent LDO regulators. These regulators keep a minimal amount of circuitry alive for tracking network timers, enable a 32 kHz clock, and some minor interface circuitry.

6.1.3 Oscillator Calibration State

When the picoNode is in Deep Sleep state, it attempts to maintain accuracy of its low power 32 kHz clock to enable faster network synchronizing when it wakes up. The CPU of the picoNode is not activated during this calibration state. The picoNode will periodically (and briefly) wake up in a very low power mode to calibrate its 32 kHz clock to its very accurate 26 MHz clock. This is especially important when the temperature varies substantially causing the 32 kHz oscillator to drift.

This plot is an example of the picoNode performing a self-calibration of its 32 kHz oscillator. The pulses represent the TCXO being turned on periodically to perform the calibration. The picoNode wakes itself from Deep Sleep, Calibrates, and then falls back to sleep. Minimal power is consumed during this self-calibration process. As can be seen, the picoNode does this approximately every 900 seconds.

6.1.4 Idle State

Idle state has various sub-states but generally refers to a state where the picoNode is "awake" and its system clock is on, the CPU is awake, but the RF is OFF.

6.1.5 RX State

The picoNode turns on all its clocks, the main CPU and the RF in an RX-only state. The RF transceiver, in RX state, consumes a moderate amount of power.

6.1.6 TX State

When the picoNode transmits, it uses a variable transmit power that is correlated to its received RSSI. In this state, the picoNode is likely at its highest power states, but this is somewhat dependent on RSSI. This is at approximately 23 dBm output power. This is the highest power state for the picoNode.

7 SPI Messaging Protocol

The SPI Messaging between the Host processor and the node is controlled/handled by the Host Common Library, compiled onto the user's apps processor. The following section (all of Chapter 7) is informational only.

The details of Host/Node messaging are typically not necessary for integrators to implement.

7.1 Arbitration

Arbitration is the process a Host uses to signal to the Node that it supports the bi-directional messaging protocol. The arbitration sequence is designed to reduce the probability that an arbitrary non-Host transfer sequence can mirror a valid arbitration sequence.

Arbitration consists of both Host and Node transmitting an arbitration request/reply pair. After a defined turn-around delay, both transmit a validation request/reply. The turn-around delay avoids race conditions between Host and Node and provides enough time to allow ISR execution to complete before the next SPI transfer.

If the Node does not reply to the Host request, the Host needs to wait for a turn-around delay and retry the arbitration request.

The Host must perform the arbitration sequence before any other SPI Bus communication can take place between the Host and the Node.

The Host must initiate this arbitration sequence on boot up. Additionally, the Host must perform the arbitration sequence when the Node sends to the Host an arbitration message. This can occur due to the Node going into Deep Sleep and then waking up. Since the Node requires the arbitration sequence after waking from Deep Sleep and since the Host is not aware of when the Node goes to Deep Sleep, the Host must be able to detect that the Node is requesting arbitration and the Host must then reset its Host interface state machine and perform arbitration. For more information on the Host interface SPI bus state machine, refer to section 7.3: Host Interface SPI Bus State Machine.

7.2 Message Protocol

Host-to-Node transfers use master message command pairs and Node-to-Host transfers use slave message command pairs. Both transfers use identical command sequences with only the encoding of the commands differing. The command sequence for a message transfer consists of a request/acknowledgement pair followed by a defined turn-around delay and then a message composed of a header pair and a payload.

Variable length payloads are supported by encoding the payload size in the second half of the message request. The second half of the message reply contains the available receive buffer size. If the message payload size exceeds the receive buffer size, then a new request must be made after a turn-around delay with a payload size that does not exceed the receive buffer size.

After a successful message request transfer, the Host waits a turn-around delay and then initiates the transfer with a message header command. The payload immediately follows the header and, if necessary, is zero padded to match the payload size indicated in the message request. After the payload, the Host waits a turn-around delay before proceeding with any other further messages.

The Host interface SPI bus is a standard SPI bus (with MISO, MOSI, CS, and SCLK) with the addition of three lines (MRQ, SRQ, and SRDY). These three additional lines are used to provide the Host with the ability to wake up the Node over the SPI Bus as well as providing the Node with the ability to prompt the Host to begin a SPI Bus transaction. The Node is also exceptional in that it must be the only slave present on the SPI Bus, since MOSI, CS, and SCLK must be undriven (tri-stated) any time that MRQ is low.

