Features

- High Performance RF-CMOS 2.4 GHz Radio Transceiver Targeted for IEEE 802.15.4[™],
 ZigBee®, 6LoWPAN, RF4CE, SP100, WirelessHART[™] and ISM Applications
- Industry Leading Link Budget (104 dB)
 - Receiver Sensitivity -101 dBm
 - Programmable Output Power from -17 dBm up to +3 dBm
- Ultra-Low Current Consumption:

- SLEEP = 0.02 µA - TRX_OFF = 0.4 mA - RX_ON = 12.3 mA

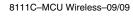
BUSY_TX = 14 mA (at max. Transmit Power of +3 dBm)

- Ultra-Low Supply Voltage (1.8V to 3.6V) with Internal Regulator
- Optimized for Low BoM Cost and Ease of Production:
 - Few External Components Necessary (Crystal, Capacitors and Antenna)
 - Excellent ESD Robustness
- Easy to Use Interface:
 - Registers, Frame Buffer and AES Accessible through Fast SPI
 - Only Two Microcontroller GPIO Lines Necessary
 - One Interrupt Pin from Radio Transceiver
 - Clock Output with Prescaler from Radio Transceiver
- Radio Transceiver Features:
 - 128-byte FIFO (SRAM) for Data Buffering
 - Programmable Clock Output, to Clock the Host Microcontroller or as Timer Reference
 - Integrated RX/TX Switch
 - Fully Integrated, Fast Settling PLL to support Frequency Hopping
 - Battery Monitor
 - Fast Wake-Up Time < 0.4 msec
- Special IEEE 802.15.4-2006 Hardware Support:
 - FCS Computation and Clear Channel Assessment
 - RSSI Measurement, Energy Detection and Link Quality Indication
- MAC Hardware Accelerator:
 - Automated Acknowledgement, CSMA-CA and Retransmission
 - Automatic Address Filtering
 - Automated FCS Check
- Extended Feature Set Hardware Support:
 - AES 128-bit Hardware Accelerator
 - RX/TX Indication (external RF Front-End Control)
 - RX Antenna Diversity
 - Supported PSDU data rates: 250 kb/s, 500 kb/s, 1 Mb/s and 2 Mb/s
 - True Random Number Generation for Security Application
- Industrial and Extended Temperature Range:
 - -40°C to +85°C and -40°C to +125°C
- I/O and Packages:
 - 32-pin Low-Profile QFN Package 5 x 5 x 0.9 mm³
 - RoHS/Fully Green
- Compliant to IEEE 802.15.4-2006 and IEEE 802.15.4-2003
- Compliant to EN 300 328/440, FCC-CFR-47 Part 15, ARIB STD-T66, RSS-210



Low Power
2.4 GHz
Transceiver for
ZigBee,
IEEE 802.15.4,
6LoWPAN,
RF4CE, SP100,
WirelessHART,
and ISM
Applications

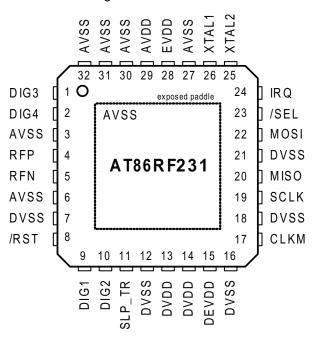
AT86RF231-ZU AT86RF231-ZF





1. Pin-out Diagram

Figure 1-1. AT86RF231 Pin-out Diagram



Note: The exposed paddle is electrically connected to the die inside the package. It shall be soldered to the board to ensure electrical and thermal contact and good mechanical stability.



1.1 Pin Descriptions

Table 1-1. Pin Description AT86RF231

	lable 1-1. Pin Description A186RF231								
Pins	Name	Туре	Description						
1	DIG3	Digital output (Ground)	 1. RX/TX Indicator, see Section 11.5 2. If disabled, pull-down enabled (AVSS) 						
2	DIG4	Digital output (Ground)	RX/TX indicator (DIG3 inverted), see Section 11.5 If disabled, pull-down enabled (AVSS)						
3	AVSS	Ground	Ground for RF signals						
4	RFP	RF I/O	Differential RF signal						
5	RFN	RF I/O	Differential RF signal						
6	AVSS	Ground	Ground for RF signals						
7	DVSS	Ground	Digital ground						
8	/RST	Digital input	Chip reset; active low						
9	DIG1	Digital output (Ground)	Antenna Diversity RF switch control, see Section 11.4 If disabled, pull-down enabled (DVSS)						
10	DIG2	Digital output (Ground)	Antenna Diversity RF switch control (DIG1 inverted), see Section 11.4 Signal IRQ_2 (RX_START) for RX Frame Time Stamping, see Section 11.6 If functions disabled, pull-down enabled (DVSS)						
11	SLP_TR	Digital input	Controls sleep, transmit start, receive states; active high, see Section 6.5						
12	DVSS	Ground	Digital ground						
13	DVDD	Supply	Regulated 1.8V voltage regulator; digital domain, see Section 9.4						
14	DVDD	Supply	Regulated 1.8V voltage regulator; digital domain, see Section 9.4						
15	DEVDD	Supply	External supply voltage; digital domain						
16	DVSS	Ground	Digital ground						
17	CLKM	Digital output	Master clock signal output; low if disabled, see Section 9.6						
18	DVSS	Ground	Digital ground						
19	SCLK	Digital input	SPI clock						
20	MISO	Digital output	SPI data output (Master Input Slave Output)						
21	DVSS	Ground	Digital ground						
22	MOSI	Digital input	SPI data input (Master Output Slave Input)						
23	/SEL	Digital input	SPI select, active low						
24	IRQ	Digital output	Interrupt request signal; active high or active low; configurable Frame Buffer Empty Indicator; active high, see Section 11.7						
25	XTAL2	Analog input	Crystal pin, see Section 9.6						
26	XTAL1	Analog input	Crystal pin or external clock supply, see Section 9.6						
27	AVSS	Ground	Analog ground						
28	EVDD	Supply	External supply voltage, analog domain						



 Table 1-1.
 Pin Description AT86RF231 (Continued)

Pins	Name	Туре	Description			
29	AVDD	Supply	Regulated 1.8V voltage regulator; analog domain, see Section 9.4			
30	AVSS	Ground	Analog ground			
31	AVSS	Ground	Analog ground			
32	AVSS	Ground	Analog ground			
Paddle	AVSS	Ground	Analog ground; Exposed paddle of QFN package			



1.2 Analog and RF Pins

1.2.1 Supply and Ground Pins

EVDD, DEVDD

EVDD and DEVDD are analog and digital supply voltage pins of the AT86RF231 radio transceiver.

AVDD, DVDD

AVDD and DVDD are outputs of the internal 1.8V voltage regulators. The voltage regulators are controlled independently by the radio transceivers state machine and are activated dependent on the current radio transceiver state. The voltage regulators can be configured for external supply.

For details, refer to Section 9.4 "Voltage Regulators (AVREG, DVREG)" on page 110.

AVSS, DVSS

AVSS and DVSS are analog and digital ground pins respectively. The analog and digital power domains should be separated on the PCB.

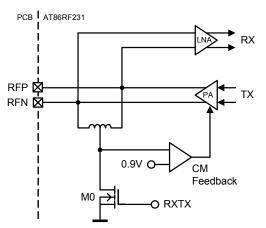
1.2.2 RF Pins

RFN, RFP

A differential RF port (RFP/RFN) provides common-mode rejection to suppress the switching noise of the internal digital signal processing blocks. At board-level, the differential RF layout ensures high receiver sensitivity by rejecting any spurious emissions originated from other digital ICs such as a microcontroller.

A simplified schematic of the RF front end is shown in Figure 1-2 on page 5.

Figure 1-2. Simplified RF Front-end Schematic



The RF port is designed for a 100Ω differential load. A DC path between the RF pins is allowed. A DC path to ground or supply voltage is not allowed. Therefore, when connecting an RF-load providing a DC path to the power supply or ground, AC-coupling is required as indicated in Table 1-2 on page 6.



The RF port DC values depend on the operating state, refer to Section 7. "Operating Modes" on page 33. In TRX_OFF state, when the analog front-end is disabled (see Section 7.1.2.3 "TRX_OFF - Clock State" on page 35), the RF pins are pulled to ground, preventing a floating voltage.

In transmit mode, a control loop provides a common-mode voltage of 0.9V. Transistor M0 is off, allowing the PA to set the common-mode voltage. The common-mode capacitance at each pin to ground shall be < 30 pF to ensure the stability of this common-mode feedback loop.

In receive mode, the RF port provides a low-impedance path to ground when transistor M0, see Figure 1-2 on page 5, pulls the inductor center tap to ground. A DC voltage drop of 20 mV across the on-chip inductor can be measured at the RF pins.

1.2.3 Crystal Oscillator Pins

XTAL1, XTAL2

The pin XTAL1 is the input of the reference oscillator amplifier (XOSC), XTAL2 is the output. A detailed description of the crystal oscillator setup and the related XTAL1/XTAL2 pin configuration can be found in Section 9.6 "Crystal Oscillator (XOSC)" on page 116.

When using an external clock reference signal, XTAL1 shall be used as input pin.

For further details, refer to Section 9.6.3 "External Reference Frequency Setup" on page 117.

1.2.4 Analog Pin Summary

Table 1-2. Analog Pin Behavior - DC values

Pin	Values and Conditions	Comments				
RFP/RFN	$V_{DC} = 0.9V (BUSY_TX)$ $V_{DC} = 20 \text{ mV (receive states)}$ $V_{DC} = 0 \text{ mV (otherwise)}$	DC level at pins RFP/RFN for various transceiver states AC coupling is required if an antenna with a DC path to ground is used. Serial capacitance and capacitance of each pin to ground must be < 30 p				
XTAL1/ XTAL2	$V_{DC} = 0.9V$ at both pins $C_{PAR} = 3 \text{ pF}$	DC level at pins XTAL1/XTAL2 for various transceiver states Parasitic capacitance (C_{PAR}) of the pins must be considered as additional load capacitance to the crystal.				
DVDD	$V_{DC} = 1.8V$ (all states, except SLEEP) $V_{DC} = 0$ mV (otherwise)	DC level at pin DVDD for various transceiver states Supply pins (voltage regulator output) for the digital 1.8V voltage domain, recommended bypass capacitor 1 μ F.				
AVDD	V _{DC} = 1.8V (all states, except P_ON, SLEEP, RESET, and TRX_OFF) V _{DC} = 0 mV (otherwise)	DC level at pin AVDD for various transceiver states Supply pin (voltage regulator output) for the analog 1.8V voltage domain, recommended bypass capacitor 1 μ F.				



1.3 Digital Pins

The AT86RF231 provides a digital microcontroller interface. The interface comprises a slave SPI (/SEL, SCLK, MOSI and MISO) and additional control signals (CLKM, IRQ, SLP_TR, /RST and DIG2). The microcontroller interface is described in detail in Section 6. "Microcontroller Interface" on page 16.

Additional digital output signals DIG1...DIG4 are provided to control external blocks, i.e. for Antenna Diversity RF switch control or as an RX/TX Indicator, see Section 11.4 "Antenna Diversity" on page 142 and Section 11.5 "RX/TX Indicator" on page 147. After reset, these pins are pulled-down to digital ground (DIG1/DIG2) or analog ground (DIG3/DIG4).

1.3.1 Driver Strength Settings

The driver strength of all digital output pins (MISO, IRQ, DIG1, DIG2, DIG3, DIG4) and CLKM pin can be configured using register 0x03 (TRX_CTRL_0), see Table 1-3 on page 7.

Table 1-3. Digital Output Driver Configuration

Pins	Default Driver Strength	Recommendation/Comment		
MISO, IRQ, DIG1,, DIG4	2 mA	Adjustable to 2 mA, 4 mA, 6 mA and 8 mA		
CLKM	4 mA	Adjustable to 2 mA, 4 mA, 6 mA and 8 mA		

The capacitive load should be as small as possible as, not larger than 50 pF when using the 2 mA minimum driver strength setting. Generally, the output driver strength should be adjusted to the lowest possible value in order to keep the current consumption and the emission of digital signal harmonics low.

1.3.2 Pull-Up and Pull-Down Configuration

All digital input pins are internally pulled-up or pulled-down in radio transceiver state P_ON, see Section 7.1.2.1 "P_ON - Power-On after VDD" on page 34. Table 1-4 on page 7 summarizes the pull-up and pull-down configuration.

Table 1-4. Pull-Up / Pull-Down Configuration of Digital Input Pins in P_ON State

Pins	H $\stackrel{}{=}$ pull-up, L $\stackrel{}{=}$ pull-down
/RST	Н
/SEL	Н
SCLK	L
MOSI	L
SLP_TR	L

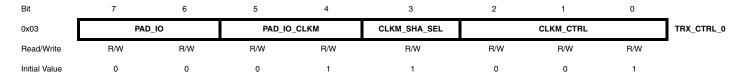
In all other radio transceiver states, no pull-up or pull-down circuitry is connected to any of the digital input pins mentioned in Table 1-4 on page 7. In RESET state, the pull-up / pull-down configuration is disabled.



1.3.3 Register Description

Register 0x03 (TRX_CTRL_0):

The TRX_CTRL_0 register controls the drive current of the digital output pads and the CLKM clock rate.



• Bit [7:6] - PAD_IO

The register bits set the output driver current of all digital output pads, except CLKM.

Table 1-5. Digital Output Driver Strength

Register Bit	Value	Description
PAD_IO	<u>O</u> ⁽¹⁾	2 mA
	1	4 mA
	2	6 mA
	3	8 mA

Note: 1. Reset values of register bits are underlined characterized in the document.

• Bit [5:4] - PAD_IO_CLKM

The register bits set the output driver current of pin CLKM. Refer also to Section 9.6 "Crystal Oscillator (XOSC)" on page 116.

Table 1-6.CLKM Driver Strength

Register Bit	Value	Description
PAD_IO_CLKM	0	2 mA
	<u>1</u>	4 mA
	2	6 mA
	3	8 mA

• Bit 3 - CLKM SHA SEL

Refer to Section 9.6 "Crystal Oscillator (XOSC)" on page 116.

• Bit [2:0] - CLKM CTRL

Refer to Section 9.6 "Crystal Oscillator (XOSC)" on page 116.



2. Disclaimer

Typical values contained in this datasheet are based on simulations and testing. Min and Max values are available when the radio transceiver has been fully characterized.

3. Overview

The AT86RF231 is a feature rich, low-power 2.4 GHz radio transceiver designed for industrial and consumer ZigBee/IEEE 802.15.4, 6LoWPAN, RF4CE and high data rate 2.4 GHz ISM band applications. The radio transceiver is a true SPI-to-antenna solution. All RF-critical components except the antenna, crystal and de-coupling capacitors are integrated on-chip. Therefore, the AT86RF231 is particularly suitable for applications like:

- 2.4 GHz IEEE 802.15.4 and ZigBee systems
- 6LoWPAN and RF4CE systems
- · Wireless sensor networks
- Industrial control, sensing and automation (SP100, WirelessHART)
- Residential and commercial automation
- · Health care
- Consumer electronics
- · PC peripherals

The AT86RF231 can be operated by using an external microcontroller like Atmel's AVR microcontrollers. A comprehensive software programming description can be found in reference [6], AT86RF231 Software Programming Model.

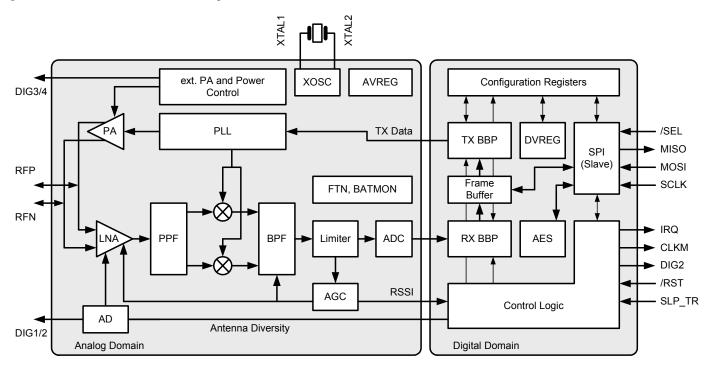


4. General Circuit Description

This single-chip radio transceiver provides a complete radio transceiver interface between an antenna and a microcontroller. It comprises the analog radio, digital modulation and demodulation including time and frequency synchronization and data buffering. The number of external components is minimized such that only the antenna, the crystal and decoupling capacitors are required. The bidirectional differential antenna pins (RFP, RFN) are used for transmission and reception, thus no external antenna switch is needed.

The AT86RF231 block diagram is shown in Figure 4-1 on page 10.

Figure 4-1. AT86RF231 Block Diagram



The received RF signal at pins RFN and RFP is differentially fed through the low-noise amplifier (LNA) to the RF filter (PPF) to generate a complex signal, driving the integrated channel filter (BPF). The limiting amplifier provides sufficient gain to drive the succeeding analog-to-digital converter (ADC) and generates a digital RSSI signal. The ADC output signal is sampled by the digital base band receiver (RX BBP).

The transmit modulation scheme is offset-QPSK (O-QPSK) with half-sine pulse shaping and 32-length block coding (spreading) according to [1] and [2]. The modulation signal is generated in the digital transmitter (TX BBP) and applied to the fractional-N frequency synthesis (PLL), to ensure the coherent phase modulation required for demodulation of O-QPSK signals. The frequency-modulated signal is fed to the power amplifier (PA).

A differential pin pair DIG3/DIG4 can be enabled to control an external RF front-end.

Two on-chip low-dropout voltage regulators (A|DVREG) provide the analog and digital 1.8V supply.



An internal 128-byte RAM for RX and TX (Frame Buffer) buffers the data to be transmitted or the received data.

The configuration of the AT86RF231, reading and writing of Frame Buffer is controlled by the SPI interface and additional control lines.

The AT86RF231 further contains comprehensive hardware-MAC support (Extended Operating Mode) and a security engine (AES) to improve the overall system power efficiency and timing. The stand-alone 128-bit AES engine can be accessed in parallel to all PHY operational transactions and states using the SPI interface, except during SLEEP state.

For applications not necessarily targeting IEEE 802.15.4 compliant networks, the radio transceiver also supports alternative data rates up to 2 Mb/s.

For long-range applications or to improve the reliability of an RF connection the RF performance can further be improved by using an external RF front-end or Antenna Diversity. Both operation modes are supported by the AT86RF231 with dedicated control pins without the interaction of the microcontroller.

Additional features of the Extended Feature Set, see Section 11. "AT86RF231 Extended Feature Set" on page 128, are provided to simplify the interaction between radio transceiver and microcontroller.

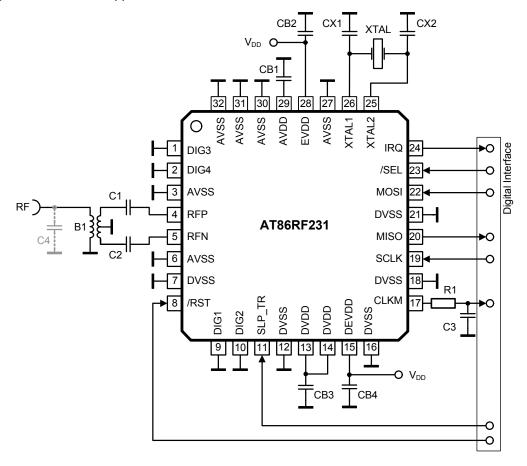


5. Application Circuits

5.1 Basic Application Schematic

A basic application schematic of the AT86RF231 with a single-ended RF connector is shown in Figure 5-1 on page 12. The 50Ω single-ended RF input is transformed to the 100Ω differential RF port impedance using balun B1. The capacitors C1 and C2 provide AC coupling of the RF input to the RF port, optional capacitor C4 improves matching if required.

Figure 5-1. Basic Application Schematic



The power supply decoupling capacitors (CB2, CB4) are connected to the external analog supply pin (EVDD, pin 28) and external digital supply pin (DEVDD, pin 15). Capacitors CB1 and CB3 are bypass capacitors for the integrated analog and digital voltage regulators to ensure stable operation. All decoupling and bypass capacitors should be placed as close as possible to the pins and should have a low-resistance and low-inductance connection to ground to achieve the best performance.

The crystal (XTAL), the two load capacitors (CX1, CX2), and the internal circuitry connected to pins XTAL1 and XTAL2 form the crystal oscillator. To achieve the best accuracy and stability of the reference frequency, large parasitic capacitances should be avoided. Crystal lines should be



routed as short as possible and not in proximity of digital I/O signals. This is especially required for the High Data Rate Modes, refer to Section 11.3 "High Data Rate Modes" on page 137.

Crosstalk from digital signals on the crystal pins or the RF pins can degrade the system performance. Therefore, a low-pass filter (C3, R1) is placed close to the CLKM output pin to reduce the emission of CLKM signal harmonics. This is not needed if the CLKM pin is not used as a microcontroller clock source. In that case, the output should be turned off during device initialization.

The ground plane of the application board should be separated into four independent fragments, the analog, the digital, the antenna and the XTAL ground plane. The exposed paddle shall act as the reference point of the individual grounds.

Table 5-1. Example Bill of Materials (BoM) for Basic Application Schematic

Designator	Designator Description		Manufacture	Part Number	Comment			
B1	SMD balun	2.45 GHz	Wuerth	748421245	2.45 GH	2.45 GHz Balun		
B1 (alternatively)	SMD balun / filter	2.45 GHz	Johanson Technology	2450FB15L0001	2.45 GH	2.45 GHz Balun / Filter		
CB1 CB3	LDO VREG bypass capacitor AVX 0603YD105KAT2A		X5R	10%	16V			
CB2 CB4	Power Supply decoupling	- 1 μF	Murata	GRM188R61C105KA12D	(0603)	10 /6	167	
CX1, CX2	Crystal load capacitor	12 pF	AVX Murata	06035A120JA GRP1886C1H120JA01	COG (0603)	5%		
			Epcos	B37930	COG 5%			
C1, C2	RF coupling capacitor	22 pF	Epcos AVX	B37920 06035A220JAT2A	(0402 or 0603)		50V	
C3	CLKM low-pass	2.2 pF	AVX Murata	06035A229DA GRP1886C1H2R0DA01	COG (0603) ±0.5 pF			
	filter capacitor		Wurata	GHP 1000C 1H2HUDAU1	Designed for f _{CLKM} =1 MHz			
C4 (optional) RF matching		0.47 pF			Depend impleme	s on final P entation	СВ	
R1	R1 CLKM low-pass filter resistor				Designe	d for f _{CLKM} =	=1 MHz	
XTAL	Crystal	CX-4025 16 MHz SX-4025 16 MHz	ACAL Taitjen Siward	XWBBPL-F-1 A207-011				

Note: Please note that pins DIG1...4 are connected to the ground in the Basic Application Schematic, refer to Figure 5-1 on page 12. Special programming of these pins require a different schematic, refer to "Extended Feature Set Application Schematic" on page 14.