Before any message is communicated over the SPI Bus, the MRQ and SRDY lines must be high. The Host guarantees this by pulling the MRQ line high and waiting for the Node to pull the SRDY line high. The Host cannot proceed with SPI Bus communication until both of these lines are high. Once MRQ and SRDY are high, the Host, being SPI Bus master, can continue with a normal SPI Bus transaction.

When the Node wishes to communicate with the Host, it pulls the SRQ line high. The Host must have the ability to detect this and start a SPI Bus transaction (by first pulling the MRQ high and waiting for SRDY to go high). A standard SPI Bus transaction is described and illustrated in Figure 2.

Message exchanges between Host and Node are shown below in Figure 1.

PicoNode Integration Specification



Figure 1 SPI Master and Slave Message Sequences

In each of the request/acknowledge command pairs shown, the top command is transmitted by the Host (master) and the bottom command is transmitted by the Node (slave). The wait bubbles indicate a predefined turn-around delay which provides ISR processing time and avoids race conditions between Host and Node.

7.3 Host Interface SPI Bus State Machine

This section illustrates the sequence of messages that can take place on the Host interface SPI bus. The design and implementation of the actual state machine on the Host software is up to the Host software designer. This diagram is provided to demonstrate the message sequence over the SPI Bus. Note the usage of the turn-around delay, which is required in between each step of message exchange. This delay is required by the Node and is currently defined as having a time of 200 μ s.



Figure 9 Host Interface SPI Bus State Machine

7.4 SPI Bus Timing Example

This section provides an example illustration of an exchange of messages first from master (Host) to slave (Node) and then from slave (Node) to master (Host). Each step in the timing sequence is described below:

Figure 20 SPI Timing Example

Note that MRQ state transitions must respect the timing requirements shown in Chapter 5.

The following items pertain to the numbered bubbles above:

- 1. Host has a message that it desires to send to Node. The first thing that it does is drive MRQ and CS high.
- 2. The Host then waits for the Node to drive SRDY high. No SPI bus transaction with the Node can occur before this.
- 3. After SRDY is high, the Host can start with the SPI data transaction. This is accomplished by driving the Node CS line low and then having the Host toggle the SCLK, and MOSI lines and having the Node toggle the MISO line according to the data to be transferred. The SPI Host interface specifies that first a MMsg pair is exchanged.
- 4. A MHdr pair is exchanged. Note that the payload of the message is appended to the MHdr.
- The Host detects that the transaction is complete and that it does not wish to send more messages to the Node at this time. It drives the MRQ line low. Since MRQ is low, CS, SCLK and MOSI are tri-stated.
- 6. At some time in the future, the Node desires to send a message to the Host. It indicates this to the Host by driving SRQ high. Since SRQ is high, the Host drives MRQ and then CS high. It then waits for SRDY to go high, which it already is.

- 7. The Host starts the SPI data transaction. This is accomplished by driving the Node CS line low and then having the Host toggle the SCLK, and MOSI lines and having the Node toggle the MISO line according to the data to be transferred. The SPI Host interface specifies that first a SMsg pair is exchanged.
- 8. A SHdr pair is exchanged. Note that the payload of the message is appended to the SHdr.
- 9. The Node detects that the transaction is complete and that it does not wish to send more messages to the Host at this time. It drives the SRQ line low.
- 10. The Host detects that SRQ has gone low and that it does not have any messages to send to the Node. It drives the MRQ line low. Since MRQ is low, CS, SCLK and MOSI are tri-stated.
- 11. The Node drives the SRDY line low after MRQ goes low.

7.5 Host Message SPI Example

This section provides an example Host message exchange from master (Host) to slave (Node). In this example, the Host is sending a version request message.

This example is a zoomed-in view of the example provided previously in Figure 2. This section covers what happens in step 3, which includes the two SPI exchanges initiated by the Host.

With any SPI Host interface message, first an MMsg or SMsg pair must be exchanged. This pair contains information on how big the message is (from the message originator) and how much message queue space is available (on the message destination).