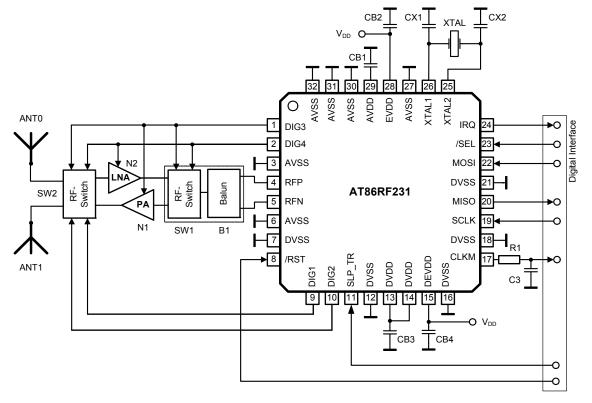
5.2 Extended Feature Set Application Schematic

The AT86RF231 supports additional features like:

•	Security Module (AES)		see Section 11.1
•	High Data Rate Mode		see Section 11.3
•	Antenna Diversity	uses pins DIG1/2	see Section 11.4
•	RX/TX indicator	uses pins DIG3/4	see Section 11.5
•	RX Frame Time Stamp	uses pin DIG2	see Section 11.6

An extended feature set application schematic illustrating the use of the AT86RF231 Extended Feature Set, see Section 11. "AT86RF231 Extended Feature Set" on page 128, is shown in Figure 5-2 on page 14. Although this example shows all additional hardware features combined, it is possible to use all features separately or in various combinations.

Figure 5-2. Extended Feature Application Schematic



In this example, a balun (B1) transforms the differential RF signal at the radio transceiver RF pins (RFP/RFN) to a single ended RF signal, similar to the Basic Application Schematic; refer to Figure 5-1 on page 12. The RF-Switches (SW1, SW2) separate between receive and transmit path in an external RF front-end.

These switches are controlled by the RX/TX Indicator, represented by the differential pin pair DIG3/DIG4, refer to Section 11.5 "RX/TX Indicator" on page 147.

During receive the radio transceiver searches for the most reliable RF signal path using the Antenna Diversity algorithm. One antenna is selected (SW2) by the Antenna Diversity RF switch



control pins DIG1/DIG2, the RF signal is amplified by an optional low-noise amplifier (N2) and fed to the radio transceiver using the second RX/TX switch (SW1).

During transmit the AT86RF231 TX signal is amplified using an external PA (N1) and fed to the antennas via an RF switch (SW2). In this example RF switch SW2 further supports Antenna Diversity controlled by the differential pin pair DIG1/DIG2.

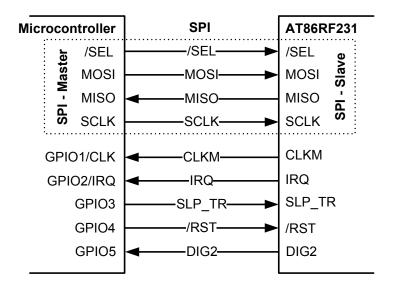
The security engine (AES) and High Data Rate Modes do not require specific circuitry to operate. The security engine (AES) has to be configured in advance, for details refer to Section 11.1 "Security Module (AES)" on page 128. The High Data Rate Modes are enabled by register bits OQPSK_DATA_RATE (register 0x0C, TRX_CTRL_2), for details refer to Section 11.3 "High Data Rate Modes" on page 137.



6. Microcontroller Interface

This section describes the AT86RF231 to microcontroller interface. The interface comprises a slave SPI and additional control signals; see Figure 6-1 on page 16. The SPI timing and protocol are described below.

Figure 6-1. Microcontroller to AT86RF231 Interface



Microcontrollers with a master SPI such as Atmel's AVR family interface directly to the AT86RF231. The SPI is used for register, Frame Buffer, SRAM and AES access. The additional control signals are connected to the GPIO/IRQ interface of the microcontroller.

Table 6-1 on page 16 introduces the radio transceiver I/O signals and their functionality.

 Table 6-1.
 Signal Description of Microcontroller Interface

Signal	Description						
/SEL	SPI select signal, active low						
MOSI	SPI data (master output slave input) signal						
MISO	SPI data (master input slave output) signal						
SCLK	SPI clock signal						
CLKM	Clock output, refer to Section 9.6.4 usable as: -microcontroller clock source -high precision timing reference -MAC timer reference						
IRQ	Interrupt request signal, further used as: -Frame Buffer Empty Indicator, refer to Section 11.7						



SLP_TR

Multipurpose control signal (functionality is state dependent, see Section 6.5):

-Sleep/Wakeup enable/disable SLEEP state

-TX start BUSY_TX_(ARET) state

-disable/enable CLKM RX_(AACK)_ON state

Table 6-1. Signal Description of Microcontroller Interface (Continued)

AT86RF231 reset signal, active low

6.1 SPI Timing Description

/RST

DIG2

Pin 17 (CLKM) can be used as a microcontroller master clock source. If the microcontroller derives the SPI master clock (SCLK) directly from CLKM, the SPI operates in synchronous mode, otherwise in asynchronous mode.

Optional, IRQ_2 (RX_START) for RX Frame Time Stamping, see Section 11.6

In synchronous mode, the maximum SCLK frequency is 8 MHz.

In asynchronous mode, the maximum SCLK frequency is limited to 7.5 MHz. The signal at pin CLKM is not required to derive SCLK and may be disabled to reduce power consumption and spurious emissions.

Figure 6-2 on page 17 and Figure 6-3 on page 17 illustrate the SPI timing and introduces its parameters. The corresponding timing parameter definitions t_1 - t_9 are defined in Section 12.4 "Digital Interface Timing Characteristics" on page 157.

Figure 6-2. SPI Timing, Global Map and Definition of Timing Parameters t₅, t₆, t₈ and t₉

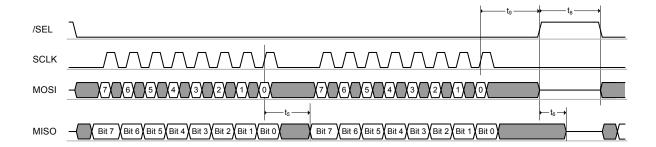
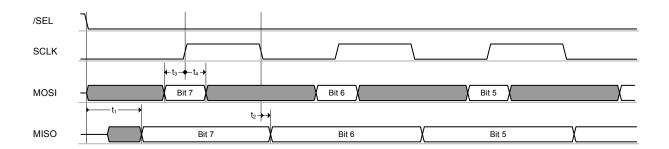


Figure 6-3. SPI Timing, Detailed Drawing of Timing Parameter t₁ to t₄





The SPI is based on a byte-oriented protocol and is always a bidirectional communication between master and slave. The SPI master starts the transfer by asserting /SEL = L. Then the master generates eight SPI clock cycles to transfer one byte to the radio transceiver (via MOSI). At the same time, the slave transmits one byte to the master (via MISO). When the master wants to receive one byte of data from the slave it must also transmit one byte to the slave. All bytes are transferred with MSB first. An SPI transaction is finished by releasing /SEL = H.

An SPI register access consists of two bytes, a Frame Buffer or SRAM access of at least two or more bytes as described in Section 6.2 "SPI Protocol" on page 19.

/SEL = L enables the MISO output driver of the AT86RF231. The MSB of MISO is valid after t1 (see Section 12.4 "Digital Interface Timing Characteristics" on page 157 parameter 12.4.3) and is updated at each falling edge of SCLK. If the driver is disabled, there is no internal pull-up circuitry connected to it. Driving the appropriate signal level must be ensured by the master device or an external pull-up resistor. Note, when both /SEL and /RST are active, the MISO output driver is also enabled.

Referring to Figure 6-2 on page 17 and Figure 6-3 on page 17 MOSI is sampled at the rising edge of the SCLK signal and the output is set at the falling edge of SCLK. The signal must be stable before and after the rising edge of SCLK as specified by t_3 and t_4 , refer to Section 12.4 "Digital Interface Timing Characteristics" on page 157 parameters 12.4.5 and 12.4.6.

This SPI operational mode is commonly known as "SPI mode 0".



6.2 SPI Protocol

Each SPI sequence starts with transferring a command byte from the SPI master via MOSI (see Table 6-2 on page 19) with MSB first. This command byte defines the SPI access mode and additional mode-dependent information.

Table 6-2. SPI Command Byte definition

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Access Mode	Access Type
1	0		R	egister ad	ddress [5:	:0]		Degister seeses	Read access
1	1		Register address [5:0]					Register access	Write access
0	0	1	Reserved					Frame Buffer access	Read access
0	1	1	Reserved					Frame Buller access	Write access
0	0	0	Reserved				CDAM	Read access	
0	1	0		Reserved				SRAM access	Write access

Each SPI transfer returns bytes back to the SPI master on MISO. The content of the first byte (see value "PHY_STATUS" in Figure 6-4 on page 19 to Figure 6-14 on page 23) is set to zero after reset. To transfer status information of the radio transceiver to the microcontroller, the content of the first byte can be configured with register bits SPI_CMD_MODE (register 0x04, TRX_CTRL_1). For details, refer to Section 6.3.1 "Register Description - SPI Control" on page 24.

In Figure 6-4 on page 19 to Figure 6-14 on page 23 and the following chapters logic values stated with XX on MOSI are ignored by the radio transceiver, but need to have a valid logic level. Return values on MISO stated as XX shall be ignored by the microcontroller.

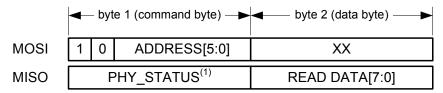
The different access modes are described within the following sections.

6.2.1 Register Access Mode

A register access mode is a two-byte read/write operation initiated by /SEL = L. The first transferred byte on MOSI is the command byte including an identifier bit (bit7 = 1), a read/write select bit (bit 6), and a 6-bit register address.

On read access, the content of the selected register address is returned in the second byte on MISO (see Figure 6-4 on page 19).

Figure 6-4. Packet Structure - Register Read Access

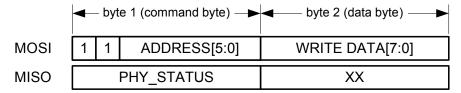


Note: 1. Each SPI access can be configured to return radio controller status information (PHY_STATUS) on MISO, for details refer to Section 6.3 "Radio Transceiver Status information" on page 24.

On write access, the second byte transferred on MOSI contains the write data to the selected address (see Figure 6-5 on page 20).



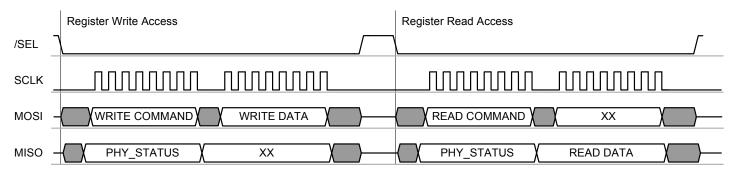
Figure 6-5. Packet Structure - Register Write Access



Each register access must be terminated by setting /SEL = H.

Figure 6-6 on page 20 illustrates a typical SPI sequence for a register access sequence for write and read respectively.

Figure 6-6. Example SPI Sequence - Register Access Mode



6.2.2 Frame Buffer Access Mode

The 128-byte Frame Buffer can hold the PHY service data unit (PSDU) data of one IEEE 802.15.4 compliant RX or one TX frame of maximum length at a time. A detailed description of the Frame Buffer can be found in Section 9.3 "Frame Buffer" on page 107. An introduction to the IEEE 802.15.4 frame format can be found in Section 8.1 "Introduction - IEEE 802.15.4 - 2006 Frame Format" on page 79.

Frame Buffer read and write accesses are used to read or write frame data (PSDU and additional information) from or to the Frame Buffer. Each access starts with /SEL = L followed by a command byte on MOSI. If this byte indicates a frame read or write access, the next byte PHR[7:0] indicates the frame length followed by the PSDU data, see Figure 6-7 on page 20 and Figure 6-8 on page 21.

On Frame Buffer read access, PHY header (PHR) and PSDU are transferred via MISO starting with the second byte. After the PSDU data, one more byte is transferred containing the link quality indication (LQI) value of the received frame, for details refer to Section 8.6 "Link Quality Indication (LQI)" on page 99. Figure 6-7 on page 20 illustrates the packet structure of a Frame Buffer read access.

Figure 6-7. Packet Structure - Frame Read Access

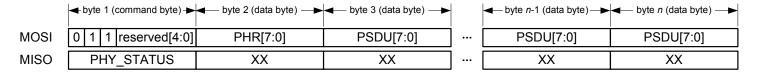
	→ byte 1 (command byte) →	→ byte 2 (data byte) → →	→ byte 3 (data byte) → →		→ byte <i>n-1</i> (data byte) →	→ byte n (data byte) →
MOSI	0 0 1 reserved[4:0]	XX	XX	•••	XX	XX
MISO	PHY_STATUS	PHR[7:0]	PSDU[7:0]	•••	PSDU[7:0]	LQI[7:0]



Note, the Frame Buffer read access can be terminated at any time without any consequences by setting /SEL = H, e.g. after reading the PHR byte only.

On Frame Buffer write access the second byte transferred on MOSI contains the frame length (PHR field) followed by the payload data (PSDU) as shown by Figure 6-8 on page 21.

Figure 6-8. Packet Structure - Frame Write Access



The number of bytes n for one frame access is calculated as follows:

• **Read Access:** $n = 3 + frame_length$

[PHY_STATUS, PHR byte, PSDU data, and LQI byte]

• Write Access: n = 2 + frame length

[command byte, PHR byte, and PSDU data]

The maximum value of frame_length is 127 bytes. That means that $n \le 130$ for Frame Buffer read and $n \le 129$ for Frame Buffer write accesses.

Each read or write of a data byte increments automatically the address counter of the Frame Buffer until the access is terminated by setting /SEL = H. A Frame Buffer read access may be terminated (/SEL = H) at any time without affecting the Frame Buffer content. Another Frame Buffer read operation starts again at the PHR field.

The content of the Frame Buffer is only overwritten by a new received frame or a Frame Buffer write access.

Figure 6-9 on page 21 and Figure 6-10 on page 22 illustrate an example SPI sequence of a Frame Buffer access to read and write a frame with 4-byte PSDU respectively.

Figure 6-9. Example SPI Sequence - Frame Buffer Read of a Frame with 4-byte PSDU

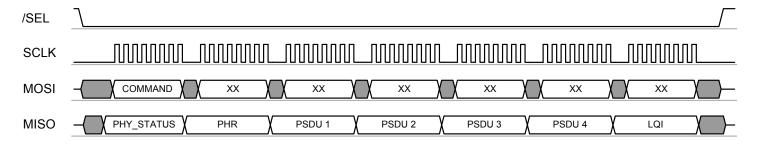
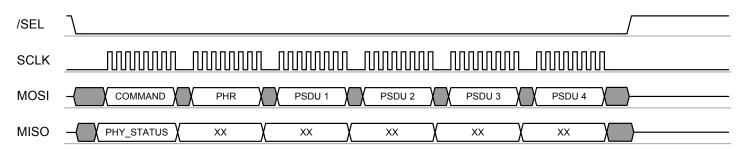




Figure 6-10. Example SPI Sequence - Frame Buffer Write of a Frame with 4 byte PSDU



Access violations during a Frame Buffer read or write access are indicated by interrupt IRQ_6 (TRX UR). For further details, refer to Section 9.3 "Frame Buffer" on page 107.

Notes

- The Frame Buffer is shared between RX and TX; therefore, the frame data are overwritten by new incoming frames. If the TX frame data are to be retransmitted, it must be ensured that no frame was received in the meanwhile.
- To avoid overwriting during receive *Dynamic Frame Buffer Protection* can be enabled, refer to Section 11.8 "Dynamic Frame Buffer Protection" on page 154.
- It is not possible to retransmit received frames without a Frame Buffer read and write access cycle.
- For exceptions, e.g. receiving acknowledgement frames in Extended Operating Mode (TX_ARET) refer to Section 7.2.4 "TX_ARET_ON - Transmit with Automatic Retry and CSMA-CA Retry" on page 64.

6.2.3 SRAM Access Mode

The SRAM access mode allows accessing dedicated bytes within the Frame Buffer. This may reduce the SPI traffic.

The SRAM access mode is useful, for instance, if a transmit frame is already stored in the Frame Buffer and dedicated bytes (e.g. sequence number, address field) need to be replaced before retransmitting the frame. Furthermore, it can be used to access only the LQI value after frame reception. A detailed description of the user accessible frame content can be found in Section 9.3 "Frame Buffer" on page 107.

Each SRAM access starts with /SEL = L. The first transferred byte on MOSI shall be the command byte and must indicate an SRAM access mode according to the definition in Table 6-2 on page 19. The following byte indicates the start address of the write or read access. The address space is 0x00 to 0x7F for radio transceiver receive or transmit operations.

On SRAM read access, one or more bytes of read data are transferred on MISO starting with the third byte of the access sequence (see Figure 6-11 on page 22).

Figure 6-11. Packet Structure - SRAM Read Access

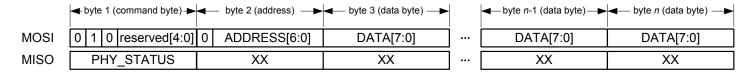
	→ byte 1 (command byte) →	→ byte 2 (address) → → → → → → → → → → → → → → → → → →	→ byte 3 (data byte) → → → → → → → → → → → → → → → → → → →		→ byte <i>n</i> -1 (data byte) →	→ byte <i>n</i> (data byte) →
MOSI	0 0 0 reserved[4:0]	0 ADDRESS[6:0]	XX	•••	XX	XX
MISO	PHY_STATUS	XX	DATA[7:0]	•••	DATA[7:0]	DATA[7:0]



On SRAM write access, one or more bytes of write data are transferred on MOSI starting with the third byte of the access sequence (see Figure 6-12 on page 23).

On SRAM read or write accesses do not attempt to read or write bytes beyond the SRAM buffer size.

Figure 6-12. Packet Structure - SRAM Write Access



As long as /SEL = L, every subsequent byte read or byte write increments the address counter of the Frame Buffer until the SRAM access is terminated by /SEL = H.

Figure 6-13 on page 23 and Figure 6-14 on page 23 illustrate an example SPI sequence of a SRAM access to read and write a data package of 5-byte length respectively.

Figure 6-13. Example SPI Sequence - SRAM Read Access of a 5 byte Data Package

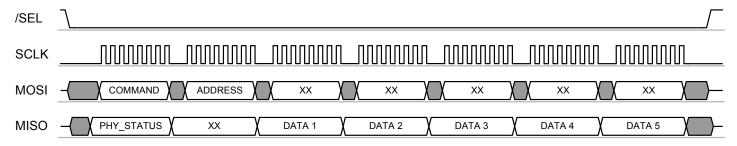
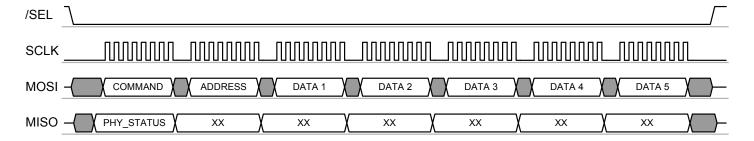


Figure 6-14. Example SPI Sequence - SRAM Write Access of a 5 byte Data Package



Notes

- The SRAM access mode is not intended to be used as an alternative to the Frame Buffer access modes (see Section 6.2.2 "Frame Buffer Access Mode" on page 20).
- If the SRAM access mode is used to read PSDU data, the Frame Buffer contains all PSDU data except the frame length byte (PHR). The frame length information can be accessed only using Frame Buffer access.
- Frame Buffer access violations are not indicated by a TRX_UR interrupt when using the SRAM access mode, for further details refer to Section 9.3.3 "Interrupt Handling" on page 109.



6.3 Radio Transceiver Status information

Each SPI access can be configured to return status information of the radio transceiver (PHY_STATUS) to the microcontroller using the first byte of the data transferred via MISO.

The content of the radio transceiver status information can be configured using register bits SPI_CMD_MODE (register 0x04, TRX_CTRL_1). After reset, the content on the first byte send on MISO to the microcontroller is set to 0x00.

6.3.1 Register Description - SPI Control

Register 0x04 (TRX_CTRL_1):

The TRX_CTRL_1 register is a multi purpose register to control various operating modes and settings of the radio transceiver.

Bit	/	6	5	4	3	2	1	0	_
+0x04	PA_EXT_EN	IRQ_2_EXT_EN	TX_AUTO_CRC_ON	RX_BL_CTRL	SPI_CME	D_MODE	IRQ_MASK_MODE	IRQ_POLARITY	TRX_CTRL_1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	,
Initial Value	0	0	1	0	0	0	0	0	

• Bit 7 - PA_EXT_EN

Refer to Section 11.5 "RX/TX Indicator" on page 147.

• Bit 6 - IRQ 2 EXT EN

Refer to Section 11.6 "RX Frame Time Stamping" on page 150.

• Bit 5 - TX_AUTO_CRC_ON

Refer to Section 8.2 "Frame Check Sequence (FCS)" on page 85.

• Bit 4 - RX_BL_CTRL

Refer to Section 11.7 "Frame Buffer Empty Indicator" on page 152.

• Bit [3:2] - SPI CMD MODE

Each SPI transfer returns bytes back to the SPI master. The content of the first byte can be configured using register bits SPI_CMD_MODE. The transfer of the following status information can be configured as follows:

Table 6-3. Radio Transceiver Status Information - PHY STATUS

Register Bit	Value	Description
SPI_CMD_MODE	<u>0</u>	default (empty, all bits 0x00)
	1	monitor TRX_STATUS register; see Section 7.1.5
	2	monitor PHY_RSSI register; see Section 8.3
	3	monitor IRQ_STATUS register; see Section 6.6

• Bit 1 - IRQ_MASK_MODE

Refer to Section 6.6 "Interrupt Logic" on page 29.

• Bit 0 - IRQ POLARITY

Refer to Section 6.6 "Interrupt Logic" on page 29.



6.4 Radio Transceiver Identification

The AT86RF231 can be identified by four registers. One register contains a unique part number and one register the corresponding version number. Two additional registers contain the JEDEC manufacture ID.

6.4.1 Register Description - AT86RF231 Identification

Register 0x1C (PART_NUM):

Bit	7	6	5	4	3	2	1	0	_
+0x1C				PART_N	UM[7:0]				PART_NUM
Read/Write	R	R	R	R	R	R	R	R	•
Reset Value	0	0	0	0	0	0	1	1	

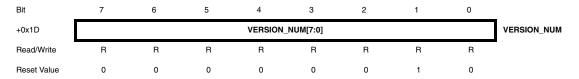
• Bit [7:0] - PART_NUM

This register contains the radio transceiver part number.

Table 6-4. Radio Transceiver Part Number

Register Bit	Value	Description
PART_NUM	<u>3</u>	AT86RF231 part number

Register 0x1D (VERSION_NUM):



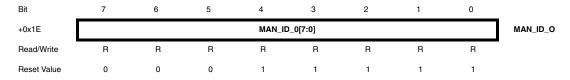
• Bit [7:0] - VERSION_NUM

This register contains the radio transceiver version number.

Table 6-5. Radio Transceiver Version Number

Register Bit	Value	Description
VERSION_NUM	<u>2</u>	Revision A

Register 0x1E (MAN_ID_0):



• Bit [7:0] - MAN_ID_0

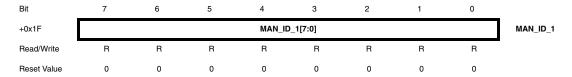
Bits [7:0] of the 32-bit JEDEC manufacturer ID are stored in register bits MAN_ID_0. Bits [15:8] are stored in register 0x1F (MAN_ID_1). The highest 16 bits of the ID are not stored in registers.



Table 6-6. JEDEC Manufacturer ID - Bits [7:0]

Register Bit	Value	Description
MAN_ID_0	<u>0x1F</u>	Atmel JEDEC manufacturer ID,
		Bits [7:0] of 32 bit manufacturer ID: 00 00 00 1F

Register 0x1F (MAN_ID_1):



• Bit [7:0] - MAN_ID_1

Bits [15:8] of the 32-bit JEDEC manufacturer ID are stored in register bits MAN_ID_1. Bits [7:0] are stored in register 0x1E (MAN_ID_0). The higher 16 bits of the ID are not stored in registers.

Table 6-7. JEDEC Manufacturer ID - Bits [15:8]

Register Bit	Value	Description
MAN_ID_1	<u>0x00</u>	Atmel JEDEC manufacturer ID, Bits [15:8] of 32 bit manufacturer ID: 00 00 00 1F



6.5 Sleep/Wake-up and Transmit Signal (SLP_TR)

Pin 11 (SLP_TR) is a multi-functional pin. Its function relates to the current state of the AT86RF231 and is summarized in Table 6-8 on page 27. The radio transceiver states are explained in detail Section 7. "Operating Modes" on page 33.

Table 6-8. SLP TR Multi-functional Pin

Transceiver Status	Function	Transition	Description
PLL_ON	TX start	L⇒H	Starts frame transmission
TX_ARET_ON	TX start	L⇒H	Starts TX_ARET transaction
TRX_OFF	Sleep	L⇒H	Takes the radio transceiver into SLEEP state, CLKM disabled
SLEEP	Wakeup	H⇒L	Takes the radio transceiver back into TRX_OFF state, level sensitive
RX_ON	Disable CLKM	L⇒H	Takes the radio transceiver into RX_ON_NOCLK state and disables CLKM
RX_ON_NOCLK	Enable CLKM	H⇒L	Takes the radio transceiver into RX_ON state and enables CLKM
RX_AACK_ON	Disable CLKM	L⇒H	Takes the radio transceiver into RX_AACK_ON_NOCLK state and disables CLKM
RX_AACK_ON_NOCLK	Enable CLKM	H⇒L	Takes the radio transceiver into RX_AACK_ON state and enables CLKM

In states PLL_ON and TX_ARET_ON, pin SLP_TR is used as trigger input to initiate a TX transaction. Here pin SLP_TR is sensitive on rising edge only.

After initiating a state change by a rising edge at pin SLP_TR in radio transceiver states TRX_OFF, RX_ON or RX_AACK_ON, the radio transceiver remains in the new state as long as the pin is logical high and returns to the preceding state with the falling edge.

SLEEP state

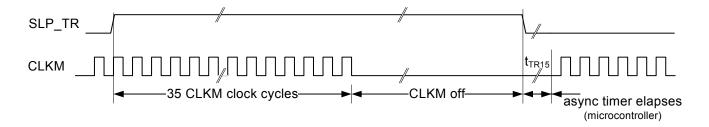
The SLEEP state is used when radio transceiver functionality is not required, and thus the AT86RF231 can be powered down to reduce the overall power consumption.

A power-down scenario is shown in Figure 6-15 on page 28. When the radio transceiver is in TRX_OFF state the microcontroller forces the AT86RF231 to SLEEP by setting SLP_TR = H. If pin 17 (CLKM) provides a clock to the microcontroller this clock is switched off after 35 clock cycles. This enables a microcontroller in a synchronous system to complete its power-down routine and prevent deadlock situations. The AT86RF231 awakes when the microcontroller releases pin SLP_TR. This concept provides the lowest possible power consumption.

The CLKM clock frequency settings for 250 kHz and 62.5 kHz are not intended to directly clock the microcontroller. When using these clock rates, CLKM is turned off immediately when entering SLEEP state.



Figure 6-15. Sleep and Wake-up Initiated by Asynchronous Microcontroller Timer



Note: Timing figure t_{TR15} refer to section Table 7-1 on page 42.

RX_ON and RX_AACK_ON states

For synchronous systems, where CLKM is used as a microcontroller clock source and the SPI master clock (SCLK) is directly derived from CLKM, the AT86RF231 supports an additional power-down mode for receive operating states (RX_ON and RX_AACK_ON).

If an incoming frame is expected and no other applications are running on the microcontroller, it can be powered down without missing incoming frames.

This can be achieved by a rising edge on pin SLP_TR that turns off the CLKM. Then the radio transceiver state changes from RX_ON or RX_AACK_ON (Extended Operating Mode) to RX_ON_NOCLK or RX_AACK_ON_NOCLK respectively.

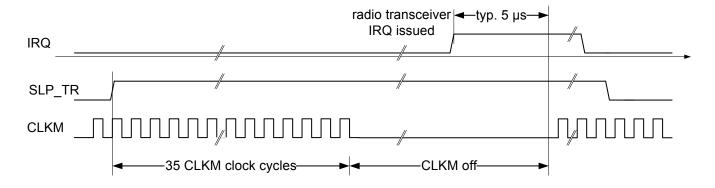
In case that a frame is received (e.g. indicated by an IRQ_2 (RX_START) interrupt) the clock output CLKM is automatically switched on again.

This scenario is shown in Figure 6-16 on page 28. In RX_ON state, the clock at pin 17 (CLKM) is switched off after 35 clock cycles when setting the pin SLP_TR = H.

The CLKM clock frequency settings for 250 kHz and 62.5 kHz are not intended to directly clock the microcontroller. When using these clock rates, CLKM is turned off immediately when entering RX_ON_NOCLK and RX_AACK_ON_NOCLK respectively.

In states RX_(AACK)_ON_NOCLK and RX_(AACK)_ON, the radio transceiver current consumptions are equivalent. However, the RX_(AACK)_ON_NOCLK current consumption is reduced by the current required for driving pin 17 (CLKM).

Figure 6-16. Wake-Up Initiated by Radio Transceiver Interrupt





6.6 Interrupt Logic

6.6.1 Overview

The AT86RF231 differentiates between nine interrupt events (eight physical interrupt registers, one shared by two functions). Each interrupt is enabled by setting the corresponding bit in the interrupt mask register 0x0E (IRQ_MASK). Internally, each pending interrupt is stored in a separate bit of the interrupt status register. All interrupt events are OR-combined to a single external interrupt signal (IRQ, pin 24). If an interrupt is issued (pin IRQ = H), the microcontroller shall read the interrupt status register 0x0F (IRQ_STATUS) to determine the source of the interrupt. A read access to this register clears the interrupt status register and thus the IRQ pin, too.

Interrupts are not cleared automatically when the event that caused them vanishes. Exceptions are IRQ_0 (PLL_LOCK) and IRQ_1 (PLL_UNLOCK) because the occurrence of one clears the other.

The supported interrupts for the Basic Operating Mode are summarized in Table 6-9 on page 29.

 Table 6-9.
 Interrupt Description in Basic Operating Mode

IRQ Name	Description	Section
IRQ_7 (BAT_LOW)	Indicates a supply voltage below the programmed threshold.	9.5.4
IRQ_6 (TRX_UR)	Indicates a Frame Buffer access violation.	9.3.3
IRQ_5 (AMI)	Indicates address matching.	7.2.3.5
IRQ_4 (CCA_ED_DONE)	Multi-functional interrupt: 1. AWAKE_END: • Indicates radio transceiver reached TRX_OFF state after P_ON, RESET, or SLEEP states. 2. CCA_ED_DONE: • Indicates the end of a CCA or ED measurement.	7.1.2.3 8.4.4 8.5.4
IRQ_3 (TRX_END)	RX: Indicates the completion of a frame reception. TX: Indicates the completion of a frame transmission.	7.1.3 7.1.3
IRQ_2 (RX_START)	Indicates the start of a PSDU reception. The TRX_STATE changes to BUSY_RX, the PHR is valid to read from Frame Buffer.	7.1.3
IRQ_1 (PLL_UNLOCK)	Indicates PLL unlock. If the radio transceiver is BUSY_TX / BUSY_TX_ARET state, the PA is turned off immediately.	9.7.5
IRQ_0 (PLL_LOCK)	Indicates PLL lock.	9.7.5

The interrupt IRQ_4 has two meanings, depending on the current radio transceiver state, refer to register 0x01 (TRX_STATUS).

After P_ON, SLEEP, or RESET, the radio transceiver issues an interrupt IRQ_4 (AWAKE_END) when it enters state TRX_OFF.

The second meaning is only valid for receive states. If the microcontroller initiates an energy-detect (ED) or clear-channel-assessment (CCA) measurement, the completion of the measurement is indicated by interrupt IRQ_4 (CCA_ED_DONE), refer to Section 8.4.4 "Interrupt Handling" on page 92 and Section 8.5.4 "Interrupt Handling" on page 95 for details.

After P_ON or RESET all interrupts are disabled. During radio transceiver initialization it is recommended to enable IRQ_4 (AWAKE_END) to be notified once the TRX_OFF state is entered.



Note that AWAKE_END interrupt can usually not be seen when the transceiver enters TRX_OFF state after RESET, because register 0x0E (IRQ_MASK) is reset to mask all interrupts. In this case, state TRX_OFF is normally entered before the microcontroller could modify the register.

The interrupt handling in Extended Operating Mode is described in Section 7.2.5 "Interrupt Handling" on page 67.

If register bit IRQ_MASK_MODE (register 0x04, TRX_CTRL_1) is set, an interrupt event can be read from IRQ_STATUS register even if the interrupt itself is masked. However, in that case no timing information for this interrupt is provided.

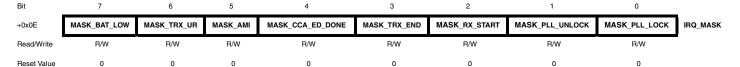
The IRQ pin polarity can be configured with register bit IRQ_POLARITY (register 0x04, TRX_CTRL_1). The default behavior is active high, which means that pin IRQ = H issues an interrupt request.

If "Frame Buffer Empty Indicator" is enabled during Frame Buffer read access the IRQ pin has an alternative functionality, refer to Section 11.7 "Frame Buffer Empty Indicator" on page 152 for details.

6.6.2 Register Description

Register 0x0E (IRQ_MASK):

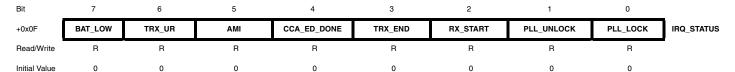
The IRQ_MASK register is used to enable or disable individual interrupts. An interrupt is enabled if the corresponding bit is set to 1. All interrupts are disabled after power up sequence (P_ON state) or reset (RESET state).



If an interrupt is enabled it is recommended to read the interrupt status register 0x0F (IRQ_STATUS) first to clear the history.

Register 0x0F (IRQ STATUS):

The IRQ_STATUS register contains the status of the pending interrupt requests.



By reading the register after an interrupt is signaled at pin 24 (IRQ) the source of the issued interrupt can be identified. A read access to this register resets all interrupt bits, and so clears the IRQ_STATUS register.

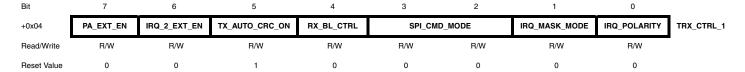
If register bit IRQ_MASK_MODE (register 0x04, TRX_CTRL_1) is set, an interrupt event can be read from IRQ_STATUS register even if the interrupt itself is masked. However in that case no timing information for this interrupt is provided.

If register bit IRQ_MASK_MODE is set, it is recommended to read the interrupt status register 0x0F (IRQ_STATUS) first to clear the history.



Register 0x04 (TRX_CTRL_1):

The TRX_CTRL_1 register is a multi purpose register to control various operating modes and settings of the radio transceiver.



• Bit 7 - PA EXT EN

Refer to Section 11.5 "RX/TX Indicator" on page 147.

Bit 6 - IRQ_2_EXT_EN

The timing of a received frame can be determined by a separate pin. If register bit IRQ_2_EXT_EN is set to 1, the reception of a PHR is directly issued on pin 10 (DIG2), similar to interrupt IRQ_2 (RX_START). Note that this pin is also active even if the corresponding interrupt event IRQ_2 (RX_START) mask bit in register 0x0E (IRQ_MASK) is set to 0. The pin remains at high level until the end of the frame receive procedure.

For further details refer to Section 11.6 "RX Frame Time Stamping" on page 150.

Bit 5 - TX AUTO CRC ON

Refer to Section 8.2 "Frame Check Sequence (FCS)" on page 85.

• Bit 4 - RX BL CTRL

Refer to Section 11.7 "Frame Buffer Empty Indicator" on page 152.

• Bit [3:2] - SPI_CMD_MODE

Refer to Section 6.3 "Radio Transceiver Status information" on page 24.

• Bit 1 - IRQ_MASK_MODE

The AT86RF231 supports polling of interrupt events. Interrupt polling can be enabled by register bit IRQ_MASK_MODE. Even if an interrupt request is masked by the corresponding bit in register 0x0E (IRQ_MASK), the event is indicated in register 0x0F (IRQ_STATUS).

 Table 6-10.
 Interrupt Polling Configuration

Register Bit	Value	Description
IRQ_MASK_MODE	<u>0</u>	Interrupt polling disabled
	1	Interrupt polling enabled



• Bit 0 - IRQ_POLARITY

The default polarity of the IRQ pin is active high. The polarity can be configured to active low via register bit IRQ_POLARITY, see Table 6-11 on page 32.

Table 6-11. Configuration of Pin 24 (IRQ)

Register Bit	Value	Description
IRQ_POLARITY	<u>0</u>	pin IRQ high active
	1	pin IRQ low active

This setting does not affect the polarity of the Frame Buffer Empty Indicator, refer to Section 11.7 "Frame Buffer Empty Indicator" on page 152. The Frame Buffer Empty Indicator is always active high.

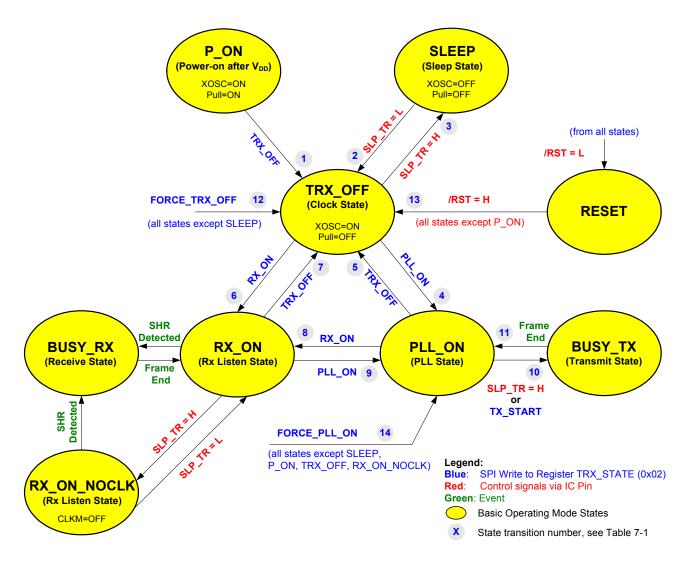


7. Operating Modes

7.1 Basic Operating Mode

This section summarizes all states to provide the basic functionality of the AT86RF231, such as receiving and transmitting frames, the power up sequence and sleep. The Basic Operating Mode is designed for IEEE 802.15.4 and ISM applications; the corresponding radio transceiver states are shown in Figure 7.1 on page 33.

Figure 7-1. Basic Operating Mode State Diagram (for timing refer to Table 7-1 on page 42)



7.1.1 State Control

The radio transceiver states are controlled either by writing commands to register bits TRX_CMD (register 0x02, TRX_STATE), or directly by two signal pins: pin 11 (SLP_TR) and



pin 8 (/RST). A successful state change can be verified by reading the radio transceiver status from register 0x01 (TRX_STATUS).

If TRX_STATUS = 0x1F (STATE_TRANSITION_IN_PROGRESS) the AT86RF231 is on a state transition. Do not try to initiate a further state change while the radio transceiver is in STATE_TRANSITION_IN_PROGRESS.

Pin SLP_TR is a multifunctional pin, refer to Section 6.5 "Sleep/Wake-up and Transmit Signal (SLP_TR)" on page 27. Dependent on the radio transceiver state, a rising edge of pin SLP_TR causes the following state transitions:

TRX_OFF ⇒ SLEEP (level sensitive)
 RX_ON ⇒ RX_ON_NOCLK (level sensitive)
 PLL_ON ⇒ BUSY TX

Whereas the falling edge of pin SLP_TR causes the following state transitions:

• SLEEP ⇒ TRX_OFF (level sensitive)
• RX_ON_NOCLK ⇒ RX_ON (level sensitive)

Pin 8 (/RST) causes a reset of all registers (register bits CLKM_CTRL are shadowed, for details refer to Section 9.6.4 "Master Clock Signal Output (CLKM)" on page 117) and forces the radio transceiver into TRX_OFF state. However, if the device was in P_ON state it remains in the P_ON state.

For all states except SLEEP, the state change commands FORCE_TRX_OFF or TRX_OFF lead to a transition into TRX_OFF state. If the radio transceiver is in active receive or transmit states (BUSY_*), the command FORCE_TRX_OFF interrupts these active processes, and forces an immediate transition to TRX_OFF. In contrast a TRX_OFF command is stored until an active state (receiving or transmitting) has been finished. After that the transition to TRX_OFF is performed.

For a fast transition from receive or active transmit states to PLL_ON state the command FORCE_PLL_ON is provided. In contrast to FORCE_TRX_OFF this command does not disable the PLL and the analog voltage regulator AVREG. It is not available in states SLEEP, P_ON, RESET, TRX_OFF, and all *_NOCLK states.

The completion of each requested state change shall always be confirmed by reading the register bits TRX_STATUS (register 0x01, TRX_STATUS).

7.1.2 Basic Operating Mode Description

7.1.2.1 P_ON - Power-On after V_{DD}

When the external supply voltage (V_{DD}) is firstly applied to the AT86RF231, the radio transceiver goes into the P_ON state performing an on-chip reset. The crystal oscillator is activated and the default 1 MHz master clock is provided at pin 17 (CLKM) after the crystal oscillator has stabilized. CLKM can be used as a clock source to the microcontroller. The SPI interface and digital voltage regulator are enabled.

The on-chip power-on-reset sets all registers to their default values. A dedicated reset signal from the microcontroller at pin 8 (/RST) is not necessary, but recommended for hardware / software synchronization reasons.



All digital inputs are pulled-up or pulled-down during P_ON state, refer to Section 1.3.2 "Pull-Up and Pull-Down Configuration" on page 7. This is necessary to support microcontrollers where GPIO signals are floating after power on or reset. The input pull-up and pull-down circuitry is disabled when the radio transceiver leaves the P_ON state. Output pins DIG1/DIG2 are pulled-down to digital ground, whereas pins DIG3/DIG4 are pulled-down to analog ground, unless their configuration is changed.

Prior to leaving P_ON, the microcontroller must set the pins to the default operating values: SLP TR = L, /RST = H and /SEL = H.

All interrupts are disabled by default. Thus, interrupts for state transition control are to be enabled first, e.g. enable IRQ_4 (AWAKE_END) to indicate a state transition to TRX_OFF state or interrupt IRQ_0 (PLL_LOCK) to signal a locked PLL in PLL_ON state. In P_ON state a first access to the radio transceiver registers is possible after a default 1 MHz master clock is provided at pin 17 (CLKM), refer to Table 7-1 on page 42.

Once the supply voltage has stabilized and the crystal oscillator has settled (see Section 12.5 "General RF Specifications" on page 158, parameter 12.5.7), a valid SPI write access to register bits TRX_CMD (register 0x02, TRX_STATE) with the command TRX_OFF or FORCE_TRX_OFF initiate a state change from P_ON towards TRX_OFF state, which is then indicated by an AWAKE_END interrupt if enabled.

7.1.2.2 SLEEP - Sleep State

In SLEEP state, the entire radio transceiver is disabled. No circuitry is operating. The radio transceiver current consumption is reduced to leakage current only. This state can only be entered from state TRX_OFF, by setting the pin SLP_TR = H.

If CLKM is enabled, the SLEEP state is entered 35 CLKM cycles after the rising edge at pin 11 (SLP_TR). At that time CLKM is turned off. If the CLKM output is already turned off (bits CLKM_CTRL = 0 in register 0x03), the SLEEP state is entered immediately. At clock rates 250 kHz and 62.5 kHz, the main clock at pin 17 (CLKM) is turned off immediately.

Setting SLP_TR = L returns the radio transceiver to the TRX_OFF state. During SLEEP the register contents remains valid while the content of the Frame Buffer and the security engine (AES) are cleared.

/RST = L in SLEEP state returns the radio transceiver to TRX_OFF state and thereby sets all registers to their default values. Exceptions are register bits CLKM_CTRL (register 0x03, TRX_CTRL_0). These register bits require a specific treatment, for details see Section 9.6.4 "Master Clock Signal Output (CLKM)" on page 117.

7.1.2.3 TRX OFF - Clock State

In TRX_OFF the crystal oscillator is running and the master clock is available at pin 17 (CLKM) after the crystal oscillator has stabilized. The SPI interface and digital voltage regulator are enabled, thus the radio transceiver registers, the Frame Buffer and security engine (AES) are accessible (see Section 9.3 "Frame Buffer" on page 107 and Section 11.1 "Security Module (AES)" on page 128).

In contrast to P_ON state the pull-up and pull-down configuration is disabled.

Pin 11 (SLP_TR) and pin 8 (/RST) are available for state control. Note that the analog front-end is disabled during TRX_OFF.



Entering the TRX_OFF state from P_ON, SLEEP, or RESET state is indicated by interrupt IRQ_4 (AWAKE_END).

7.1.2.4 PLL ON - PLL State

Entering the PLL_ON state from TRX_OFF state enables the analog voltage regulator (AVREG) first. After the voltage regulator has been settled, the PLL frequency synthesizer is enabled. When the PLL has been settled at the receive frequency to a channel defined by register bits CHANNEL (register 0x08, PHY_CC_CCA), a successful PLL lock is indicated by issuing an interrupt IRQ_0 (PLL_LOCK).

If an RX_ON command is issued in PLL_ON state, the receiver is immediately enabled. If the PLL has not been settled before the state change nevertheless takes place. Even if the register bits TRX_STATUS (register 0x01, TRX_STATUS) indicates RX_ON, actual frame reception can only start once the PLL has locked.

The PLL_ON state corresponds to the TX_ON state in IEEE 802.15.4.

7.1.2.5 RX_ON and BUSY_RX - RX Listen and Receive State

In RX ON state the receiver blocks and the PLL frequency synthesizer are enabled.

The AT86RF231 receive mode is internally separated into RX_ON state and BUSY_RX state. There is no difference between these states with respect to the analog radio transceiver circuitry, which are always turned on. In both states the receiver and the PLL frequency synthesizer are enabled.

During RX_ON state the receiver listens for incoming frames. After detecting a valid synchronization header (SHR), the AT86RF231 automatically enters the BUSY_RX state. The reception of a valid PHY header (PHR) generates an IRQ_2 (RX_START) and receives and demodulates the PSDU data.

During PSDU reception the frame data are stored continuously in the Frame Buffer until the last byte was received. The completion of the frame reception is indicated by an interrupt IRQ_3 (TRX_END) and the radio transceiver reenters the state RX_ON. At the same time the register bit RX_CRC_VALID (register 0x06, PHY_RSSI) is updated with the result of the FCS check (see Section 8.2 "Frame Check Sequence (FCS)" on page 85).