The following diagram shows such an example:



Figure 31 Host Message on SPI – MMsg Pair

The SPI clock edging is configurable with a polarity and phase. In order to communicate with the Node, the SPI clock polarity must be set to "the inactive state value of SPI clock is logic level zero" and the SPI clock phase must be set to "data is captured on the leading edge of SPI clock and changed on the following edge of SPI clock." This means that the data lines (both MISO and MOSI) are read on the SCLK rising edge and are set or cleared on the SCLK falling edge, and is commonly referred to as CPOL=0, CPHA=0.

This illustration shows that the bit streams for MISO and MOSI are:

- MISO: 011010011111111
- MOSI: 101010010000100

These bits indicate:

MISO: from slave to master (01)

length of message=2 (10)

opcode=MMsgACK (1001)

buffer size=255 (1111111)

MOSI: from master to slave (10)

length of message=2 (10)

opcode =MMsgREQ (1001)

payload size=4 (00000100)

An MMsg pair or SMsg pair is immediately followed by the corresponding MHdr pair or SHdr pair. This is illustrated below:



Figure 4 Host Message on SPI – MHdr Pair

For purpose of brevity, this timing diagram shows only a portion of the data exchange. The complete bit streams for MISO and MOSI are as follows:

These bits indicate:

- MISO: from slave to master (01) length of message=2 (10) opcode=MHdrACK (1010) Hard coded byte=1 (0000001) Unused Extra Data (0000.....0)
- MOSI: from master to slave (10) length of message=2 (10) opcode =MhdrREQ (1010) Hard coded byte=1 (00000001)

Payload:

length=8 (000010000000000) message type=VERSION (00010101000000) trailing sequence (111100001111000010100101010101)

- The payload is Little Endian. The least significant byte is transmitted over SPI first.
- All MHdr and SHdr payloads are terminated by the fixed trailing sequence 11110000111100001010010110100101.
- The example above shows a message going from master to slave, thereby having a
 payload in the master to slave direction appended at the end of the MhdrREQ and no
 payload appended at the end of the MhdrACK.

7.6 Host Message "Connect" SPI Example

This section provides an example Host message exchange of the CONNECT message from master/Host to slave/Node and subsequent response from the slave to the master.

The timing is similar to the timing illustrated in the previous section, but the data and length of data is different.

The steps involved in this exchange are as follows:

The Host desires to send the CONNECT message to the Node. As described in the previous section, this starts with an MmsgREQ/MmsgACK exchange over the SPI bus.

- MISO: 011010011111111
- MOSI: 1010100100000110

These bits indicate:

MISO: from slave to master (01)

length of message=2 (10)

opcode=MMsgACK (1001)

buffer size=255 (1111111)

MOSI: from master to slave (10)

length of message=2 (10)

opcode =MmsgREQ (1001)

payload size=6 (0000110)

The MMsg exchange is followed by the MHdr exchange, which includes the payload of the CONNECT message.

These bits indicate:

MISO: from slave to master (01)

length of message=2 (10) opcode=MHdrACK (1010) Hard coded byte=1 (00000001) Unused Extra Data (0000......0)

- MOSI: from master to slave (10)
 - length of message=2 (10)
 - opcode =MhdrREQ (1010)
 - Hard coded byte=1 (0000001)

Payload:

length=12 (000011000000000)

- message type=CONNECT (001100100100000)
- host interface=True (000000100000000000000000000000)
- trailing sequence (1111000011110000101001010101010)

The payload of the message includes first the length, which is the number of bytes in the payload including the length and the trailing sequence.

It is followed by the message type, which in this case is 0x4032, and corresponds with CONNECT.

The CONNECT message has a 4-byte field that is a Boolean flag specifying whether or not the Node should send asynchronous SPI messages to the Host. To specify that the Node should send messages to the Host, the value of oxooooooo1 is used.

It is then followed by the standard fixed trailing sequence.

This message exchange is followed by a Node-initiated message exchange for the purpose of sending an ACK of the CONNECT message to the Host. This starts with a SmsgREQ/SMsgACK exchange over the SPI bus.

- MISO: 0110101100000100
- MOSI: 101010111111111

These bits indicate:

MISO: from slave to master (01)

length of message=2 (10) opcode=SMsgACK (1011) buffer size=255 (1111111) payload size=4 (00000100)

MOSI: from master to slave (10) length of message=2 (10) opcode =SmsgREQ (1011)

buffer size=255 (1111111)

The SMsg exchange is followed by the SHdr exchange, which includes the payload of the ACK message.