Received frames are passed to the frame filtering unit, refer to Section 7.2.3.5 "Frame Filtering" on page 61. If the content of the MAC addressing fields (refer to IEEE 802.15.4-2006, Section 7.2.1) of a frame matches to the expected addresses, which is further dependent on the addressing mode, an address match interrupt IRQ_5 (AMI) is issued, refer to Section 6.6 "Interrupt Logic" on page 29. The expected address values are to be stored in registers 0x20 - 0x2B (Short address, PAN-ID and IEEE address). Frame filtering is available in Basic and Extended Operating Mode, refer to Section 7.2.3.5 "Frame Filtering" on page 61.

Leaving state RX_ON is only possible by writing a state change command to register bits TRX_CMD in register 0x02 (TRX_STATE).

7.1.2.6 RX_ON_NOCLK - RX Listen State without CLKM

If the radio transceiver is listening for an incoming frame and the microcontroller is not running an application, the microcontroller may be powered down to decrease the total system power consumption. This specific power-down scenario for systems running in clock synchronous mode (see Section 6. "Microcontroller Interface" on page 16), is supported by the AT86RF231 using the state RX_ON_NOCLK.



This state can only be entered by setting pin 11 (SLP_TR) = H while the radio transceiver is in the RX_ON state, refer to Section 7.1.2.5 "RX_ON and BUSY_RX - RX Listen and Receive State" on page 36. Pin 17 (CLKM) is disabled 35 clock cycles after the rising edge at the SLP_TR pin, see Figure 6-16 on page 28. This allows the microcontroller to complete its power-down sequence.

Note that for CLKM clock rates 250 kHz and 62.5 kHz the master clock signal CLKM is switched off immediately after rising edge of SLP_TR.

The reception of a frame shall be indicated to the microcontroller by an interrupt indicating the receive status. CLKM is turned on again, and the radio transceiver enters the BUSY_RX state (see Section 6.5 "Sleep/Wake-up and Transmit Signal (SLP_TR)" on page 27 and Figure 6-16 on page 28). Using this radio transceiver state it is essential to enable at least one interrupt indicating the reception status. Otherwise the reception of a frame does not activate CLKM and the microcontroller remains in its power-down mode.

After the receive transaction has been completed, the radio transceiver enters the RX_ON state. The radio transceiver only reenters the RX_ON_NOCLK state, when the next rising edge of pin SLP_TR pin occurs.

If the AT86RF231 is in the RX_ON_NOCLK state, and pin SLP_TR is reset to logic low, it enters the RX_ON state, and it starts to supply clock on the CLKM pin again.

In states RX_ON_NOCLK and RX_ON, the radio transceiver current consumptions are equivalent. However, the RX_ON_NOCLK current consumption is reduced by the current required for driving pin 17 (CLKM).

Note

• A reset in state RX_ON_NOCLK requires further to reset pin SLP_TR to logic low, otherwise the radio transceiver enters directly the SLEEP state.

7.1.2.7 BUSY_TX - Transmit State

A transmission can only be initiated in state PLL_ON. There are two ways to start a transmission:

- Rising edge of pin 11 (SLP_TR)
- TX_START command to register bits TRX_CMD (register 0x02, TRX_STATE).

Either of these causes the radio transceiver into the BUSY_TX state.

During the transition to BUSY_TX state, the PLL frequency shifts to the transmit frequency. The actual transmission of the first data chip of the SHR starts after 16 µs to allow PLL settling and PA ramp-up, see Figure 7-6 on page 41. After transmission of the SHR, the Frame Buffer content is transmitted. In case the PHR indicates a frame length of zero, the transmission is aborted.

After the frame transmission has completed, the AT86RF231 automatically turns off the power amplifier, generates an IRQ_3 (TRX_END) interrupt and returns into PLL_ON state.

7.1.2.8 RESET State

The RESET state is used to set back the state machine and to reset all registers of the AT86RF231 to their default values, exception are register bits CLKM_CTRL (register 0x03, TRX_CTRL_0). These register bits require a specific treatment, for details see Section 9.6.4 "Master Clock Signal Output (CLKM)" on page 117.



A reset forces the radio transceiver into TRX_OFF state. If the device is still in the P_ON state it remains in the P_ON state though.

A reset is initiated with pin /RST = L and the state is left after setting /RST = H. The reset pulse should have a minimum length as specified in Section 12.4 "Digital Interface Timing Characteristics" on page 157 see parameter 12.4.13.

During reset the microcontroller has to set the radio transceiver control pins SLP_TR and /SEL to their default values.

An overview about the register reset values is provided in Table 14-1 on page 170.

7.1.3 Interrupt Handling

All interrupts provided by the AT86RF231 (see Table 6-9 on page 29) are supported in Basic Operating Mode.

For example, interrupts are provided to observe the status of radio transceiver RX and TX operations.

On receive IRQ_2 (RX_START) indicates the detection of a valid PHR first, IRQ_5 (AMI) an address match and IRQ_3 (TRX_END) the completion of the frame reception.

On transmit IRQ_3 (TRX_END) indicates the completion of the frame transmission.

Figure 7-2 on page 39 shows an example for a transmit/receive transaction between two devices and the related interrupt events in Basic Operating Mode. Device 1 transmits a frame containing a MAC header (in this example of length 7), payload and valid FCS. The frame is received by Device 2 which generates the interrupts during the processing of the incoming frame. The received frame is stored in the Frame Buffer.

The first interrupt IRQ 2 (RX START) signals the reception of a valid PHR.

If the received frame passes the address filter, refer to Section 7.2.3.5 "Frame Filtering" on page 61, an address match interrupt IRQ_5 (AMI) is issued after the reception of the MAC header (MHR).

In Basic Operating Mode the third interrupt IRQ_3 (TRX_END) is issued at the end of the received frame. In Extended Operating Mode, refer to Section 7.2 "Extended Operating Mode" on page 47; the interrupt is only issued if the received frame passes the address filter and the FCS is valid. Further exceptions are explained in Section 7.2 "Extended Operating Mode" on page 47.

Processing delay t_{IRQ} is a typical value, refer to Section 12.4 "Digital Interface Timing Characteristics" on page 157.



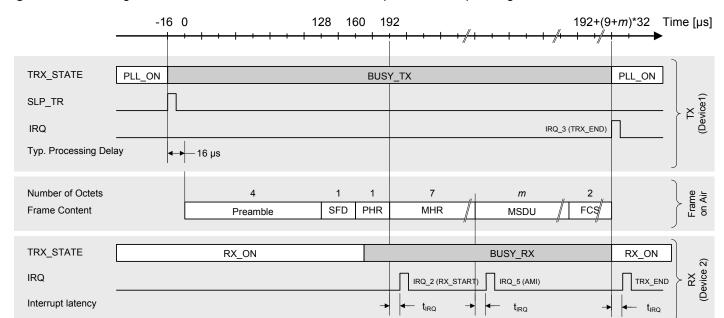


Figure 7-2. Timing of RX_START, AMI and TRX_END Interrupts in Basic Operating Mode

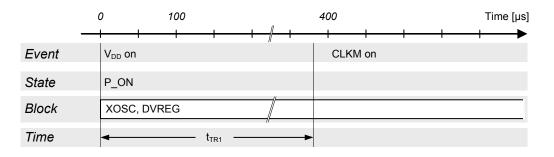
7.1.4 Basic Operating Mode Timing

The following paragraphs depict state transitions and their timing properties. Timing figures are explained in Table 7-1 on page 42 and Section 12.4 "Digital Interface Timing Characteristics" on page 157.

7.1.4.1 Power-on Procedure

The power-on procedure to P_ON state is shown in Figure 7-3 on page 39.

Figure 7-3. Power-on Procedure to P_ON State



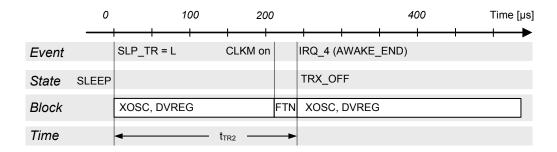
When the external supply voltage (V_{DD}) is firstly supplied to the AT86RF231, the radio transceiver enables the crystal oscillator (XOSC) and the internal 1.8 V voltage regulator for the digital domain (DVREG). After t_{TR1} = 330 μ s (typ.), the master clock signal is available at pin 17 (CLKM) at default rate of 1 MHz. If CLKM is available the SPI is already enabled and can be used to control the transceiver. As long as no state change towards state TRX_OFF is performed the radio transceiver remains in P_ON state.



7.1.4.2 Wake-up Procedure

The wake-up procedure from SLEEP state is shown in Figure 7-4 on page 40.

Figure 7-4. Wake-up Procedure from SLEEP State



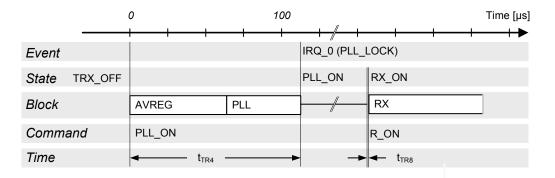
The radio transceivers SLEEP state is left by releasing pin SLP_TR to logic low. This restarts the XOSC and DVREG. After t_{TR2} = 380 μ s (typ.) the radio transceiver enters TRX_OFF state. The internal clock signal is available and provided to pin 17 (CLKM), if CLKM was enabled.

This procedure is similar to the Power-On Procedure. However the radio transceiver continues the state change automatically to the TRX_OFF state. During this the filter-tuning network (FTN) calibration is performed. Entering TRX_OFF state is signaled by IRQ_4 (AWAKE_END), if this interrupt was enabled by the appropriate mask register bit.

7.1.4.3 PLL_ON and RX_ON States

The transition from TRX_OFF to PLL_ON and RX_ON mode is shown in Figure 7-5 on page 40.

Figure 7-5. Transmission from TRX_OFF to PLL_ON and RX_ON State



Note: If TRX_CMD = RX_ON in TRX_OFF state RX_ON state is entered immediately, even if the PLL has not settled.

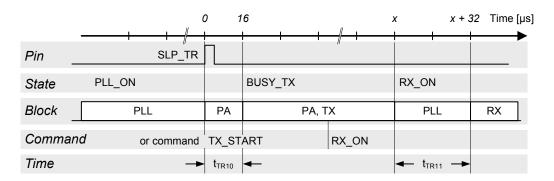
In TRX_OFF state, entering the commands PLL_ON or RX_ON initiates a ramp-up sequence of the internal 1.8V voltage regulator for the analog domain (AVREG). RX_ON state can be entered any time from PLL_ON state regardless whether the PLL has already locked, which is indicated by IRQ_0 (PLL_LOCK).



7.1.4.4 BUSY_TX and RX_ON States

The transition from PLL_ON to BUSY_TX state and subsequent to RX_ON state is shown in Figure 7-6 on page 41.

Figure 7-6. PLL_ON to BUSY_TX to RX_ON Timing



Starting from PLL_ON state it is further assumed that the PLL is already locked. A transmission is initiated either by a rising edge of pin 11 (SLP_TR) or by command TX_START. The PLL settles to the transmit frequency and the PA is enabled.

 t_{TR10} = 16 µs after initiating the transmission the AT86RF231 changes into BUSY_TX state and the internally generated SHR is transmitted. After that the PSDU data are transmitted from the Frame Buffer.

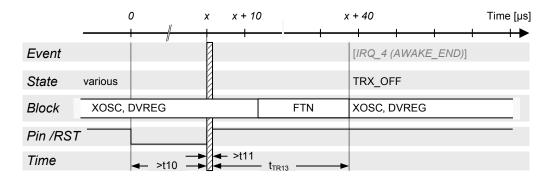
After completing the frame transmission, indicated by IRQ_3 (TRX_END), the PLL settles back to the receive frequency within $t_{TR11} = 32 \mu s$ in state PLL_ON.

If during TX_BUSY the radio transmitter is programmed to change to a receive state it automatically proceeds the state change to RX_ON state after finishing the transmission.

7.1.4.5 Reset Procedure

The radio transceiver reset procedure is shown in Figure 7-7 on page 41.

Figure 7-7. Reset Procedure



Note: Timing figure t_{TR13} refers to Table 7-1 on page 42, t₁₀, t₁₁ refers to Section 12.4 "Digital Interface Timing Characteristics" on page 157.



/RST = L sets all registers to their default values. Exceptions are register bits CLKM_CTRL (register 0x03, TRX_CTRL_0), refer to Section 9.6.4 "Master Clock Signal Output (CLKM)" on page 117.

After releasing the reset pin (/RST = H) the wake-up sequence including an FTN calibration cycle is performed, refer to Section 9.8 "Automatic Filter Tuning (FTN)" on page 125. After that the TRX_OFF state is entered.

Figure 7-7 on page 41 illustrates the reset procedure once the P_ON state was left and the radio transceiver was not in SLEEP state.

The reset procedure is identical for all originating radio transceiver states except of state P_ON and SLEEP state. Instead, here the procedure described in Section 7.1.2.1 "P_ON - Power-On after VDD" on page 34 must be followed to enter the TRX_OFF state.

If the radio transceiver was in SLEEP state, the XOSC and DVREG are enabled before entering TRX OFF state.

If register TRX_STATUS indicates STATE_TRANSITION_IN_PROGRESS during system initialization until the AT86RF231 reaches TRX_OFF, do not try to initiate a further state change while the radio transceiver is in this state.

Notes

- The reset impulse should have a minimum length t₁₀ = 625 ns as specified in Section 12.4 "Digital Interface Timing Characteristics" on page 157, see parameter 12.4.13.
- An access to the device should not occur earlier than t₁₁ □ 625 ns after releasing the pin /RST; refer to Section 12.4 "Digital Interface Timing Characteristics" on page 157, parameter 12.4.14.
- A reset overrides an SPI command request that might be gueued.

7.1.4.6 State Transition Timing Summary

The transition numbers correspond to Figure 7-1 on page 33 and do not include SPI access time if not otherwise stated. See measurement setup in Figure 5-1 on page 12.

Table 7-1. State Transition Timing

No	Symbol	Tra	ansiti	on	Time [µs], (type)	Comments
1	t _{TR1}	P_ON	\Rightarrow	until CLKM available	330	Depends on external capacitor at DVDD (1 μF nom) and crystal oscillator setup (CL = 10 pF)
2	t _{TR2}	SLEEP	\Rightarrow	TRX_OFF	380	Depends on external capacitor at DVDD (1 μ F nom) and crystal oscillator setup (CL = 10 pF) TRX_OFF state indicated by IRQ_4 (AWAKE_END)
3	t _{TR3}	TRX_OFF	\Rightarrow	SLEEP	35*1/f _{CLKM}	For f _{CLKM} > 250 kHz
4	t _{TR4}	TRX_OFF	\Rightarrow	PLL_ON	110	Depends on external capacitor at AVDD (1 µF nom)
5	t _{TR5}	PLL_ON	\Rightarrow	TRX_OFF	1	
6	t _{TR6}	TRX_OFF	\Rightarrow	RX_ON	110	Depends on external capacitor at AVDD (1 µF nom)
7	t _{TR7}	RX_ON	\Rightarrow	TRX_OFF	1	
8	t _{TR8}	PLL_ON	\Rightarrow	RX_ON	1	
9	t _{TR9}	RX_ON	\Rightarrow	PLL_ON	1	Transition time is also valid for TX_ARET_ON, RX_AACK_ON



Table 7-1. State Transition Timing (Continued)

No	Symbol	Tra	ansiti	on	Time [µs], (type)	Comments
10	t _{TR10}	PLL_ON	⇒	BUSY_TX	16	When asserting pin 11 (SLP_TR) or TRX_CMD = TX_START first symbol transmission is delayed by 16 μ s delay (PLL settling and PA ramp up)
11	t _{TR11}	BUSY_TX	\Rightarrow	PLL_ON	32	PLL settling time from TX_BUSY to PLL_ON state
12	t _{TR12}	All states	\Rightarrow	TRX_OFF	1	Using TRX_CMD = FORCE_TRX_OFF (see register 0x02, TRX_STATE), Not valid for SLEEP state
13	t _{TR13}	RESET	\Rightarrow	TRX_OFF	37	Valid for P_ON or SLEEP state
14	t _{TR14}	Various states	\Rightarrow	PLL_ON	1	Using TRX_CMD = FORCE_PLL_ON (see register 0x02, TRX_STATE), Not valid for SLEEP, P_ON, RESET, TRX_OFF and *_NOCLK

The state transition timing is calculated based on the timing of the individual blocks shown in Figure 7-3 on page 39 to Figure 7-7 on page 41. The worst case values include maximum operating temperature, minimum supply voltage, and device parameter variations.

Table 7-2. Analog Block Initialization and Settling Time

No	Symbol	Block	Time [µs], (type)	Time [µs], (max)	Comment
15	t _{TR15}	xosc	330	1000	Leaving SLEEP state, depends on crystal Q factor and load capacitor
16	t _{TR16}	FTN		25	FTN tuning time fixed
17	t _{TR17}	DVREG	60	1000	Depends on external bypass capacitor at DVDD (CB3 = 1 μ F nom., 10 μ F worst case), depends on V _{DD}
18	t _{TR18}	AVREG	60	1000	Depends on external bypass capacitor at AVDD (CB1 = 1 μ F nom, 10 μ F worst case), depends on V _{DD}
19	t _{TR19}	PLL, initial	110	155	PLL settling time TRX_OFF⇒PLL_ON, including 60 μs AVREG settling time
20	t _{TR20}	PLL settling	11	24	Settling time between channels switch
21	t _{TR21}	PLL, CF cal	35		PLL center frequency calibration, refer to Section 9.7.4
22	t _{TR22}	PLL, DCU cal	6	3	PLL DCU calibration, refer to Section 9.7.4
23	t _{TR23}	PLL, RX⇒TX	1	6	Maximum PLL settling time RX⇒TX
24	t _{TR24}	PLL, TX⇒RX	3	2	Maximum PLL settling time TX⇒RX
25	t _{TR25}	RSSI, update	2	2	RSSI update period in receive states, refer to Section 8.3.2
26	t _{TR26}	ED	14	10	ED measurement period, refer to Section 8.4.2
27	t _{TR27}	SHR, sync	96		Typical SHR synchronisation period, refer to Section 8.4.2
28	t _{TR28}	CCA	140		CCA measurement period, refer to Section 8.5.2
29	t _{TR29}	Random value	1		Random value update period, refer to Section 11.2.1



7.1.5 Register Description

Register 0x01 (TRX_STATUS):

A read access to TRX_STATUS register signals the current radio transceiver state. A state change is initiated by writing a state transition command to register bits TRX_CMD (register 0x02, TRX_STATE). Alternatively a state transition can be initiated by the rising edge of pin 11 (SLP_TR) in the appropriate state.

This register is used for Basic and Extended Operating Mode, refer to Section 7.2 "Extended Operating Mode" on page 47.

Bit	7	6	5	4	3	2	1	0	_
+0x01	CCA_DONE	CCA_STATUS	Reserved			TRX_STATUS			TRX_STATUS
Read/Write	R	R	R	R	R	R	R	R	•
Reset Value	0	0	0	0	0	0	0	0	

• Bit 7 - CCA_DONE

Refer to Section 8.5 "Clear Channel Assessment (CCA)" on page 94.

• Bit 6 - CCA STATUS

Refer to Section 8.5 "Clear Channel Assessment (CCA)" on page 94.

• Bit 5 - Reserved

• Bit [4:0] - TRX STATUS

The register bits TRX_STATUS signals the current radio transceiver status. If the requested state transition is not completed yet, the TRX_STATUS returns STATE_TRANSITION_IN_PROGRESS. Do not try to initiate a further state change while the radio transceiver is in STATE_TRANSITION_IN_PROGRESS. State transition timings are defined in Table 7-1 on page 42.



 Table 7-3.
 Radio Transceiver Status, Register Bits TRX_STATUS

Register Bits	Value	State Description
TRX_STATUS	<u>0x00</u>	P_ON
	0x01	BUSY_RX
	0x02	BUSY_TX
	0x06	RX_ON
	0x08	TRX_OFF (CLK Mode)
	0x09	PLL_ON (TX_ON)
	0x0F ⁽³⁾	SLEEP
	0x11 ⁽¹⁾	BUSY_RX_AACK
	0x12 ⁽¹⁾	BUSY_TX_ARET
	0x16 ⁽¹⁾	RX_AACK_ON
	0x19 ⁽¹⁾	TX_ARET_ON
	0x1C	RX_ON_NOCLK
	0x1D ⁽¹⁾	RX_AACK_ON_NOCLK
	0x1E ⁽¹⁾	BUSY_RX_AACK_NOCLK
	0x1F ⁽²⁾	STATE_TRANSITION_IN_PROGRESS
		All other values are reserved

Notes: 1. Extended Operating Mode only, refers to Section 7.2 "Extended Operating Mode" on page 47.

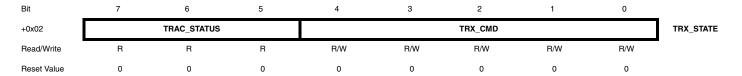
- 2. Do not try to initiate a further state change while the radio transceiver is in STATE_TRANSITION_IN_PROGRESS state.
- 3. In SLEEP state register not accessible.



Register 0x02 (TRX_STATE):

The radio transceiver states are controlled via register bits TRX_CMD, which receives the state transition commands.

This register is used for Basic and Extended Operating Mode, refer to Section 7.2 "Extended Operating Mode" on page 47.



• Bit [7:5] - TRAC_STATUS

Refer to Section 7.2.7 "Register Description - Control Registers" on page 68.

• Bit [4:0] - TRX_CMD

A write access to register bits TRX_CMD initiate a radio transceiver state transition towards the new state as defined by the write access:

Table 7-4. State Control Command, Register Bits TRX_CMD

Register Bit	Value	State Description
TRX_CMD	<u>0x00</u>	NOP
	0x02	TX_START
	0x03	FORCE_TRX_OFF
	0x04 ⁽¹⁾	FORCE_PLL_ON
	0x06	RX_ON
	0x08	TRX_OFF (CLK Mode)
	0x09	PLL_ON (TX_ON)
	0x16 ⁽²⁾	RX_AACK_ON
	0x19 ⁽²⁾	TX_ARET_ON
		All other values are reserved and mapped to NOP

Notes: 1. FORCE_PLL_ON is not valid for states SLEEP, P_ON, RESET, TRX_OFF, and all *_NOCLK states, as well as STATE_TRANSITION_IN_PROGRESS towards these states.

2. Extended Operating Mode only, refers to Section 7.2.7 "Register Description - Control Registers" on page 68.



7.2 Extended Operating Mode

The Extended Operating Mode is a hardware MAC accelerator and goes beyond the basic radio transceiver functionality provided by the Basic Operating Mode. It handles time critical MAC tasks, requested by the IEEE 802.15.4 standard, by hardware, such as automatic acknowledgement, automatic CSMA-CA and retransmission. This results in a more efficient IEEE 802.15.4 software MAC implementation including reduced code size and may allow the use of a smaller microcontroller or to operate at low clock rates.

The Extended Operating Mode is designed to support IEEE 802.15.4-2006 compliant frames; the mode is backward compatible to IEEE 802.15.4-2003 and supports non IEEE 802.15.4 compliant frames. This mode comprises the following procedures:

Automatic acknowledgement (RX_AACK) divides into the tasks:

- · Frame reception and automatic FCS check
- Configurable addressing fields check
- · Interrupt indicating address match
- Interrupt indicating frame reception, if it passes address filtering and FCS check
- Automatic ACK frame transmission (if the received frame passed the address filter and FCS check and if an ACK is required by the frame type and ACK request)
- Support of slotted acknowledgment using SLP_TR pin

Automatic CSMA-CA and Retransmission (TX_ARET) divides into the tasks:

- CSMA-CA including automatic CCA retry and random back-off
- Frame transmission and automatic FCS field generation
- Reception of ACK frame (if an ACK was requested)
- Automatic frame retry if ACK was expected but not received
- · Interrupt signaling with transaction status

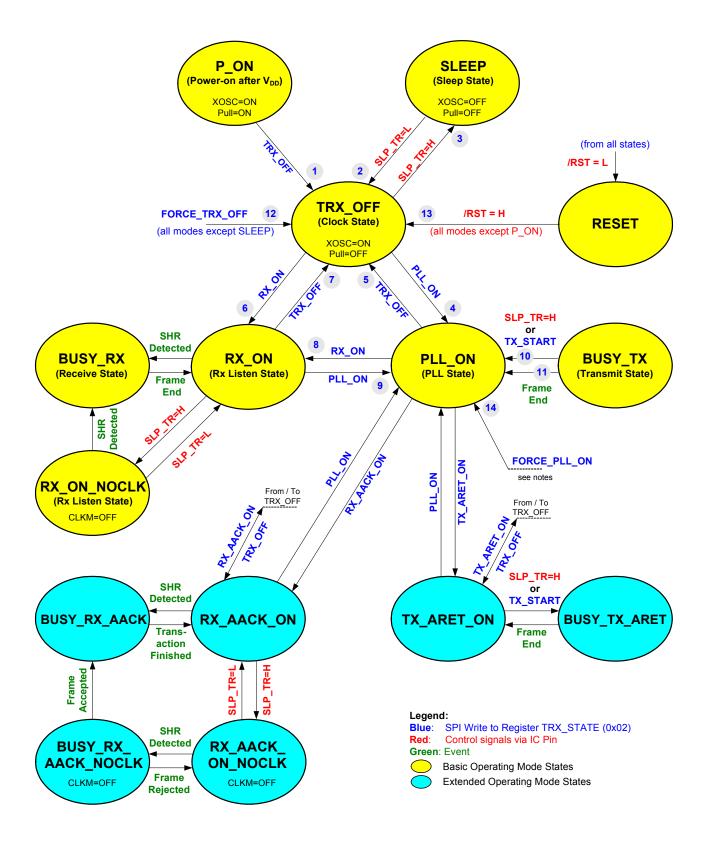
Automatic FCS check and generation, refer to Section 8.2 "Frame Check Sequence (FCS)" on page 85, is used by the RX_AACK and TX_ARET modes. In RX_AACK mode, an automatic FCS check is always performed for incoming frames.

In TX_ARET mode, an ACK, received within the time required by IEEE 802.15.4, is accepted if the FCS is valid, and if the sequence number of the ACK matches the sequence number of the previously transmitted frame. Dependent on the value of the frame pending subfield in the received acknowledgement frame the transaction status is set, see Table 7-16 on page 70.

An AT86RF231 state diagram including the Extended Operating Mode states is shown in Figure 7-8 on page 48. Yellow marked states represent the Basic Operating Mode; blue marked states represent the Extended Operating Mode.



Figure 7-8. Extended Operating Mode State Diagram



7.2.1 State Control

The Extended Operating Mode states RX_AACK and TX_ARET are controlled via register bits TRX_CMD (register 0x02, TRX_STATE), which receives the state transition commands. The states are entered from TRX_OFF or PLL_ON state as illustrated by Figure 7-8 on page 48. The completion of each state change command shall always be confirmed by reading the register 0x01 (TRX_STATUS).

RX AACK - Receive with Automatic ACK

A state transition to RX_AACK_ON from PLL_ON or TRX_OFF is initiated by writing the command RX_AACK_ON to the register bits TRX_CMD. The state change can be confirmed by reading register 0x01 (TRX_STATUS), those changes to RX_AACK_ON or BUSY_RX_AACK on success. The latter one is returned if a frame is currently about being received.

The RX_AACK state is left by writing command TRX_OFF or PLL_ON to the register bits TRX_CMD. If the AT86RF231 is within a frame receive or acknowledgment procedure (BUSY_RX_AACK) the state change is executed after finish. Alternatively, the commands FORCE_TRX_OFF or FORCE_PLL_ON can be used to cancel the RX_AACK transaction and change into radio transceiver state TRX_OFF or PLL_ON, respectively.

TX_ARET - Transmit with Automatic Retry and CSMA-CA Retry

Similarly, a state transition to TX_ARET_ON from PLL_ON or TRX_OFF is initiated by writing command TX_ARET_ON to register bits TRX_CMD. The radio transceiver is in the TX_ARET_ON state after TRX_STATUS (register 0x01) changes to TX_ARET_ON. The TX_ARET transaction is started with a rising edge of pin 11 (SLP_TR) or writing the command TX_START to register bits TRX_CMD.

The TX_ARET state is left by writing the command TRX_OFF or PLL_ON to the register bits TRX_CMD. If the AT86RF231 is within a CSMA-CA, a frame-transmit or an acknowledgment procedure (BUSY_TX_ARET) the state change is executed after finish. Alternatively the command FORCE_TRX_OFF or FORCE_PLL_ON can be used to instantly terminate the TX_ARET transaction and change into radio transceiver state TRX_OFF or PLL_ON, respectively.

Note

 A state change request from TRX_OFF to RX_AACK_ON or TX_ARET_ON internally passes the state PLL_ON to initiate the radio transceiver. Thus the readiness to receive or transmit data is delayed accordingly. It is recommended to use interrupt IRQ_0 (PLL_LOCK) as an indicator.



7.2.2 Configuration

The use of the Extended Operating Mode is based on Basic Operating Mode functionality. Only features beyond the basic radio transceiver functionality are described in the following sections. For details on the Basic Operating Mode refer to Section 7.1 "Basic Operating Mode" on page 33.

When using the RX_AACK or TX_ARET modes, the following registers needs to be configured.

RX_AACK configuration steps:

• Short address, PAN-ID and IEEE address registers 0x20 - 0x2B

Configure RX_AACK properties

- Handling of Frame Version SubfieldHandling of Pending Data Indicator
- Characterize as PAN coordinator
- Handling of Slotted Acknowledgement
- Additional Frame Filtering Properties

registers 0x17, 0x2E

registers 0x2C, 0x2E

- Promiscuous Mode
- Enable or disable automatic ACK generation
- Handling of reserved frame types

The addresses for the address match algorithm are to be stored in the appropriate address registers. Additional control of the RX_AACK mode is done with register 0x17 (XAH_CTRL_1) and register 0x2E (CSMA_SEED_1).

As long as a short address has not been set, only broadcast frames and frames matching the IEEE address can be received.

Configuration examples for different device operating modes and handling of various frame types can be found in Section 7.2.3.1 "Description of RX_AACK Configuration Bits" on page 54.

TX ARET configuration steps:

Leave register bit TX_AUTO_CRC_ON = 1 register 0x04, TRX_CTRL_1

• Configure CSMA-CA

MAX_FRAME_RETRIES
 MAX_CSMA_RETRIES
 CSMA_SEED
 MAX_BE, MIN_BE
 register 0x2C, XAH_CTRL_0
 register 0x2D, 0x2E
 register 0x2F, CSMA_BE

Configure CCA (see Section 8.5)

MAX_FRAME_RETRIES (register 0x2C) defines the maximum number of frame retransmissions.

The register bits MAX_CSMA_RETRIES (register 0x2C) configure the number of CSMA-CA retries after a busy channel is detected.



The CSMA_SEED_0 and CSMA_SEED_1 register bits (registers 0x2D, 0x2E) define a random seed for the back-off-time random-number generator in the AT86RF231.

The MAX_BE and MIN_BE register bits (register 0x2F) sets the maximum and minimum CSMA back-off exponent (according to [1]).

7.2.3 RX_AACK_ON - Receive with Automatic ACK

The general functionality of the RX_AACK procedure is shown in Figure 7-9 on page 53.

The gray shaded area is the standard flow of an RX_AACK transaction for IEEE 802.15.4 compliant frames, refer Section 7.2.3.2 "Configuration of IEEE Scenarios" on page 55. All other procedures are exceptions for specific operating modes or frame formats, refer to Section 7.2.3.3 "Configuration of non IEEE 802.15.4 Compliant Scenarios" on page 58.

The frame filtering operations is described in detail in Section 7.2.3.5 "Frame Filtering" on page 61.

In RX_AACK_ON state, the radio transceiver listens for incoming frames. After detecting a valid PHR, the radio transceiver parses the frame content of the MAC header (MHR), refer to Section 8.1.2 "MAC Protocol Layer Data Unit (MPDU)" on page 80.

Generally, at nodes, configured as a normal device or PAN coordinator, a frame is not indicated if the frame filter does not match and the FCS is invalid. Otherwise, the interrupt IRQ_3 (TRX_END) is issued after the completion of the frame reception. The microcontroller can then read the frame. An exception applies if promiscuous mode is enabled; see Section 7.2.3.2 "Configuration of IEEE Scenarios" on page 55, in that case an IRQ_3 (TRX_END) interrupt is issued, even if the FCS fails.

If the content of the MAC addressing fields of the received frame (refer to IEEE 802.15.4 section 7.2.1) matches one of the configured addresses, dependent on the addressing mode, an address match interrupt IRQ_5 (AMI) is issued, refer to Section 7.2.3.5 "Frame Filtering" on page 61. The expected address values are to be stored in registers 0x20 - 0x2B (Short address, PAN-ID and IEEE address). Frame filtering as described in Section 7.2.3.5 "Frame Filtering" on page 61 is also valid for Basic Operating Mode.

During reception the AT86RF231 parses bit [5] (ACK Request) of the frame control field of the received data or MAC command frame to check if an ACK reply is expected. In that case and if the frame passes the third level of filtering, see IEEE 802.15.4-2006, section 7.5.6.2, the radio transceiver automatically generates and transmits an ACK frame.

The content of the frame pending subfield of the ACK response is set by register bit AACK_SET_PD (register 0x2E, CSMA_SEED_1) when the ACK frame is sent in response to a data request MAC command frame, otherwise this subfield is set to 0. The sequence number is copied from the received frame.

Optionally, the start of the transmission of the acknowledgement frame can be influenced by register bit AACK_ACK_TIME. Default value (according to standard IEEE 802.15.4) is 12 symbol times after the reception of the last symbol of a data or MAC command frame.

If the register bit AACK_DIS_ACK (register 0x2E, CSMA_SEED_1) is set, no acknowledgement frame is sent even if an acknowledgment frame was requested. This is useful for operating the MAC hardware accelerator in promiscuous mode, see Section 7.2.3.2 "Configuration of IEEE Scenarios" on page 55.

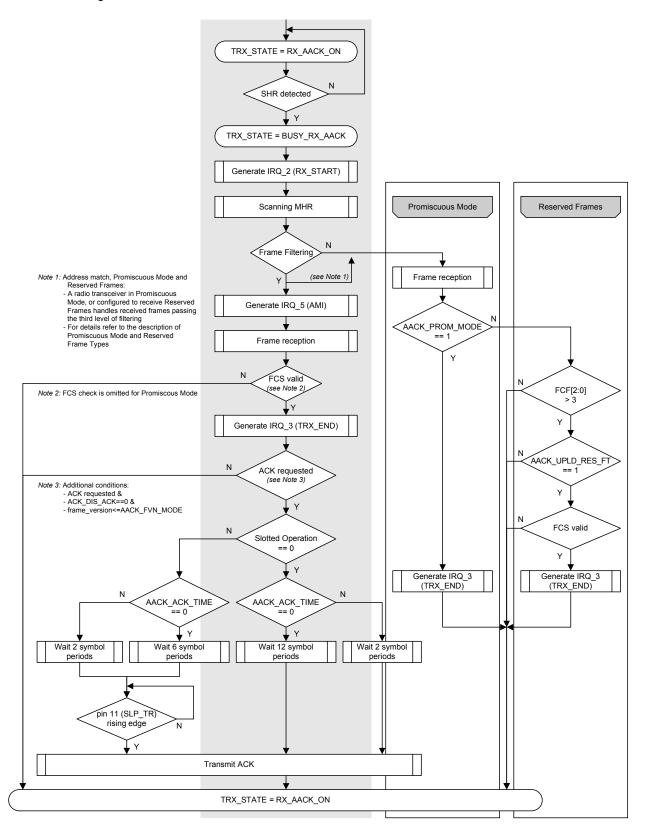


The status of the RX_AACK operation is indicated by register bits TRAC_STATUS (register 0x02, TRAC_STATUS), see Section 7.2.7 "Register Description - Control Registers" on page 68.

During the operations described above the AT86RF231 remains in BUSY_RX_AACK state.



Figure 7-9. Flow Diagram of RX_AACK





7.2.3.1 Description of RX_AACK Configuration Bits

Overview

Table 7-5 on page 54 summarizes all register bits which affect the behavior of an RX_AACK transaction. For address filtering it is further required to setup address registers to match to the expected address.

Configuration and address bits are to be set in TRX_OFF or PLL_ON state prior to switching to RX_AACK mode.

A graphical representation of various operating modes is illustrated in Figure 7-9 on page 53.

Table 7-5. Overview of RX_AACK Configuration Bits

Register	Register	Register Name	Description
Address	Bits		
0x20,0x21 0x22,0x23 0x24		SHORT_ADDR_0/1 PAN_ADDR_0/1 IEEE_ADDR_0	Set node addresses
0x2B	_	IEEE_ADDR_7	D
0x0C	7	RX_SAFE_MODE	Protect buffer after frame receive
0x17	1	AACK_PROM_MODE	Support promiscuous mode
0x17	2	AACK_ACK_TIME	Change auto acknowledge start time
0x17	4	AACK_UPLD_RES_FT	Enable reserved frame type reception, needed to receive non-standard compliant frames
0x17	5	AACK_FLTR_RES_FT	Filter reserved frame types like data frame type, needed for filtering of non-standard compliant frames
0x2C	0	SLOTTED_OPERATION	If set, acknowledgment transmission has to be triggered by pin 11 (SLP_TR)
0x2E	3	AACK_I_AM_COORD	If set, the device is a PAN coordinator
0x2E	4	AACK_DIS_ACK	Disable generation of acknowledgment
0x2E	5	AACK_SET_PD	Set frame pending subfield in Frame Control Field (FCF), refer to Section 8.1.2.2
0x2E	7:6	AACK_FVN_MODE	Controls the ACK behavior, depending on FCF frame version number

The usage of the RX_AACK configuration bits for various operating modes of a node is explained in the following sections. Configuration bits not mentioned in the following two sections should be set to their reset values according to Table 14-1 on page 170.

All registers mentioned in Table 7-5 on page 54 are described in Section 7.2.6 "Register Summary" on page 68.

Note, that the general behavior of the "AT86RF231 Extended Feature Set", Section 11. "AT86RF231 Extended Feature Set" on page 128, settings:



OQPSK_DATA_RATE (PSDU data rate)
 SFD_VALUE (alternative SFD value)
 ANT_DIV (Antenna Diversity)

• RX_PDT_LEVEL (blocking frame reception of lower power signals)

are completely independent from RX_AACK mode. Each of these operating modes can be combined with the RX_AACK mode.

7.2.3.2 Configuration of IEEE Scenarios

Normal Device

Table 7-6 on page 55 shows a typical RX_AACK configuration of an IEEE 802.15.4 device operating as a normal device, rather than a PAN coordinator or router.

Table 7-6. Configuration of IEEE 802.15.4 Devices

Register	Register	Register Name	Description
Address	Bits		
0x20,0x21 0x22,0x23 0x24, 0x2B		SHORT_ADDR_0/1 PAN_ADDR_0/1 IEEE_ADDR_0 IEEE_ADDR_7	Set node addresses
0x0C	7	RX_SAFE_MODE	Q: disable frame protection1: enable frame protection
0x2C	0	SLOTTED_OPERATION	O: if transceiver works in unslotted mode1: if transceiver works in slotted mode
0x2E	7:6	AACK_FVN_MODE	Controls the ACK behavior, depending on FCF frame version number 0x00: acknowledges only frames with version number 0, i.e. according to IEEE 802.15.4-2003 frames 0x01: acknowledges only frames with version number 0 or 1, i.e. frames according to IEEE 802.15.4-2006 0x10: acknowledges only frames with version number 0 or 1 or 2 0x11: acknowledges all frames, independent of the FCF frame version number

Notes

- If no short address has been configured before the device has been assigned one by the PAN-coordinator, only frames directed to either the broadcast address or the IEEE address are received.
- In IEEE 802.15.4-2003 standard the frame version subfield did not yet exist but was marked as reserved. According to this standard, reserved fields have to be set to zero. On the other hand, IEEE 802.15.4-2003 standard requires ignoring reserved bits upon reception. Thus, there is a contradiction in the standard which can be interpreted in two ways:



- 1. If a network should only allow access to nodes which use the IEEE 802.15.4-2003, then AACK_FVN_MODE should be set to 0.
- 2. If a device should acknowledge all frames independent of its frame version, AACK_FVN_MODE should be set to 3. However, this can result in conflicts with co-existing IEEE 802.15.4-2006 standard compliant networks.

The same holds for PAN coordinators, see Table 7-7 on page 56.

PAN-Coordinator

Table 7-7. Configuration of a PAN Coordinator

Register	Register	Register Name	Description
Address	Bits		
0x20,0x21 0x22,0x23 0x24, 0x2B		SHORT_ADDR_0/1 PAN_ADDR_0/1 IEEE_ADDR_0 IEEE_ADDR_7	Set node addresses
0x0C	7	RX_SAFE_MODE	O: disable frame protection 1: enable frame protection
0x2C	0	SLOTTED_OPERATION	O: if transceiver works in unslotted mode 1: if transceiver works in slotted mode
0x2E	3	AACK_I_AM_COORD	1: device is PAN coordinator
0x2E	5	AACK_SET_PD	O: frame pending subfield is not set in FC 1: frame pending subfield is set in FCF
0x2E	7:6	AACK_FVN_MODE	Controls the ACK behavior, depending on FCF frame version number <i>0x00</i> : acknowledges only frames with version number 0, i.e. according to IEEE 802.15.4-2003 frames 0x01: acknowledges only frames with version number 0 or 1, i.e. frames according to IEEE 802.15.4-2006 0x10: acknowledges only frames with version number 0 or 1 or 2 0x11: acknowledges all frames, independent of the FCF frame version number

Promiscuous Mode

The promiscuous mode is described in IEEE 802.15.4-2006, section 7.5.6.2. This mode is further illustrated in Figure 7-9 on page 53. According to IEEE 802.15.4-2006 when in promiscuous mode, the MAC sub layer shall pass received frames with correct FCS to the next higher layer and shall not be processed further. That implies that frames should never be acknowledged.



Only second level filter rules as defined by IEEE 802.15.4-2006, section 7.5.6.2, are applied to the received frame.

Table 7-8 on page 57 shows the typical configuration of a device operating promiscuous mode.

Table 7-8. Configuration of Promiscuous Mode

Register	Register	Register Name	Description
Address	Bits		
0x20,0x21 0x22,0x23 0x24, 0x2B		SHORT_ADDR_0/1 PAN_ADDR_0/1 IEEE_ADDR_0 IEEE_ADDR_7	Address shall be set: 0x00
0x17	1	AACK_PROM_MODE	1: Enable promiscuous Mode
0x2E	4	AACK_DIS_ACK	1: Disable generation of acknowledgment
0x2E	7:6	AACK_FVN_MODE	Controls the ACK behavior, depending on FCF frame version number <i>0x00</i> : acknowledges only frames with version number 0, i.e. according to IEEE 802.15.4-2003 frames <i>0x01</i> : acknowledges only frames with version number 0 or 1, i.e. frames according to IEEE 802.15.4-2006 <i>0x10</i> : acknowledges only frames with version number 0 or 1 or 2 <i>0x11</i> : acknowledges all frames, independent of the FCF frame version number

If the radio transceiver is in promiscuous mode, second level of filtering according to IEEE 802.15.4-2006, section 7.5.6.2, is applied to a received frame. However, an IRQ_3 (TRX_END) is issued even if the FCS is invalid. Thus, it is necessary to read register bit RX_CRC_VALID (register 0x06, PHY_RSSI) after IRQ_3 (TRX_END) in order to verify the reception of a frame with a valid FCS.

If a device, operating in promiscuous mode, receives a frame with a valid FCS which further passed the third level of filtering according to IEEE 802.15.4-2006, section 7.5.6.2, an acknowledgement frame would be transmitted. According to the definition of the promiscuous mode a received frame shall not be acknowledged, even if it is requested. Thus register bit AACK_DIS_ACK (register 0x2E, CSMA_SEED_1) has to be set to 1.

In all receive modes an IRQ_5 (AMI) interrupt is issued, when the received frame matches the node's address according to the filter rules described in Section 7.2.3.5 "Frame Filtering" on page 61

Alternatively, in Basic Operating Mode RX_ON state, when a valid PHR is detected, an IRQ_2 (RX_START) is generated and the frame is received. The end of the frame reception is signalized with an IRQ_3 (TRX_END). At the same time the register bit RX_CRC_VALID (register 0x06, PHY_RSSI) is updated with the result of the FCS check (see Section 8.2 "Frame Check Sequence (FCS)" on page 85). According to the promiscuous mode definition the RX_CRC_VALID bit needs to be checked in order to dismiss corrupted frames.



7.2.3.3 Configuration of non IEEE 802.15.4 Compliant Scenarios

Sniffer

Table 7-9 on page 58 shows an RX_AACK configuration to setup a sniffer device. Other RX_AACK configuration bits, refer to Table 7-5 on page 54, should be set to their reset values.

All frames received are indicated by an IRQ_2 (RX_START) and IRQ_3 (TRX_END). After frame reception register bit RX_CRC_VALID (register 0x06, PHY_RSSI) is updated with the result of the FCS check (see Section 8.2 "Frame Check Sequence (FCS)" on page 85). The RX_CRC_VALID bit needs to be checked in order to dismiss corrupted frames.

 Table 7-9.
 Configuration of a Sniffer Device

Register Register		Register Name	Description
Address	Bits		
0x17	1	AACK_PROM_MODE	1: Enable promiscuous Mode
0x2E	4	AACK_DIS_ACK	1: Disable generation of acknowledgment

This operating mode is similar to the promiscuous mode.

Reception of Reserved Frames

In RX_AACK mode, frames with reserved frame types, refer to Section 8.1.2.2 "Frame Control Field (FCF)" on page 80, can also be handled. This might be required when implementing proprietary, non-standard compliant, protocols. It is an extension of the address filtering in RX_AACK mode. Received frames are either handled similar to data frames, or may be allowed to completely bypass the address filter.

Table 7-10 on page 58 shows the required configuration for a node to receive reserved frames, Figure 7-9 on page 53 shows the corresponding flow chart.

Table 7-10. RX_AACK Configuration to Receive Reserved Frame Types

Register	Register	Register Name	Description
Address	Bits		
0x20,0x21 0x22,0x23 0x24, 0x2B		SHORT_ADDR_0/1 PAN_ADDR_0/1 IEEE_ADDR_0 IEEE_ADDR_7	Set node addresses
0x0C	7	RX_SAFE_MODE	O: disable frame protection1: enable frame protection
0x17	4	AACK_UPLD_RES_FT	1: Enable reserved frame type reception
0x17	5	AACK_FLTR_RES_FT	Filter reserved frame types like data frame type, see note below O: disable 1: enable
0x2C	0	SLOTTED_OPERATION	O: if transceiver works in unslotted mode 1: if transceiver works in slotted mode



Table 7-10. RX_AACK Configuration to Receive Reserved Frame Types (Continued)

0x2E	3	AACK_I_AM_COORD	Q: device is not PAN coordinator1: device is PAN coordinator
0x2E	4	AACK_DIS_ACK	O: Enable generation of acknowledgment1: Disable generation of acknowledgment
0x2E	7:6	AACK_FVN_MODE	Controls the ACK behavior, depending on FCF frame version number 0x00: acknowledges only frames with version number 0, i.e. according to IEEE 802.15.4-2003 frames 0x01: acknowledges only frames with version number 0 or 1, i.e. frames according to IEEE 802.15.4-2006 0x10: acknowledges only frames with version number 0 or 1 or 2 0x11: acknowledges all frames, independent of the FCF frame version number

There are two different options for handling reserved frame types.