These bits indicate:

MISO: from slave to master (01)

length of message=2 (10)

opcode=SHdrACK (1100)

Hard coded byte=1 (0000001)

Payload:

length=8 (00001000000000)

message type=ACK (00110000000000)

trailing sequence (11110000111100001010010110100101)

MOSI: from master to slave (10)

length of message=2 (10)

opcode =ShdrREQ (1100)

Hard coded byte=1 (0000001)

Unused Extra Data (oooo.....o)

8 picoNode Provisioning

Provisioning a node consists of updating (if necessary) node firmware version, applying a configuration to the node, and providing security keys to the node. The node configuration and security keys must match the target network where the node will be deployed and, after provisioning, the node-specific key must be provided to the network operator. The provisioning software package is used to perform these steps.

In the near future, the provisioning step will not be required for Global and U.S. public networks. Pre-Provisioned Nodes will be the default of what ships from the factory.

9 Antenna Diversity

The picoNode supports Antenna Diversity for optimal System performance. In many cases, the picoNode and Host system are mounted in fixed locations that often experience nulls in the RF coverage. Antenna Diversity can help with optimization of the RX and TX paths. In marginal coverage areas, an RF null could easily disadvantage the picoNode to force it to transmit at a higher TX Power (more battery energy) or cause network loss and frequent rescanning to reacquire the network (again, more energy).



Figure 13 Antenna Diversity

9.1 Antenna Design Considerations

 Good antenna design is also crucial to success. It is important to consider some pertinent issues. Compal and our partners provide a service to review your PCB layout to ensure good RF design practices have been followed. Compal and our partners provide a service to field test the antenna performance of your device, vs. a perfect reference. Over time we have found that good RF performance is possible, even in tight enclosures, if we work cooperatively.

- Ceramic antennas can work well but may sometimes have issues. Careful testing must be done to ensure desired gains and radiation patterns.
- The product must be researched in conjunction to the Access Point, its deployment, and its antenna radiation pattern. Nominally the Access Point will be mounted on a tower or mountain with a downward tilt. The picoNode and System may be mounted vertically or horizontally—forcing requirements on the optimal radiation pattern of the picoNode.
- The antenna must be well matched and with low loss between picoNode and antenna. It is important to follow the manufacturer's recommendations. The use of low tolerance ceramic capacitors and low tolerance thin film inductors are recommended. Examples include the Murata GJM series of capacitors and LQP series of inductors. If using stripline RF port feeds, care must be employed to ensure low loss and proper impedance. The antenna match may change when fully integrated into a product. It is advised to recheck the match after full integration. During tuning this may require the use of so called "RF pigtails" in an ad hoc fashion. If the Bill of Materials (BOM) cost will allow, a special connector can be implemented to support this verification/optimization.
- Metallic objects nearby to the antenna can affect radiation gains, patterns, and power match. Typically anything within about 4-5 inches can affect the match significantly particularly if the nearby metal is resonant at 2.4 GHz. A little pattern distortion usually is not of too much concern unless deep wide angular nulls in the antenna pattern results. Other types of pattern distortion can be caused by absorptive losses due to lossy dielectrics nearby the antenna, which represents real power loss dissipated as heat in the loss object. This represents power that is completely lost and not radiated in a useful direction.
- Noisy System clocks with harmonics can fall into the operating band of the picoNode and can be picked up by the antennas—degrading sensitivity, or causing Electromagnetic Compatibility (EMC) regulatory issues.
- It is best to put chip antennas on the side of the PCB that faces outward in your enclosure. If the antennas are on the opposite side of the PCB from the picoNode, make sure to use the proper PCB via for 50 ohms.

9.2 Diversity Considerations

The operating frequency of the picoNode is the ISM 2.4 GHz band. This has a wavelength of 12.3 cm in air. For optimal null/peak diversity detection, the antennas must be separated by at least 2.5" (5 cm). It is a good idea on the diversity antenna to orient it 90 degrees from the main antenna in order to improve the polarization diversity and increase spatial de-correlation.