1. AACK_UPLD_RES_FT = 1, AACK_FLT_RES_FT = 0:

Any non-corrupted frame with a reserved frame type is indicated by an IRQ_3 (TRX_END) interrupt. No further address filtering is applied on those frames. An IRQ_5 (AMI) interrupt is never generated and the acknowledgment subfield is ignored.

2. AACK_UPLD_RES_FT = 1, AACK_FLT_RES_FT = 1:

If AACK_FLT_RES_FT = 1 any frame with a reserved frame type is filtered by the address filter similar to a data frame as described in the standard. This implies the generation of the IRQ_5 (AMI) interrupts upon address match. An IRQ_3 (TRX_END) interrupt is only generated if the address matched and the frame was not corrupted. An acknowledgment is only send, when the ACK request subfield was set in the received frame and an IRQ_3 (TRX_END) interrupt occurred.

3. AACK UPLD RES FT = 0:

Any received frame indicated as a reserved frame is discarded.

Short Acknowledgment Frame (ACK) Start Timing

Register bit AACK_ACK_TIME (register 0x17, XAH_CTRL_1), see Table 7-11 on page 60, defines the symbol time between frame reception and transmission of an acknowledgment frame.



Register	Register	Register Name	Description	
Address	Bit			
0x17	2	AACK_ACK_TIME	Q: Standard compliant acknowledgement timing of 12 symbol periods. In slotted acknowledge- ment operation mode, the acknowledgment frame transmission can be triggered 6 symbol periods after reception of the frame earliest.	
			1: Reduced acknowledgment timing of 2 symbol periods (32 µs).	

Table 7-11. Overview of RX AACK Configuration Bits

Note that this feature can be used in all scenarios, independent of other configurations. However, shorter acknowledgment timing is especially useful when using High Data Rate Modes to increase battery lifetime and to improve the overall data throughput; refer to Section 11.3 "High Data Rate Modes" on page 137.

7.2.3.4 RX_AACK_NOCLK - RX_AACK_ON without CLKM

If the AT86RF231 is listening for an incoming frame and the microcontroller is not running an application, the microcontroller can be powered down to decrease the total system power consumption. This special power-down scenario for systems running in clock synchronous mode (see Section 6. "Microcontroller Interface" on page 16) is supported by the AT86RF231 using the state RX_AACK_ON_NOCLK. The radio transceiver functionality in this state is based on that in state RX_AACK_ON with pin 17 (CLKM) disabled.

The RX_AACK_NOCLK state is entered from RX_AACK_ON by a rising edge at pin 11 (SLP_TR). The return to RX_AACK_ON state results either from a successful frame reception or a falling edge on pin SLP_TR.

The CLKM pin is disabled 35 clock cycles after the rising edge at SLP_TR pin. This allows the microcontroller to complete its power-down sequence. This is not valid for clock rates 250 kHz and 62.5 kHz, where the main clock at pin 17 (CLKM) is switched off immediately.

In case of the reception of a valid frame, IRQ_3 (TRX_END) is issued and pin 17 (CLKM) is turned on. A timing diagram is shown in Figure 6-16 on page 28. A received frame is considered valid if it passes address filtering and has a correct FCS. If an ACK was requested the radio transceiver enters BUSY_RX_AACK state and follows the procedure described in Section 7.2.3 "RX_AACK_ON - Receive with Automatic ACK" on page 51.

After the transaction has been completed, the radio transceiver reenters the RX_AACK_ON state.

The radio transceiver reenters the RX_AACK_ON_NOCLK state only, when the next rising edge at SLP_TR pin occurs.

It is not recommended to operate the receiver in state RX_AACK_NOCLK with register bit SLOTTED_OPERATION (register 0x2C, XAH_XTRL_0) set, refer to "Register Description - Control Registers" on page 68.



7.2.3.5 Frame Filtering

Frame Filtering is an evaluation whether or not a received frame is dedicated for this node. To accept a received frame and to generate an address match interrupt IRQ_5 (AMI) a filtering procedure as described in IEEE 802.15.4-2006, section 7.5.6.2 (Third level of filtering) is applied to the frame. The AT86RF231 RX_AACK mode accepts only frames that satisfy all of the following requirements (quote from IEEE 802.15.4-2006, section 7.5.6.2):

- 1. The Frame Type subfield shall not contain a reserved frame type.
- 2. The Frame Version subfield shall not contain a reserved value.
- 3. If a destination PAN identifier is included in the frame, it shall match macPANId or shall be the broadcast PAN identifier (0xFFFF).
- 4. If a short destination address is included in the frame, it shall match either macShortAddress or the broadcast address (0xFFFF). Otherwise, if an extended destination address is included in the frame, it shall match aExtendedAddress.
- 5. If the frame type indicates that the frame is a beacon frame, the source PAN identifier shall match macPANId unless macPANId is equal to 0xFFF, in which case the beacon frame shall be accepted regardless of the source PAN identifier.
- If only source addressing fields are included in a data or MAC command frame, the frame shall be accepted only if the device is the PAN coordinator and the source PAN identifier matches macPANId.

The AT86RF231 requires satisfying two additional rules:

- 7. The frame type indicates that the frame is not an ACK frame (refer to Table 8-4 on page 82).
- 8. At least one address field must be configured.

Address match, indicated by interrupt IRQ_5 (AMI), is further controlled by the content of sub-fields of the frame control field of a received frame according to the following rule:

If (Destination Addressing Mode = 0 OR 1) AND (Source Addressing Mode = 0) no IRQ_5 (AMI) is generated, refer to Section 8.1.2.2 "Frame Control Field (FCF)" on page 80. This effectively causes all acknowledgement frames not to be announced, which otherwise always pass the filter, regardless of whether they are intended for this device or not.

For backward compatibility to IEEE 802.15.4-2003 third level filter rule 2 (Frame Version) can be disabled by register bits AACK_FVN_MODE (register 0x2E, CSMA_SEED_1).

Frame filtering is available in Extended and Basic Operating Mode, refer to Section 7.1 "Basic Operating Mode" on page 33, a frame passing the frame filtering generates an IRQ_5 (AMI), if enabled.

Notes

- Filter rule 1 is affected by register bits AACK_FLTR_RES_FT and AACK_UPLD_RES_FT, Section 7.2.7 "Register Description Control Registers" on page 68.
- Filter rule 2 is affected by register bits AACK_FVN_MODE, Section 7.2.7 "Register Description Control Registers" on page 68.



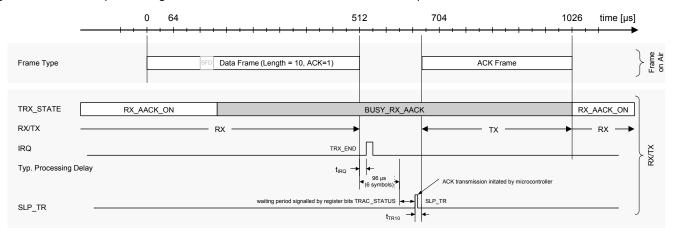
7.2.3.6 RX AACK Slotted Operation - Slotted Acknowledgement

AT86RF231 supports slotted acknowledgement operation, refer to IEEE 802.15.4-2006, section 7.5.6.4.2, in conjunction with the microcontroller.

In RX_AACK mode with register bit SLOTTED_OPERATION (register 0x2C, XAH_CTRL_0) set, the transmission of an acknowledgement frame has to be controlled by the microcontroller. If an ACK frame has to be transmitted, the radio transceiver expects a rising edge on pin 11 (SLP_TR) to actually start the transmission. This waiting state is signaled 6 symbol periods after the reception of the last symbol of a data or MAC command frame by register bits TRAC_STATUS (register 0x02, XAH_CTRL_0), which are set to SUCCESS_WAIT_FOR_ACK in that case. In networks using slotted operation the start of the acknowledgment frame, and thus the exact timing, must be provided by the microcontroller.

A timing example of an RX_AACK transaction with register bit SLOTTED_OPERATION (register 0x2C, XAH_CTRL_0) set is shown in Figure 7-10 on page 62. The acknowledgement frame is ready to transmit 6 symbol times after the reception of the last symbol of a data or MAC command frame. The transmission of the acknowledgement frame is initiated by the microcontroller with the rising edge of pin 11 (SLP_TR) and starts t_{TR10} = 16 µs later. The interrupt latency t_{IRQ} is specified in Section 12.4 "Digital Interface Timing Characteristics" on page 157, parameter 12.4.17.

Figure 7-10. Example Timing of an RX_AACK Transaction for Slotted Operation



If register bit AACK_ACK_TIME (register 0x17, XAH_CTRL_1) is set, an acknowledgment frame can be sent already 2 symbol times after the reception of the last symbol of a data or MAC command frame.

7.2.3.7 RX_AACK Mode Timing

A timing example of an RX_AACK transaction is shown in Figure 7-11 on page 63. In this example a data frame of length 10 with an ACK request is received. The AT86RF231 changes to state BUSY_RX_AACK after SFD detection. The completion of the frame reception is indicated by a TRX_END interrupt. Interrupts IRQ_2 (RX_START) and IRQ_5 (AMI) are disabled in this example. The ACK frame is automatically transmitted after a default wait period of 12 symbols (192 μ s), register bit AACK_ACK_TIME = 0 (reset value). The interrupt latency t_{IRQ} is specified in Section 12.4 "Digital Interface Timing Characteristics" on page 157, parameter 12.4.17.



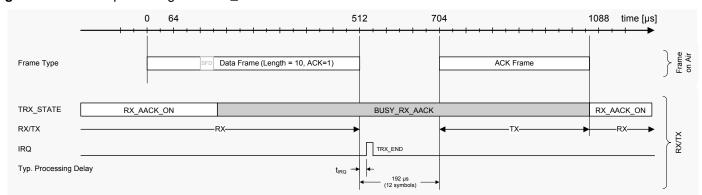


Figure 7-11. Example Timing of an RX_AACK Transaction

If register bit AACK_ACK_TIME (register 0x17, XAH_CTRL_1) is set, an acknowledgment frame is sent already 2 symbol times after the reception of the last symbol of a data or MAC command frame.



7.2.4 TX_ARET_ON - Transmit with Automatic Retry and CSMA-CA Retry

TRX_STATE = TX_ARET_ON $frame_rctr = 0$ TRX_STATE = BUSY_TX_ARET TRAC_STATUS = INVALID (see Note 1) Note 1: If MAX_CSMA_RETRIES = 7 no retry is performed MAX_CSMA_RETRIES csma_rctr = 0 Random Back-Off csma_rctr = csma_rctr + 1
CCA Failure csma_rctr > MAX_CSMA_RETRIES Success Transmit Frame frame_rctr = frame_rctr + 1 ACK requested Receive ACK until timeout ACK valid Ν frame_rctr >
MAX_FRAME_RETRIES Ν Data Pending TRAC_STATUS = SUCCESS TRAC_STATUS = NO_ACK TRAC_STATUS = SUCCESS_DATA_PENDING TRAC_STATUS = CHANNEL_ACCESS_FAILURE Issue IRQ_3 (TRX_END) interrupt

TRX_STATE = TX_ARET_ON

Figure 7-12. Flow Diagram of TX_ARET



Overview

The implemented TX_ARET algorithm is shown in Figure 7-12 on page 64.

In TX_ARET mode, the AT86RF231 first executes the CSMA-CA algorithm, as defined by IEEE 802.15.4-2006, section 7.5.1.4, initiated by a transmit start event. If the channel is IDLE a frame is transmitted from the Frame Buffer. If the acknowledgement frame is requested the radio transceiver additionally checks for an ACK reply.

The completion of the TX_ARET transmit transaction is indicated by an IRQ_3 (TRX_END) interrupt.

Description

Configuration and address bits are to be set in TRX_OFF or PLL_ON state prior to switching to TX_ARET mode. It is further recommended to transfer the PSDU data to the Frame Buffer in advance. The transaction is started by either using pin 11 (SLP_TR), refer to Section 6.5 "Sleep/Wake-up and Transmit Signal (SLP_TR)" on page 27, or writing a TX_START command to register 0x02 (TRX_STATE).

If the CSMA-CA detects a busy channel, it is retried as specified by the register bits MAX_CSMA_RETRIES (register 0x2C, XAH_CTRL_0). In case that CSMA-CA does not detect a clear channel after MAX_CSMA_RETRIES, it aborts the TX_ARET transaction, issues interrupt IRQ_3 (TRX_END), and set the value of the TRAC_STATUS register bits to CHANNEL ACCESS FAILURE.

During transmission of a frame the radio transceiver parses bit 5 (ACK Request) of the MAC header (MHR) frame control field of the PSDU data (PSDU octet #1) to be transmitted to check if an ACK reply is expected.

If an ACK is expected, the radio transceiver automatically switches into receive mode to wait for a valid ACK reply. After receiving an ACK frame the Frame Pending subfield of that frame is parsed and the status register bits TRAC_STATUS are updated accordingly, refer to Table 7-12 on page 66. This receive procedure does not overwrite the Frame Buffer content. Transmit data in the Frame Buffer is not changed during the entire TX_ARET transaction. Received frames other than the expected ACK frame are discarded.

If no valid ACK is received or after timeout of 54 symbol periods (864 μ s), the radio transceiver retries the entire transaction, (including CSMA-CA) until the maximum number of retransmissions (as set by the register bits MAX_FRAME_RETRIES in register 0x2C (XAH_CTRL_0) is exceeded.

After that, the microcontroller may read the value of the register bits TRAC_STATUS (register 0x02, TRX_STATE) to verify whether the transaction was successful or not. The register bits are set according to the following cases, additional exit codes are described in Section 7.2.6 "Register Summary" on page 68:



Value Name **Description** 0 **SUCCESS** The transaction was responded by a valid ACK, or, if no ACK is requested, after a successful frame transmission 1 SUCCESS_DATA_PENDING Equivalent to SUCCESS, indicates pending frame data according to the MHR frame control field of the received ACK response 3 CHANNEL_ACCESS_FAILURE Channel is still busy after MAX_CSMA_RETRIES of CSMA-CA 5 NO_ACK No acknowledgement frames were received during all retry

attempts

Table 7-12. Interpretation of TRAC STATUS register bits

Note that if no ACK is expected (according to the content of the received frame in the Frame Buffer), the radio transceiver issues IRQ_3 (TRX_END) directly after the frame transmission has been completed. The value of register bits TRAC_STATUS (register 0x02, TRX_STATE) is set to SUCCESS.

Entering TX_ARET mode sets TRAC_STATUS = 7

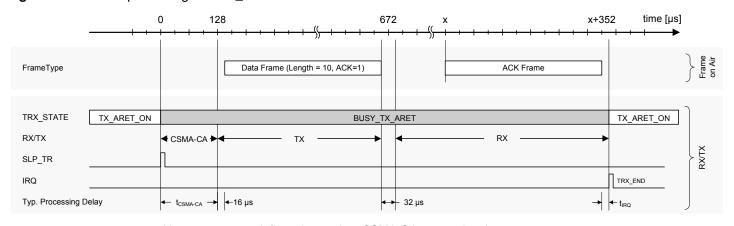
A value of MAX_CSMA_RETRIES = 7 initiates an immediate TX_ARET transaction without performing CSMA-CA. This is required to support slotted acknowledgement operation. Further the value MAX_FRAME_RETRIES is ignored and the TX_ARET transaction is performed only once.

A timing example of a TX_ARET transaction is shown in Figure 7-13 on page 66.

Figure 7-13. Example Timing of a TX ARET Transaction

7

INVALID



Note: t_{CSMA-CA} defines the random CSMA-CA processing time

Here an example data frame of length 10 with an ACK request is transmitted, see Table 7-13 on page 67. After that the AT86RF231 switches to receive mode and expects an acknowledgement response. During the whole transaction including frame transmit, wait for ACK and ACK receive the radio transceiver status register TRX_STATUS (register 0x01, TRX_STATUS) signals BUSY_TX_ARET.

A successful reception of the acknowledgment frame is indicated by IRQ_3 (TRX_END). The status register TRX_STATUS (register 0x01, TRX_STATUS) changes back to TX_ARET_ON. The TX_ARET status register TRAC_STATUS changes as well to TRAC_STATUS = SUCCESS



or TRAC_STATUS = SUCCESS_DATA_PENDING if the frame pending subfield of the received ACK frame was set to 1.

7.2.5 Interrupt Handling

The interrupt handling in the Extended Operating Mode is similar to the Basic Operating Mode, refer to Section 7.1.3 "Interrupt Handling" on page 38. The microcontroller enables interrupts by setting the appropriate bit in register 0x0E (IRQ_MASK).

For RX_AACK and TX_ARET the following interrupts inform about the status of a frame reception and transmission:

 Table 7-13.
 Interrupt Handling in Extended Operating Mode

Mode	Interrupt	Description
RX_AACK	IRQ_2 (RX_START)	Indicates a PHR reception
	IRQ_5 (AMI)	Issued at address match
	IRQ_3 (TRX_END)	Signals completion of RX_AACK transaction if successful
		- A received frame must pass the address filter
		- The FCS is valid
TX_ARET	IRQ_3 (TRX_END)	Signals completion of TX_ARET transaction
Both	IRQ_0 (PLL_LOCK)	Entering RX_AACK_ON or TX_ARET_ON state from TRX_OFF state, the PLL_LOCK interrupt signals that the transaction can be started

RX AACK

For RX_AACK it is recommended to enable IRQ_3 (TRX_END). This interrupt is issued only if a frame passes the frame filtering, refer to Section 7.2.3.5 "Frame Filtering" on page 61 and has a valid FCS. This is in contrast to Basic Operating Mode, refer to Section 7.1.3 "Interrupt Handling" on page 38. The use of the other interrupts is optional.

On reception of a valid PHR an IRQ_2 (RX_START) is issued. IRQ_5 (AMI) indicates address match, refer to filter rules in Section 7.2.3.5 "Frame Filtering" on page 61, and the completion of a frame reception with a valid FCS is indicated by interrupt IRQ_3 (TRX_END).

Thus, it can happen that an IRQ_2 (RX_START) and/or IRQ_5 (AMI) are issued, but no IRQ_3 (TRX_END) interrupt.

TX_ARET

In TX_ARET interrupt IRQ_3 (TRX_END) is only issued after completing the entire TX_ARET transaction.

Acknowledgement frames do not issue IRQ_5 (AMI) or IRQ_3 (TRX_END) interrupts.

All other interrupts as described in Section 6.6 "Interrupt Logic" on page 29, are also available in Extended Operating Mode.



7.2.6 Register Summary

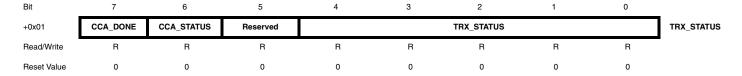
The following registers are to be configured to control the Extended Operating Mode:

Table 7-14. Register Summary

RegAddr	Register Name	Description
0x01	TRX_STATUS	Radio transceiver status, CCA result
0x02	TRX_STATE	Radio transceiver state control, TX_ARET status
0x04	TRX_CTRL_1	TX_AUTO_CRC_ON
0x08	PHY_CC_CCA	CCA mode control, see Section 8.5.6
0x09	CCA_THRES	CCA threshold settings, see Section 8.5.6
0x17	XAH_CTRL_1	RX_AACK control
0x20 - 0x2B		Address filter configuration
		- Short address, PAN-ID and IEEE address
0x2C	XAH_CTRL_0	TX_ARET control, retries value control
0x2D	CSMA_SEED_0	CSMA-CA seed value
0x2E	CSMA_SEED_1	CSMA-CA seed value, RX_AACK control
0x2F	CSMA_BE	CSMA-CA back-off exponent control

7.2.7 Register Description - Control Registers Register 0x01 (TRX_STATUS):

The read-only register TRX_STATUS signals the present state of the radio transceiver as well as the status of a CCA application. A state change is initiated by writing a state transition command to register bits TRX_CMD (register 0x02, TRX_STATE).



• Bit 7 - CCA_DONE

Refer to Section 8.5 "Clear Channel Assessment (CCA)" on page 94, not updated in Extended Operating Mode.

• Bit 6 - CCA_STATUS

Refer to Section 8.5 "Clear Channel Assessment (CCA)" on page 94, not updated in Extended Operating Mode.

• Bit 5 - Reserved

• Bit [4:0] - TRX_STATUS

The register bits TRX_STATUS signal the current radio transceiver status.



Table 7-15. Radio Transceiver Status

Register Bit	Value	State Description
TRX_STATUS	<u>0x00</u>	P_ON
	0x01	BUSY_RX
	0x02	BUSY_TX
	0x06	RX_ON
	0x08	TRX_OFF (CLK Mode)
	0x09	PLL_ON (TX_ON)
	0x0F ⁽¹⁾	SLEEP
	0x11	BUSY_RX_AACK
	0x12	BUSY_TX_ARET
	0x16	RX_AACK_ON
	0x19	TX_ARET_ON
	0x1C	RX_ON_NOCLK
	0x1D	RX_AACK_ON_NOCLK
	0x1E	BUSY_RX_AACK_NOCLK
	0x1F ⁽²⁾	STATE_TRANSITION_IN_PROGRESS
		All other values are reserved

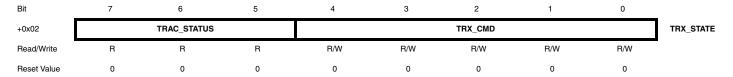
- Notes: 1. In SLEEP state register not accessible.
 - 2. Do not try to initiate a further state change while the radio transceiver is in STATE_TRANSITION_IN_PROGRESS state.

Register 0x02 (TRX_STATE):

The AT86RF231 radio transceiver states are controlled via register TRX_STATE using register bits TRX_CMD. The read-only register bits TRAC_STATUS indicate the status or result of an Extended Operating Mode transaction.

A successful state transition shall be confirmed by reading register bits TRX STATUS (register 0x01, TRX_STATUS).

Register bits TRX_CMD are used for Extended and Basic Operating Mode, refer to Section 7.1 "Basic Operating Mode" on page 33.



• Bit [7:5] - TRAC_STATUS

The status of the RX_AACK and TX_ARET procedure is indicated by register bits TRAC_STATUS. Details of the algorithm and a description of the status information are given in



Section 7.2.3 "RX_AACK_ON - Receive with Automatic ACK" on page 51 and Section 7.2.4 "TX_ARET_ON - Transmit with Automatic Retry and CSMA-CA Retry" on page 64.

Table 7-16. TRAC_STATUS Transaction Status

Register Bits	Value	Description	RX_AACK	TX_ARET
TRAC_STATUS	<u>0</u> ⁽¹⁾	SUCCESS	X	X
	1	SUCCESS_DATA_PENDING		X
	2	SUCCESS_WAIT_FOR_ACK	Х	
	3	CHANNEL_ACCESS_FAILURE		X
	5	NO_ACK		X
	7 ⁽¹⁾	INVALID	Х	Х
		All other values are reserved		

Notes: 1. Even though the reset value for register bits TRAC_STATUS is 0, the RX_AACK and TX_ARET procedures set the register bits to TRAC_STATUS = 7 (INVALID) when it is started.

TX_ARET

SUCCESS_DATA_PENDING: Indicates a successful reception of an ACK frame with

frame pending bit set to 1.

RX AACK

SUCCESS_WAIT_FOR_ACK:

Indicates an ACK frame is about to sent in RX_AACK slotted acknowledgement. Slotted acknowledgement operation must be enabled with register bit SLOTTED_OPERATION (register 0x2C, XAH_XTRL_0). The microcontroller must pulse pin 11 (SLP_TR) at the next back-off slot boundary in order to initiate a transmission of the ACK frame. For details refer to IEEE 802.15.4-2006, section 7.5.6.4.2.

• Bit [4:0] - TRX_CMD

A write access to register bits TRX CMD initiate a radio transceiver state transition:

 Table 7-17.
 State Control Register

Register Bit	Value	State Description
TRX_CMD	<u>0x00</u>	NOP
	0x02	TX_START
	0x03	FORCE_TRX_OFF
	0x04 ⁽¹⁾⁽²⁾	FORCE_PLL_ON
	0x06	RX_ON
	0x08	TRX_OFF (CLK Mode)
	0x09	PLL_ON (TX_ON)
	0x16	RX_AACK_ON
	0x19	TX_ARET_ON
		All other values are reserved and mapped to NOP

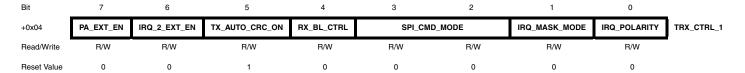


Notes: 1. FORCE_PLL_ON is not valid for states SLEEP, P_ON, RESET, TRX_OFF, and all *_NOCLK states, as well as STATE_TRANSITION_IN_PROGRESS towards these states.

2. Using FORCE_PLL_ON to interrupt an TX_ARET transaction, it is recommended to check register bits [7:5] of register address 0x32 for value 0. If this value is different, TRX_CMD sequence FORCE_TRX_OFF shall be used immediately followed by TRX_CMD sequence PLL_ON. This performs a state transition to PLL_ON.

Register 0x04 (TRX CTRL 1):

The TRX_CTRL_1 register is a multi purpose register to control various operating modes and settings of the radio transceiver.



• Bit 7 - PA EXT EN

Refer to Section 11.5 "RX/TX Indicator" on page 147.

• Bit 6 - IRQ_2_EXT_EN

Refer to Section 11.6 "RX Frame Time Stamping" on page 150.

• Bit 5 - TX_AUTO_CRC_ON

If set, register bit TX_AUTO_CRC_ON enables the automatic FCS generation. For further details refer to Section 8.2 "Frame Check Sequence (FCS)" on page 85.

• Bit 4 - RX_BL_CTRL

Refer to Section 11.7 "Frame Buffer Empty Indicator" on page 152.

Bit [3:2] - SPI_CMD_MODE

Refer to Section 6.3 "Radio Transceiver Status information" on page 24.

Bit 1 - IRQ MASK MODE

Refer to Section 6.6 "Interrupt Logic" on page 29.

• Bit 0 - IRQ_POLARITY

Refer to Section 6.6 "Interrupt Logic" on page 29.

Register 0x17 (XAH_CTRL_1):

The XAH_CTRL_1 register is a control register for Extended Operating Mode.



• Bit [7:6] - Reserved



• Bit 5 - AACK FLTR RES FT

This register bit shall only be set if AACK_UPLD_RES_FT = 1.

If AACK_FLTR_RES_FT = 1 reserved frame types are filtered similar to data frames as specified in IEEE 802.15.4-2006. Reserved frame types are explained in IEEE 802.15.4, section 7.2.1.1.1.

If AACK_FLTR_RES_FT = 0 the received reserved frame is only checked for a valid FCS.

• Bit 4 - AACK_UPLD_RES_FT

If AACK_UPLD_RES_FT = 1 received frames indicated as a reserved frame are further processed. For those frames, an IRQ 3 (TRX_END) interrupt is generated if the FCS is valid.

In conjunction with the configuration bit AACK_FLTR_RES_FT set, these frames are handled like IEEE 802.15.4 compliant data frames during RX_AACK transaction. An IRQ_5 (AMI) interrupt is issued, if the addresses in the received frame match the node's addresses.

That means, if a reserved frame passes the third level filter rules, an acknowledgement frame is generated and transmitted if it was requested by the received frame. If this is not wanted register bit AACK_DIS_ACK (register 0x2E, CSMA_SEED_1) has to be set.

· Bit 3 - Reserved

Bit 2 - AACK_ACK_TIME

According to IEEE 802.15.4, section 7.5.6.4.2, the transmission of an acknowledgment frame shall commence 12 symbols (aTurnaroundTime) after the reception of the last symbol of a data or MAC command frame. This is achieved with the reset value of the register bit AACK_ACK_TIME.

Alternatively, if AACK_ACK_TIME = 1 an acknowledgment frame is sent already 2 symbol periods after the reception of the last symbol of a data or MAC command frame. This may be applied to proprietary networks or networks using the High Data Rate Modes to increase battery lifetime and to improve the overall data throughput; refer to Section 11.3 "High Data Rate Modes" on page 137.

This setting affects also to acknowledgment frame response time for slotted acknowledgement operation, see Section 7.2.3.6 "RX_AACK Slotted Operation - Slotted Acknowledgement" on page 62.

• Bit 1 - AACK PROM MODE

Register bit AACK_PROM_MODE enables the promiscuous mode, within the RX_AACK mode; refer to IEEE 802.15.4-2006, section 7.5.6.5.

If this bit is set, every incoming frame with a valid PHR finishes with IRQ_3 (TRX_END) interrupt even if the third level filter rules do not match or the FCS is not valid. Register bit RX_CRC_VALID (register 0x06, PHY_RSSI) is set accordingly.

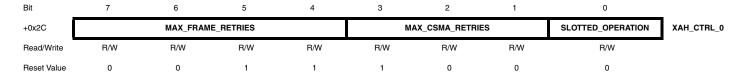
Here, if a frame passes the third level filter rules, an acknowledgement frame is generated and transmitted unless disabled by register bit AACK_DIS_ACK (register 0x2E, CSMA_SEED_1).

· Bit 0 - Reserved



Register 0x2C (XAH_CTRL_0):

Register 0x2C (XAH_CTRL_0) is a control register for Extended Operating Mode.



Bit [7:4] - MAX_FRAME_RETRIES

The setting of MAX_FRAME_RETRIES in TX_ARET mode specifies the number of attempts to retransmit a frame, when it was not acknowledged by the recipient, before the transaction gets cancelled.

• Bit [3:1] - MAX_CSMA_RETRIES

MAX_CSMA_RETRIES specifies the number of retries in TX_ARET mode to repeat the CSMA-CA procedure before the transaction gets cancelled. According IEEE 802.15.4 the valid range of MAX_CSMA_RETRIES is [0, 1, ..., 5].

A value of MAX_CSMA_RETRIES = 7 initiates an immediate frame transmission without performing CSMA-CA. This may especially be required for slotted acknowledgement operation. MAX_CSMA_RETRIES = 6 is reserved.

• Bit 0 - SLOTTED_OPERATION

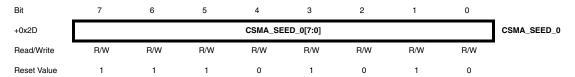
Using RX_AACK mode in networks operating in beacon or slotted mode, refer to IEEE 802.15.4 2006, section 5.5.1, register bit SLOTTED_OPERATION indicates that acknowledgement frames are to be sent on back-off slot boundaries (slotted acknowledgement).

If this register bit is set the acknowledgement frame transmission has to be initiated by the microcontroller using the rising edge of pin 11 (SLP_TR). This waiting state is signaled in sub register TRAC_STATUS (register 0x02, TRX_STATE) with value SUCCESS_WAIT_FOR_ACK.

 Table 7-18.
 Register Bit Slotted Acknowledgement Operation

Register Bit	Value	State Description
SLOTTED_OPERATION	<u>0</u>	The radio transceiver operates in unslotted mode. An acknowledgment frame is automatically sent if requested.
	1	Refer to Section 7.2.3.6. The transmission of an acknowledgement frame has to be controlled by the microcontroller.

Register 0x2D (CSMA_SEED_0):





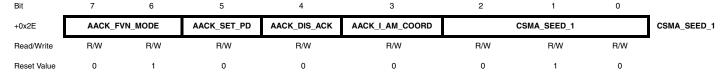
• Bit [7:0] - CSMA SEED 0

This register contains the lower 8-bit of the CSMA_SEED, bits [7:0]. The higher 3 bit are part of register bits CSMA_SEED_1 (register 0x2E, CSMA_SEED_1). CSMA_SEED is the seed for the random number generation that determines the length of the back-off period in the CSMA-CA algorithm.

It is recommended to initialize registers CSMA_SEED by random values. This can be done using register bits RND_VALUE (register 0x06, PHY_RSSI), refer to Section 11.2 "Random Number Generator" on page 136.

Register 0x2E (CSMA SEED 1):

The CSMA_SEED_1 register is a control register for RX_AACK and contains a part of the CSMA_SEED for the CSMA-CA algorithm.



Note: The register bits CSMA_SEED_0/1 content initializes the TX_ARET random backoff generator after leaving SLEEP state. To prevent a reinitialization with the same value it is recommended to reinitialize all register bits with random values before entering SLEEP state.

• Bit [7:6] - AACK_FVN_MODE

The frame control field of the MAC header (MHR) contains a frame version subfield. The setting of AACK_FVN_MODE specifies the frame filtering behavior of the AT86RF231. According to the content of these register bits the radio transceiver passes frames with a specific frame version number, number group, or independent of the frame version number.

Thus the register bit AACK_FVN_MODE defines the maximum acceptable frame version. Received frames with a higher frame version number than configured do not pass the address filter and are not acknowledged.

 Table 7-19.
 Register Bit Slotted Acknowledgement Operation

Register Bit	Value	State Description
AACK_FVN_MODE	0	Acknowledge frames with version number 0
	1	Acknowledge frames with version number 0 or 1
	2	Acknowledge frames with version number 0 or 1 or 2
	3	Acknowledge independent of frame version number

The frame version field of the acknowledgment frame is set to 0x00 according to IEEE 802.15.4-2006, section 7.2.2.3.1, Acknowledgment frame MHR fields.

Bit 5 - AACK_SET_PD

The content of AACK_SET_PD bit is copied into the frame pending subfield of the acknowledgment frame if the ACK is the answer to a data request MAC command frame.

In addition, if register bits AACK_FVN_MODE (register 0x2E, CSMA_SEED_1) are configured to accept frames with a frame version other than 0 or 1, the content of register bit



AACK_SET_PD is also copied into the frame pending subfield of the acknowledgment frame for any MAC command frame with a frame version of 2 or 3 that have the security enabled subfield set to 1. This is done in the assumption that a future version of the standard [1] might change the length or structure of the auxiliary security header, so it is not possible to safely detect whether the MAC command frame is actually a data request command or not.

Bit 4 - AACK_ DIS_ACK

If this bit is set no acknowledgment frames are transmitted in RX_AACK Extended Operating Mode, even if requested.

• Bit 3 - AACK I AM COORD

This register bit has to be set if the node is a PAN coordinator. It is used for address filtering in RX AACK.

If AACK_I_AM_COORD = 1 and if only source addressing fields are included in a data or MAC command frame, the frame shall be accepted only if the device is the PAN coordinator and the source PAN identifier matches macPANId, for details refer to IEEE 802.15.4, section 7.5.6.2 (third-level filter rule 6).

• Bit [2:0] - CSMA_SEED_1

These register bits are the higher 3-bit of the CSMA_SEED, bits [10:8]. The lower part is in register 0x2D (CSMA_SEED_0), see register CSMA_SEED_0 for details.

Register 0x2F (CSMA_BE):

Bit	7	6	5	4	3	2	1	0	
+0x2F		MAX	_BE			MIN_	BE		CSMA_BE
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	-
Reset Value	0	1	0	1	0	0	1	1	

• Bit [7:4] - MAX_BE

Register bits MAX_BE defines the maximum back-off exponent used in the CSMA-CA algorithm to generate a pseudo random number for back off the CCA. For details refer to IEEE 802.15.4-2006, Section 7.5.1.4.

Valid values are [4'd8, 4'd7, ..., 4'd3].

Bit [3:0] - MIN_BE

Register bits MIN_BE defines the minimum back-off exponent used in the CSMA-CA algorithm to generate a pseudo random number for back off the CCA. For details refer to IEEE 802.15.4-2006, Section 7.5.1.4.

Valid values are [MAX_BE, (MAX_BE - 1), ..., 4'd0].

Note

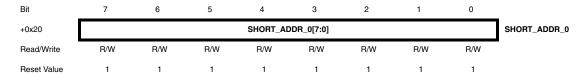
• If MIN_BE = 0 and MAX_BE = 0 the CCA back off period is always set to 0.



7.2.8 Register Description - Address Registers

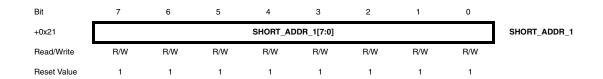
Register 0x20 (SHORT_ADDR_0):

This register contains the lower 8 bit of the MAC short address for Frame Filter address recognition, bits [7:0].



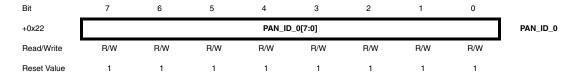
Register 0x21 (SHORT_ADDR_1):

This register contains the upper 8 bit of the MAC short address for Frame Filter address recognition, bits [15:8].



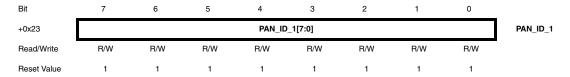
Register 0x22 (PAN_ID_0):

This register contains the lower 8 bit of the MAC PAN ID for Frame Filter address recognition, bits [7:0].



Register 0x23 (PAN_ID_1):

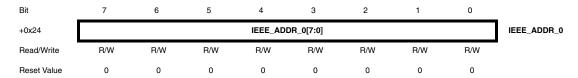
This register contains the upper 8 bit of the MAC PAN ID for Frame Filter address recognition, bits [15:8].





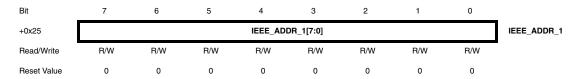
Register 0x24 (IEEE_ADDR_0):

This register contains the lower 8 bit of the MAC IEEE address for Frame Filter address recognition, bits [7:0].



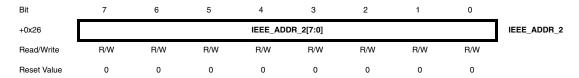
Register 0x25 (IEEE_ADDR_1):

This register contains 8 bit of the MAC IEEE address for Frame Filter address recognition, bits [15:8].



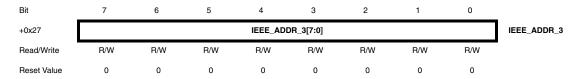
Register 0x26 (IEEE_ADDR_2):

This register contains 8 bit of the MAC IEEE address for Frame Filter address recognition, bits [23:16].



Register 0x27 (IEEE_ADDR_3):

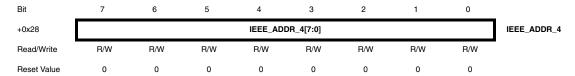
This register contains 8 bit of the MAC IEEE address for Frame Filter address recognition, bits [31:24].





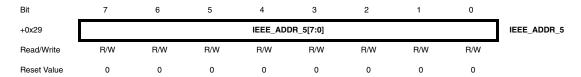
Register 0x28 (IEEE_ADDR_4):

This register contains 8 bit of the MAC IEEE address for Frame Filter address recognition, bits [39:32].



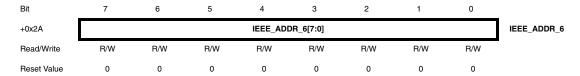
Register 0x29 (IEEE_ADDR_5):

This register contains 8 bit of the MAC IEEE address for Frame Filter address recognition, bits [47:40].



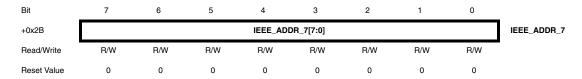
Register 0x2A (IEEE_ADDR_6):

This register contains 8 bit of the MAC IEEE address for Frame Filter address recognition, bits [55:48].



Register 0x2B (IEEE_ADDR_7):

This register contains the most significant 8 bits of the MAC IEEE Frame Filter address for address recognition, bits [63:56].





8. Functional Description

8.1 Introduction - IEEE 802.15.4 - 2006 Frame Format

Figure 8-1 on page 79 provides an overview of the physical layer (PHY) frame structure as defined by IEEE 802.15.4. Figure 8-2 on page 80 shows the frame structure of the medium access control (MAC) layer.

Figure 8-1. IEEE 802.15.4 Frame Format - PHY-Layer Frame Structure (PPDU)

		PHY Protocol Da	ta Unit (PPDU)
Preamble Sequence	SFD	Frame Length	PHY Payload
5 octets Synchronization Header (SHR)		1 octet (PHR)	max. 127 octets PHY Service Data Unit (PSDU)
			MAC Protocol Data Unit (MPDU)

8.1.1 PHY Protocol Layer Data Unit (PPDU)

8.1.1.1 8.1.1.1 Synchronization Header (SHR)

The SHR consists of a four-octet preamble field (all zero), followed by a single byte start-of-frame delimiter (SFD) which has the predefined value 0xA7. During transmit, the SHR is automatically generated by the AT86RF231, thus the Frame Buffer shall contain PHR and PSDU only.

The transmission of the SHR requires 160 μ s (10 symbols). As the SPI data rate is normally higher than the over-air data rate, this allows the microcontroller to initiate a transmission without having transferred the full frame data already. Instead it is possible to subsequently write the frame content.

During frame reception, the SHR is used for synchronization purposes. The matching SFD determines the beginning of the PHR and the following PSDU payload data.

8.1.1.2 PHY Header (PHR)

The PHY header is a single octet following the SHR. The least significant 7 bits denote the frame length of the following PSDU, while the most significant bit of that octet is reserved, and shall be set to 0 for IEEE 802.15.4 compliant frames.

On receive the PHR is returned as the first octet during Frame Buffer read access. Even though the standard only defines frame lengths ≤127 bytes, AT86RF231 is able to transmit and receive frame length values >127. For IEEE 802.15.4 compliant operation bit 7 has to be masked by SW. The reception of a valid PHR is signaled by an interrupt IRQ_2 (RX_START).

On transmit the PHR is to be supplied by the microcontroller during Frame Buffer write access as the first octet.

8.1.1.3 PHY Payload (PHY Service Data Unit, PSDU)

The PSDU has a variable length between 0 and aMaxPHYPacketSize (127, maximum PSDU size in octets) whereas the last two octets are used for the Frame Check Sequence (FCS). The length of the PSDU is signaled by the frame length field (PHR), refer to Table 8-1 on page 80. The PSDU contains the MAC Protocol Layer Data Unit (MPDU).



Received frames with a frame length field set to 0x00 (invalid PHR) are not signaled to the microcontroller.

Table 8-1 on page 80 summarizes the type of payload versus the frame length value.

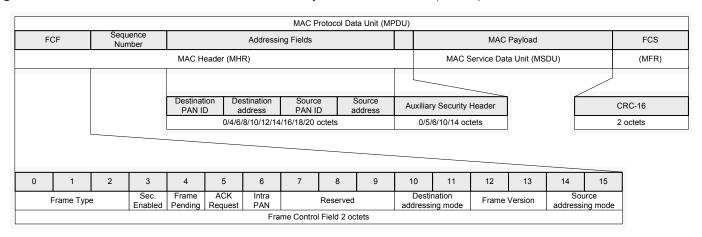
Table 8-1. Frame Length Field - PHR

Frame Length Value	Payload
0 - 4	Reserved
5	MPDU (Acknowledgement)
6 - 8	Reserved
9 - aMaxPHYPacketSize	MPDU

8.1.2 MAC Protocol Layer Data Unit (MPDU)

Figure 8-2 on page 80 shows the frame structure of the MAC layer.

Figure 8-2. IEEE 802.15.4 Frame Format - MAC-Layer Frame Structure (MPDU)



8.1.2.1 MAC Header (MHR) Fields

The MAC header consists of the Frame Control Field (FCF), a sequence number, and the addressing fields (which are of variable length, and can even be empty in certain situations).

8.1.2.2 Frame Control Field (FCF)

The FCF consists of 16 bits, and occupies the first two octets of the MPDU or PSDU, respectively.

Figure 8-3. IEEE 802.15.4-2006 Frame Control Field (FCF)

0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15
	Frame Type	е	Sec. Enabled	Frame Pending	ACK Request	Intra PAN		Reserved		Destii addressi		Frame	Version	Sou addressi	
	Frame Control Field 2 octets														



• Bit [2:0]:

describes the frame type. Table 8-2 on page 81 summarizes frame types defined by IEEE 802.15.4, section 7.2.1.1.1.

Table 8-2. Frame Control Field - Frame Type Subfield

Frame Control Fie	eld Bit Assignments	Description
Frame Type Value b ₂ b ₁ b ₀	Value	
000	0	Beacon
001	1	Data
010	2	Acknowledge
011	3	MAC command
100 - 111	4 - 7	Reserved

This subfield is used for address filtering by the third level filter rules. Only frame types 0 - 3 pass the third level filter rules, refer to Section 7.2.3.5 "Frame Filtering" on page 61 Automatic address filtering by the AT86RF231 is enabled when using the RX_AACK mode, refer to Section 7.2.3 "RX_AACK_ON - Receive with Automatic ACK" on page 51.

However, a reserved frame (frame type value > 3) can be received if register bit AACK_UPLD_RES_FT (register 0x17, XAH_CTRL_1) is set, for details refer to Section 7.2.3.3 "Configuration of non IEEE 802.15.4 Compliant Scenarios" on page 58.

Address filtering is also provided in Basic Operating Mode, refer to Section 7.1 "Basic Operating Mode" on page 33.

• Bit 3:

indicates whether security processing applies to this frame.

• Bit 4:

is the "Frame Pending" subfield. This field can be set in an acknowledgment frame (ACK) in response to a data request MAC command frame. This bit indicates that the node, which transmitted the ACK, has more data to send to the node receiving the ACK.

For acknowledgment frames automatically generated by the AT86RF231, this bit is set according to the content of register bit AACK_SET_PD in register 0x2E (CSMA_SEED_1) if the received frame was a data request MAC command frame.

• Bit 5:

forms the "Acknowledgment Request" subfield. If this bit is set within a data or MAC command frame that is not broadcast, the recipient shall acknowledge the reception of the frame within the time specified by IEEE 802.15.4 (i.e. within 192 µs for non beacon-enabled networks).

The radio transceiver parses this bit during RX_AACK mode and transmits an acknowledgment frame if necessary.

In TX_ARET mode this bit indicates if an acknowledgement frame is expected after transmitting a frame. If this is the case, the receiver waits for the acknowledgment frame, otherwise the TX_ARET transaction is finished.



• Bit 6:

the "Intra-PAN" subfield indicates that in a frame, where both, the destination and source addresses are present, the PAN-ID of the source address field is omitted. In RX_AACK mode, this bit is evaluated by the address filter logic of the AT86RF231.