Practical ground plane-independent antennas are preferable to those that require the printed circuit board (PCB) copper for the antenna counterpoise. Examples of these are dipole antennas and some chip patch antennas. However these can be cost adders in certain cases. It should be noted however that some chip antennas that use the PCB for ground return have been shown to produce reasonable performance.

9.3 Block Diagram

Some regulatory domains require a block diagram of the module for their documentation similar to that shown in the following figure.



Figure 14 picoNode Block Diagram

9.4 Antennas

9.4.1 Host Antenna Trace Design

To connect RF signal from the module's RF pins to the antenna, RF connector, or matching network, 50 ohm-controlled impedance traces should be used. Vias may be used to transition the RF from one signal layer to another signal layer.

The following controlled impedance RF traces are allowed:

- Surface microstrip
- Surface coplanar waveguide (CPWG)
- Surface ground-backed CPWG
- Embedded microstrip
- Embedded CPWG

- Embedded ground-backed CPWG
- Stripline
- Asymmetrical stripline

9.4.2 Controlled Impedance Trace Design

9.4.2.1 Dimensions

The controlled impedance trace should have dimensions that correspond to 50 ohm \pm 10% no matter what the specific transmission line type is. The specific design depends on trace width, thickness, dielectric constant, and distance to grounds. These dimensional factors are controlled by the PCB fabrication shop. From a design perspective, on the PCB fabrication drawing, clearly mark that the microstrip trace requires controlled impedance of 50 ohm \pm 10%. Also specify that the PCB shop must test to confirm that the impedance is within 10% of 50 ohms.

9.4.2.2 Ground Planes

All transmission line types except surface CPWG and embedded CPWG requires ground planes either above the RF trace, below the RF trace, or both above and below the RF trace. In these cases with ground planes, the RF trace must not cross any splits in the ground plane. The ground plane directly above and below the RF trace must be continuous.

9.4.2.3 Matching Network

A matching network may be used at the module's RF pin to ensure good 50 ohm transition between the host board and the module. Another matching network may be used on the antenna side of the controlled impedance trace to match the antenna to 50 ohms.

9.4.2.4 Stitching Vias

Ample ground stitching vias should be used around the connectors, trace, and module. It is recommended that a via stitching pattern be used with no more than 0.25" separation between adjacent vias.

9.4.2.5 Isolation

Since the RF signal coming from the module has already been filtered, there is no isolation requirement.

9.4.2.6 Example RF Trace Design

The following figure shows surface-ground backed CPWG design with 22 mil trace width, 15 mil gap-to-surface ground plane, 2 mil thick top layer (1/2 oz copper + plating), 14 mil dielectric thickness to solid ground plane underneath, and a FR-4 dielectric constant of 4.4. This configuration achieves the desired 50 ohm ± 10% for the controlled impedance. Note that in

this design, a pi-matching network exists to match the antenna. This is recommended but not necessary if the antenna is already matched with VSWR of 1.5:1 or better. Note that antennas should not have gain at the harmonic frequencies and only dipole-type antennas should be used. Antennas with gain at the fundamental frequency of more than 9 dBi should not be used.



Figure 15 Controlled Impedance RF Trace Design

Other controlled impedance RF trace configurations are acceptable as long as the characteristic impedance stays within the boundary limit of 50 ohm \pm 10%. Most PCB fabrication shops can adjust the trace width of the controlled impedance traces such that 50 ohms is maintained. The PCB shop must guarantee the impedance of the traces to within the 50 ohm \pm 10% boundary limit by impedance testing the PCBs.

10 Regulatory Considerations

The picoNode uses LGA pads for its RF ports. This lowers the unit cost and provides greater host configuration flexibility in the final application. Modular certifications (FCC, ETSI, and others) have been obtained for the picoNode. The existence of the modular certification minimizes cost and time to market for customers. The certification documents and the results of the certification tests are available to system integrators upon request.

The modular certification of the picoNode can be re-used by customers that utilize an equivalent layout and stack-up as our reference design platform. Additional testing for verification purposes may still be required per FCC or other regulatory body guidelines and requirements but will vary on a case-by-case basis. Customers are advised to consult with the regulatory compliance test house of their choice for the best way to proceed.

10.1 FCC Warnings

This device complies with part 15 of the Federal Communications Commission (FCC) Rules. Operation is subject to the following two conditions:

- 1. This device may not cause harmful interference.
- 2. This device must accept any interference received, including interference that may cause undesired operation.

Changes or modifications not expressly approved by the manufacturer could void the user's authority to operate the equipment.

NOTE:

This equipment has been tested and found to comply with the limits for a Class B digital device, pursuant to Part 15 of the FCC Rules. These limits are designed to provide reasonable protection against harmful interference in a residential installation.

WARNING:

This equipment generates, uses, and can radiate radio frequency energy. If not installed and used in accordance with the instructions, this equipment may cause harmful interference to radio communications. However, there is no guarantee that interference will not occur in a particular installation. If this equipment does cause harmful interference to radio or television reception, which can be determined by turning the equipment off and on, the user is encouraged to try to correct the interference by one or more of the following measures:

- Re-orient or relocate the receiving antenna.
- Increase the separation between the equipment and receiver.
- Connect the equipment into an outlet on a circuit different from that to which the receiver is connected.
- Consult the dealer or an experienced radio/TV technician for help.

IMPORTANT NOTICE

This transmitter module is authorized only for use in devices where the final end product must be labeled in visible area with the following:

"Contains FCC ID: GKR-PICONODE"

The picoNode may transmit simultaneously with other collocated radio transmitters within a host device, provided the following conditions are met:

1. Each collocated radio transmitter has been certified by FCC for mobile application.

2. At least 20 cm separation distance between the antennas of the collocated transmitters and the user's body must be maintained at all times.

As long as the above conditions are met, further transmitter test will not be required.

However, the OEM integrator is still responsible for testing their end-product for any additional compliance requirements required with this module installed.

In the event that these conditions cannot be met (for example co-location with another transmitter), then the FCC authorization is no longer considered valid and the FCC ID cannot be used on the final product.

In these circumstances, the OEM integrator will be responsible for re-evaluating the end product (including the transmitter) and obtaining a separate FCC authorization.

The OEM integrator has to be aware not to provide information to the end user regarding how to install or remove this RF module in the user's manual of the end product which integrates this module.

The end user manual shall include all required regulatory information/warning as shown in this document.

11 Errata

Degraded RF Channels

Ingenu and their partners are deploying networks globally. They carefully manage the frequency allocations of these networks to avoid using the 3 (out of 38 possible) RF channels described below. RPMA networks can deploy overlapping coverage (ie. all Access Points) on a single 1MHz RF channel. The other channels are held in reserve for future use or special cases. Three lower performing channels are listed below, for reference.

The picoNode uses a Channel scheme such as the following:

- Channel 1 = 2402 MHz and each successive channel is 1.99 MHz offset to that Channel 1.
- Channel 2 = 2403.99 MHz
- Channel 3 = 2405.98 MHz
- Etc.

The picoNode uses a 26 MHz reference clock for processing and for the direct conversion radio. It has been found that 26 MHz harmonics can create strong tones that cause some RF sensitivity degradation on these harmonic channels.

- 93*26 MHz = 2418 MHz. This affects channel 9.
- 94*26 MHz = 2444 MHz. This affects channel 22.
- 95*26 MHz = 2470 MHz. This affects channel 35.

System integrators should NOT use these 3 channels as picoNode RX sensitivity can be degraded by a nominal 3-10 dB.

Appendix A Abbreviations and Terms

Abbreviation/Term	ation/Term Definition	
ACPR	Adjacent Channel Power Ratio	
AGC	Automatic Gain Control	
ALC	Automatic Level Control	
AP	Access Point	
API	Application Programming Interface	
ASIC	Application-Specific Integrated Circuit	
ATE	Automated Test Equipment	
вом	Bill of Materials	
BW	Bandwidth	
CCATS	Commodity Classification Automated Tracking System. An <u>alphanumeric code</u> assigned by the <u>Bureau of Industry and Security</u> (BIS) to products that it has classified against the <u>Export</u> <u>Administration Regulations</u> (EAR).	
CMOS	Complementary Metal-Oxide-Semiconductor	
CPOL	Clock Polarity (for SPI)	
CPU	Central Processing Unit	
DBPSK	Differential Binary Phase Shift Keying	
DFS	Dynamic Frequency Selection	
DPLL	Digital Phase-Locked Loop	
DSSS	Direct Sequence Spread Spectrum	
EIRP	Effective Isotropic Radiated Power	
EMC	Electromagnetic Compatibility	
ESD	Electrostatic Discharge	
ESR	Equivalent Series Resistance	
ETSI	European Telecommunications Standards Institute	
EVM	Error Vector Magnitude	
FCC	Federal Communications Commission	
FER	Frame Error Rate	
GND	Ground	
GPIO	General Purpose Input/Output	
НВМ	Human Body Model	
ISED (IC)	Innovation, Science and Economic Development Canada (Formerly Industry Canada	
IIP ₃	Input Third-Order Intercept Point	

PicoNode Integration Specification

Abbreviation/Term	Definition	
LDO	Low Drop Out	
LNA	Low Noise Amplifier	
LO	Local Oscillator	
MISO	Master Input, Slave Output	
MM	Machine Model	
MOSI	Master Output, Slave Input	
MRQ	Master Request	
MSL	Moisture Sensitivity Level	
picoNode	A small form factor, wireless network module developed by Compal that works in combination with various devices and sensors and communicates data to an Access Point.	
Node	The generic term used interchangeably with picoNode.	
ΟΤΑ	Over-the-Air	
Po	"Power Output" for the RF Transmitter	
PA	Power Amplifier	
PAPR	Peak-to-Average Power Ratio	
РСВ	Printed Circuit Board	
POR	Power On Reset	
QoS	Quality of Service	
RF	Radio Frequency	
RFIC	Radio Frequency Integrated Circuit	
RMA	Return Merchandise Authorization	
RoHS	Restriction of Hazardous Substances	
RPMA	Random Phase Multiple Access. The Ingenu proprietary wireless communication technology and network.	
RP-SMA	Reverse Polarity Subminiature version A connector	
RSSI	Receive Signal Strength Indicator	
RT	Remote Terminal	
RTC	Real Time Clock	
RX	Receive/Receiver	
SCLK	Serial Clock	
SMT	Surface Mount Technology	
SNR	Signal-to-Noise Ratio	
SPI	Synchronous Peripheral Interface	
SRAM	Static Random Access Memory	
SRDY	Slave Ready	
SRQ	Slave Request	

PicoNode Integration Specification

Abbreviation/Term	Definition
ТСХО	Temperature Compensated Crystal Oscillator
ТХ	Transmit/Transmitter
UART	Universal Asynchronous Receiver/Transmitter
VCO	Voltage Controlled Oscillator
VSWR	Voltage Standing Wave Ratio
ХО	Crystal Oscillator

Appendix B PCB Land Pattern and Keep-outs



Figure 16 picoNode PCB Land Pattern



Keep out area



Figure 17 picoNode PCB Land Pattern Keep-outs

Figure shows keep out areas for the Host PCB.

No silk-screen markings are to be located under the area of the module.

It is desirable to keep traces and pads/Vias away from these areas as the picoNode uses these areas as Test Points on the underside of its PCA. Some signals are high frequency and could induce noise onto Host traces or contamination at time of picoNode to Host reflow could cause long term problems.

Appendix C Application Diagrams/Schematics

The diagrams/schematics in this appendix are provided as examples.



Figure 5 Powered Example



Figure 6 Lithium Battery Example

Appendix D picoNode Mechanical Drawing



PIN2	WAkE
PIN3	GND
PIN4	GND
PIN5	GND
PIN6	RF_TXENA
PIN7	GND
PIN8	SRQ
PIN9	SRDY
PIN10	GND
PIN11	MRQ
PIN12	GND
PIN13	ON_OFF
PIN14	GND
PIN15	+VBATT
PIN16	TIME_QUAL
PIN17	TOUT
PIN18	GND
PIN19	SPI_MISO
PIN20	SPI_MOSI
PIN21	SPI_SCLK
PIN22	SPI_CSO
PIN23	+3V3
PIN24	GND
PIN25	RF_PORT1
PIN26	GND
PIN27	RF_PORT2
PIN28	GND
PIN29	GND
PIN30	GND
PIN31	GND
PIN32	GND
PIN33	RF_SHDN
PIN34	GND

PICO NODE PIN DESCRIPTION TABLE

PIN DESCRIPTION

GND

PIN#

PIN1

Figure 20 picoNode Mechanical Dimensions