• Bit [11:10]:

the "Destination Addressing Mode" subfield describes the format of the destination address of the frame. The values of the address modes are summarized in Table 8-3 on page 82, according to IEEE 802.15.4.

 Table 8-3.
 Frame Control Field - Destination and Source Addressing Mode

Frame Control Fie	eld Bit Assignments	Description
Addressing Mode b ₁₁ b ₁₀ b ₁₅ b ₁₄	Value	
00	0	PAN identifier and address fields are not present
01	1	Reserved
10	2	Address field contains a 16-bit short address
11	3	Address field contains a 64-bit extended address

If the destination address mode is either 2 or 3 (i.e. if the destination address is present), it always consists of a 16-bit PAN ID first, followed by either the 16-bit or 64-bit address as described by the mode.

• Bit [13:12]:

the "Frame Version" subfield specifies the version number corresponding to the frame. These register bits are reserved in IEEE 802.15.4-2003.

This subfield shall be set to 0 to indicate a frame compatible with IEEE 802.15.4-2003 and 1 to indicate an IEEE 802.15.4-2006 frame. All other subfield values shall be reserved for future use.

RX_AACK register bit AACK_FVN_MODE (register 0x2E, CSMA_SEED_1) controls the behavior of frame acknowledgements. This register determines if, depending on the Frame Version Number, a frame is acknowledged or not. This is necessary for backward compatibility to IEEE 802.15.4-2003 and for future use. Even if frame version numbers 2 and 3 are reserved, it can be handled by the radio transceiver, for details refer to Section 7.2.7 "Register Description - Control Registers" on page 68.

See IEEE 802.15.4-2006, section 7.2.3 for details on frame compatibility.

Table 8-4. Frame Control Field - Frame Version Subfield

Frame Control Fig	eld Bit Assignments	Description
Frame Version b ₁₃ b ₁₂	Value	
00	0	Frames are compatible with IEEE 802.15.4 2003
01	1	Frames are compatible with IEEE 802.15.4-2006
10	2	Reserved
11	3	Reserved



• Bit [15:14]:

the "Source Addressing Mode" subfield, with similar meaning as "Destination Addressing Mode", see Table 8-3 on page 82.

The subfields of the FCF (Bits 0-2, 3, 6, 10-15) affect the address filter logic of the AT86RF231 while operating in RX_AACK operation, see Section 7.2.3 "RX_AACK_ON - Receive with Automatic ACK" on page 51.

8.1.2.3 Frame Compatibility between IEEE 802.15.4-2003 and IEEE 802.15.4-2006

All unsecured frames according to IEEE 802.15.4-2006 are compatible with unsecured frames compliant with IEEE 802.15.4-2003 with two exceptions: a coordinator realignment command frame with the "Channel Page" field present (see IEEE 802.15.4-2006, section 7.3.8) and any frame with a MAC Payload field larger than *aMaxMACSafePayloadSize* octets.

Compatibility for secured frames is shown in Table 8-5 on page 83, which identifies the security operating modes for IEEE 802.15.4-2006.

Frame Control Field	d Bit Assignments	Description
Security Enabled	Frame Version	
b_3	$b_{13} b_{12}$	
0	00	No security. Frames are compatible between IEEE 802.15.4-2003 and IEEE 802.15.4-2006.
0	01	No security. Frames are not compatible between IEEE 802.15.4-2003 and IEEE 802.15.4-2006.
1	00	Secured frame formatted according to IEEE 802.15.4-2003. This frame type is not supported in IEEE 802.15.4-2006.
1	01	Secured frame formatted according to IEEE 802.15.4-2006

Table 8-5. Frame Control Field - Security and Frame Version

8.1.2.4 Sequence Number

The one-octet sequence number following the FCF identifies a particular frame, so that duplicated frame transmissions can be detected. While operating in RX_AACK mode, the content of this field is copied from the frame to be acknowledged into the acknowledgment frame.

8.1.2.5 Addressing Fields

The addressing fields of the MPDU are used by the AT86RF231 for address matching indication. The destination address (if present) is always first, followed by the source address (if present). Each address field consists of the Intra PAN ID and a device address. If both addresses are present, and the "Intra PAN-ID compression" subfield in the FCF is set to one, the source Intra PAN ID is omitted.

Note that in addition to these general rules, IEEE 802.15.4 further restricts the valid address combinations for the individual possible MAC frame types. For example, the situation where both addresses are omitted (source addressing mode = 0 and destination addressing mode = 0) is only allowed for acknowledgment frames. The address filter in the AT86RF231 has been designed to apply to IEEE 802.15.4 compliant frames. It can be configured to handle other frame formats and exceptions.



8.1.2.6 Auxiliary Security Header Field

The Auxiliary Security Header specifies information required for security processing and has a variable length. This field determines how the frame is actually protected (security level) and which keying material from the MAC security PIB is used (see IEEE 802.15.4-2006, section 7.6.1). This field shall be present only if the Security Enabled subfield b3, see Section 8.1.2.3 "Frame Compatibility between IEEE 802.15.4-2003 and IEEE 802.15.4-2006" on page 83, is set to one. For details of its structure, see IEEE 802.15.4-2006, section 7.6.2. Auxiliary security header.

8.1.2.7 MAC Service Data Unit (MSDU)

This is the actual MAC payload. It is usually structured according to the individual frame type. A description can be found in IEEE 802.15.4-2006, section 5.5.3.2.

8.1.2.8 MAC Footer (MFR) Fields

The MAC footer consists of a two-octet Frame Checksum (FCS), for details refer to Section 8.2 "Frame Check Sequence (FCS)" on page 85.



8.2 Frame Check Sequence (FCS)

The Frame Check Sequence (FCS) is characterized by:

- Indicate bit errors, based on a cyclic redundancy check (CRC) of length 16 bit
- Uses International Telecommunication Union (ITU) CRC polynomial
- · Automatically evaluated during reception
- · Can be automatically generated during transmission

8.2.1 Overview

The FCS is intended for use at the MAC layer to detect corrupted frames at a first level of filtering. It is computed by applying an ITU CRC polynomial to all transferred bytes following the length field (MHR and MSDU fields). The frame check sequence has a length of 16 bit and is located in the last two bytes of a frame (MAC footer, see Figure 8-2 on page 80).

The AT86RF231 applies an FCS check on each received frame. The FCS check result is stored in register bit RX_CRC_VALID in register 0x06 (PHY_RSSI).

On transmit the radio transceiver generates and appends the FCS bytes during the frame transmission. This behavior can be disabled by setting register bit TX_AUTO_CRC_ON = 0 (register 0x04, TRX_CTRL_1).

8.2.2 CRC Calculation

The CRC polynomial used in IEEE 802.15.4 networks is defined by:

$$G_{16}(x) = x^{16} + x^{12} + x^5 + 1$$

The FCS shall be calculated for transmission using the following algorithm:

Let

$$M(x) = b_0 x^{k-1} + b_1 x^{k-2} + b_2 x^{k-3} + ... + b_{k-2} x + b_{k-1}$$

be the polynomial representing the sequence of bits for which the checksum is to be computed. Multiply M(x) by x^{16} , giving the polynomial

$$N(x) = M(x) \bullet x^{16}$$

Divide N(x) modulo 2 by the generator polynomial, $G_{16}(x)$, to obtain the remainder polynomial,

$$R(x) = r_0 x^{15} + r_1 x^{14} + ... + r_{14} x + r_{15}$$

The FCS field is given by the coefficients of the remainder polynomial, R(x).

Example:

Considering a 5 octet ACK frame. The MHR field consists of

0100 0000 0000 0000 0101 0110.

The leftmost bit (b₀) is transmitted first in time. The FCS is in this case

0010 0111 1001 1110.

The leftmost bit (r_0) is transmitted first in time.



8.2.3 Automatic FCS generation

The automatic FCS generation is performed with register bit TX_AUTO_CRC_ON = 1 (reset value). This allows the AT86RF231 to compute the FCS autonomously. For a frame with a frame length specified as N ($3 \le N \le 127$), the FCS is calculated on the first N-2 octets in the Frame Buffer, and the resulting FCS field is transmitted in place of the last two octets from the Frame Buffer.

If the radio transceivers automatic FCS generation is enabled, the Frame Buffer write access can be stopped right after MAC payload. There is no need to write FCS dummy bytes.

In RX_AACK mode, when a received frame needs to be acknowledged, the FCS of the ACK frame is always automatically generated by the AT86RF231, independent of the TX_AUTO_CRC_ON setting.

Example:

A frame transmission of length five with TX_AUTO_CRC_ON set, is started with a Frame Buffer write access of five bytes (the last two bytes can be omitted). The first three bytes are used for FCS generation; the last two bytes are replaced by the internally calculated FCS.

8.2.4 Automatic FCS check

An automatic FCS check is applied on each received frame with a frame length $N \square 2$. Register bit RX_CRC_VALID (register 0x06, PHY_RSSI) is set if the FCS of a received frame is valid. The register bit is updated when issuing interrupt IRQ_3 (TRX_END) and remains valid until the next TRX_END interrupt caused by a new frame reception.

In RX_AACK mode, if FCS of the received frame is not valid, the radio transceiver rejects the frame and the TRX_END interrupt is not issued.

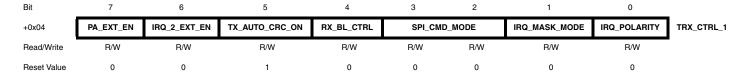
In TX_ARET mode, the FCS and the sequence number of an ACK is automatically checked. If one of these is not correct, the ACK is not accepted.



8.2.5 Register Description

Register 0x04 (TRX_CTRL_1):

The TRX_CTRL_1 register is a multi purpose register to control various operating modes and settings of the radio transceiver.



• Bit 7 - PA_EXT_EN

Refer to Section 11.5 "RX/TX Indicator" on page 147.

Bit 6 - IRQ_2_EXT_EN

Refer to Section 11.6 "RX Frame Time Stamping" on page 150.

Bit 5 - TX_AUTO_CRC_ON

Register bit TX_AUTO_CRC_ON controls the automatic FCS generation for TX operations. The automatic FCS algorithm is performed autonomously by the radio transceiver if register bit TX_AUTO_CRC_ON = 1.

Bit 4 - RX BL CTRL

Refer to Section 11.7 "Frame Buffer Empty Indicator" on page 152.

• Bit [3:2] - SPI_CMD_MODE

Refer to Section 6.3 "Radio Transceiver Status information" on page 24.

Bit 1 - IRQ_MASK_MODE

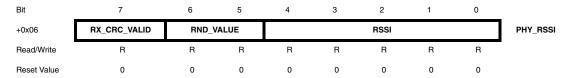
Refer to Section 6.6 "Interrupt Logic" on page 29.

• Bit 0 - IRQ POLARITY

Refer to Section 6.6 "Interrupt Logic" on page 29.

Register 0x06 (PHY_RSSI):

The PHY_RSSI register is a multi purpose register that indicates FCS validity, provides random numbers and shows the actual RSSI value.





• Bit 7 - RX_CRC_VALID

Reading this register bit indicates whether the last received frame has a valid FCS or not. The register bit is updated when issuing interrupt IRQ_3 (TRX_END) and remains valid until the next TRX_END interrupt is issued, caused by a new frame reception.

Table 8-6. RX Frame FCS Check

Register Bit	Value	State Description
RX_CRC_VALID	<u>0</u>	FCS is not valid
	1	FCS is valid

• Bit [6:5] - RND_VALUE

Refer to register description in Section 11.2.2 "Register Description" on page 136.

• Bit [4:0] - RSSI

Refer to register description in Section 8.3.4 "Register Description" on page 90.



8.3 Received Signal Strength Indicator (RSSI)

The Received Signal Strength Indicator is characterized by:

- Minimum RSSI level is -91 dBm (RSSI_BASE_VAL)
- Dynamic range is 81 dB
- Minimum RSSI value is 0
- Maximum RSSI value is 28

8.3.1 Overview

The RSSI is a 5-bit value indicating the receive power in the selected channel, in steps of 3 dB. No attempt is made to distinguish IEEE 802.15.4 signals from others, only the received signal strength is evaluated. The RSSI provides the basis for an ED measurement, see Section 8.4 "Energy Detection (ED)" on page 91.

8.3.2 Reading RSSI

In Basic Operating Mode the RSSI value is valid in any receive state, and is updated every $t_{TR25} = 2 \mu s$ to register 0x06 (PHY_RSSI).

It is not recommended to read the RSSI value when using the Extended Operating Mode. The automatically generated ED value should be used alternatively, see Section 8.4 "Energy Detection (ED)" on page 91.

8.3.3 Data Interpretation

The RSSI value is a 5-bit value indicating the receive power, in steps of 3 dB and with a range of 0 - 28.

An RSSI value of 0 indicates a receiver RF input power of $P_{RF} < -91$ dBm. For an RSSI value in the range of 1 to 28, the RF input power can be calculated as follows:

 $P_{RF} = RSSI_BASE_VAL + 3*(RSSI - 1) [dBm]$



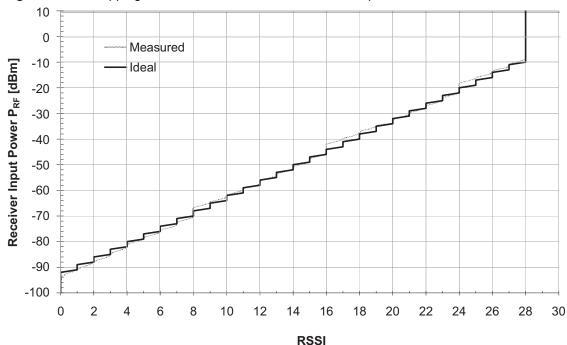
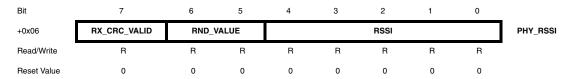


Figure 8-4. Mapping between RSSI Value and Received Input Power

8.3.4 Register Description

Register 0x06 (PHY_RSSI):



• Bit 7 - RX_CRC_VALID

Refer to register description in Section 8.2.5 "Register Description" on page 87.

• Bit [6:5] - RND_VALUE

Refer to register description in section Section 11.2.2 "Register Description" on page 136.

• Bit [4:0] - RSSI

The result of the automated RSSI measurement is stored in register bits RSSI. The value is updated every 2 μ s in receive states.

The read value is a number between 0 and 28 indicating the received signal strength as a linear curve on a logarithmic input power scale (dBm) with a resolution of 3 dB. An RSSI value of 0 indicates an RF input power of $P_{RF} < -91$ dBm (see parameter 12.7.16), a value of 28 a power of $P_{RF} = 10$ dBm (see parameter 12.7.18).



8.4 Energy Detection (ED)

The Energy Detection (ED) module is characterized by:

- 85 unique energy levels defined
- 1 dB resolution

8.4.1 Overview

The receiver ED measurement is used by the network layer as part of a channel selection algorithm. It is an estimation of the received signal power within the bandwidth of an IEEE 802.15.4 channel. No attempt is made to identify or decode signals on the channel. The ED value is calculated by averaging RSSI values over eight symbols (128 µs).

For High Data Rate Modes the automated ED measurement duration is reduced to 32 μs, refer to Section 11.3 "High Data Rate Modes" on page 137. For manually initiated ED measurements in these modes the measurement period is still 128 μs as long as the receiver is in RX_ON state.

8.4.2 Measurement Description

There are two ways to initiate an ED measurement:

- Manually, by writing an arbitrary value to register 0x07 (PHY_ED_LEVEL), or
- Automatically, after detection of a valid SHR of an incoming frame.

For manually initiated ED measurements the radio transceiver needs to be in one of the states RX_ON or BUSY_RX state. The end of the ED measurement is indicated by an interrupt IRQ_4 (CCA_ED_DONE).

An automated ED measurement is started if an SHR is detected. The end of the automated measurement is not signaled by an interrupt.

The measurement result is stored after t_{TR26} = 140 μs (128 μs measurement duration and processing delay) in register 0x07 (PHY_ED_LEVEL).

Thus by using Basic Operating Mode, a valid ED value from the currently received frame is accessible 108 µs after IRQ_2 (RX_START) and remains valid until a new RX_START interrupt is generated by the next incoming frame or until another ED measurement is initiated.

By using the Extended Operating Mode, it is recommended to mask IRQ_2 (RX_START), thus the interrupt cannot be used as timing reference. A successful frame reception is signalized by interrupt IRQ_3 (TRX_END). The minimum time span between a TRX_END interrupt and a following SFD detection is $t_{TR27}=96~\mu s$ due to the length of the SHR. Including the ED measurement time, the ED value needs to be read within 224 μs after the TRX_END interrupt; otherwise, it could be overwritten by the result of the next measurement cycle. This is important for time critical applications or if interrupt IRQ_2 (RX_START) is not used to indicate the reception of a frame.

Note, it is not recommended to manually initiate an ED measurement when using the Extended Operating Mode.

The values of the register 0x07 (PHY_ED_LEVEL) are:

Table 8-7. Register Bit PHY ED LEVEL Interpretation

PHY_ED_LEVEL	Description
<u>0xFF</u>	Reset value
0x00 0x54	ED measurement result of the last ED measurement



8.4.3 Data Interpretation

The PHY_ED_LEVEL is an 8-bit register. The ED value of the AT86RF231 has a valid range from 0x00 to 0x54 with a resolution of 1 dB. All other values do not occur; a value of 0xFF indicates the reset value. A value of PHY_ED_LEVEL = 0 indicates that the measured energy is less than -91 dBm (see parameter 12.7.16 RSSI_BASE_VAL, Section 12.7 "Receiver Characteristics" on page 160). Due to environmental conditions (temperature, voltage, semiconductor parameters, etc.) the calculated ED value has a maximum tolerance of ±5 dB, this is to be considered as constant offset over the measurement range.

An ED value of 0 indicates an RF input power of $P_{RF} \le -91$ dBm. For an ED value in the range of 0 to 84, the RF input power can be calculated as follows:

$$P_{RF} = -91 + ED [dBm]$$

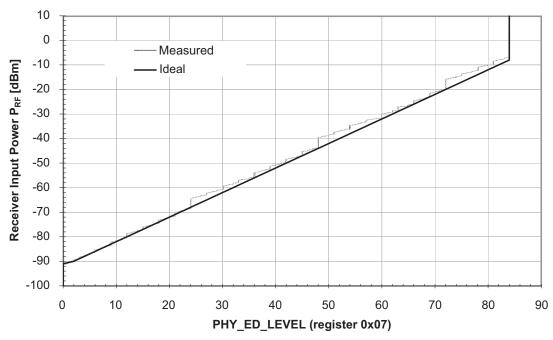


Figure 8-5. Mapping between Received Input Power and ED Value

8.4.4 Interrupt Handling

Interrupt IRQ_4 (CCA_ED_DONE) is issued at the end of a manually initiated ED measurement.

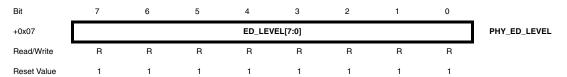
Note that an ED request should only be initiated in receive states. Otherwise the radio transceiver generates an IRQ 4 (CCA ED DONE); however no ED measurement was performed.



8.4.5 Register Description

Register 0x07 (PHY_ED_LEVEL):

The PHY_ED_LEVEL register contains the result of an ED measurement.



• Bit [7:0] - ED_LEVEL

The minimum ED value (ED_LEVEL = 0) indicates receiver power less than or equal to RSSI_BASE_VAL. The range is 84 dB with a resolution of 1 dB and an absolute accuracy of ±5 dB. A manual ED measurement can be initiated by a write access to the register. A value 0xFF signals that a measurement has never been started yet (reset value).

The measurement duration is 8 symbol periods (128 µs) for a data rate of 250 kb/s.

For High Data Rate Modes the automated measurement duration is reduced to 32 μ s, refer to Section 11.3 "High Data Rate Modes" on page 137. For manually initiated ED measurements in these modes the measurement period is still 128 μ s as long as the receiver is in RX_ON state.

A value other than 0xFF indicates the result of the last ED measurement.



8.5 Clear Channel Assessment (CCA)

The main features of the Clear Channel Assessment (CCA) module are:

- All 4 modes are available as defined by IEEE 802.15.4-2006 in section 6.9.9
- · Adjustable threshold for energy detection algorithm

8.5.1 Overview

A CCA measurement is used to detect a clear channel. Four modes are specified by IEEE 802.15.4 - 2006:

Table 8-8. CCA Mode Overview

CCA Mode	Description
1	Energy above threshold. CCA shall report a busy medium upon detecting any energy above the ED threshold.
2	Carrier sense only. CCA shall report a busy medium only upon the detection of a signal with the modulation and spreading characteristics of an IEEE 802.15.4 compliant signal. The signal strength may be above or below the ED threshold.
0, 3	Carrier sense with energy above threshold. CCA shall report a busy medium using a logical combination of — Detection of a signal with the modulation and spreading characteristics of this standard and — Energy above the ED threshold.
	Where the logical operator may be configured as either OR (mode 0) or AND (mode 3).

8.5.2 Configuration and Request

The CCA modes are configurable via register 0x08 (PHY_CC_CCA).

Using the Basic Operating Mode, a CCA request can be initiated manually by setting CCA_REQUEST = 1 (register 0x08, PHY_CC_CCA), if the AT86RF231 is in any RX state. The current channel status (CCA_STATUS) and the CCA completion status (CCA_DONE) are accessible in register 0x01 (TRX_STATUS).

The CCA evaluation is done over eight symbol periods and the result is accessible t_{TR28} = 140 µs (128 µs measurement duration and processing delay) after the request. The end of a manually initiated CCA measurement is indicated by an interrupt IRQ_4 (CCA_ED_DONE).

The sub-register CCA_ED_THRES of register 0x09 (CCA_THRES) defines the received power threshold of the "*Energy above threshold*" algorithm. The threshold is calculated by RSSI_BASE_VAL + 2 * CCA_ED_THRES [dBm]. Any received power above this level is interpreted as a busy channel.

Note, it is not recommended to manually initiate a CCA measurement when using the Extended Operating Mode.



8.5.3 Data Interpretation

The current channel status (CCA_STATUS) and the CCA completion status (CCA_DONE) are accessible in register 0x01 (TRX_STATUS). Note, register bits CCA_DONE and CCA_STATUS are cleared in response to a CCA_REQUEST.

The completion of a measurement cycle is indicated by CCA_DONE = 1. If the radio transceiver detected no signal (idle channel) during the measurement cycle, the CCA_STATUS bit is set to 1.

When using the "energy above threshold" algorithm, any received power above CCA_ED_THRES level is interpreted as a busy channel. The "carrier sense" algorithm reports a busy channel when detecting an IEEE 802.15.4 signal above the RSSI_BASE_VAL (see parameter 12.7.16). The radio transceiver is also able to detect signals below this value, but the detection probability decreases with the signal power. It is almost zero at the radio transceivers sensitivity level (see parameter 12.7.1).

8.5.4 Interrupt Handling

Interrupt IRQ_4 (CCA_ED_DONE) is issued at the end of a manually initiated CCA measurement.

Notes

- A CCA request should only be initiated in Basic Operating Mode receive states. Otherwise
 the radio transceiver generates an IRQ_4 (CCA_ED_DONE) and sets the register bit
 CCA_DONE = 1, even though no CCA measurement was performed.
- Requesting a CCA measurement in BUSY_RX state and during an ED measurement, an IRQ_4 (CCA_ED_DONE) could be issued immediately after the request. If in this case register bit CCA_DONE = 0, an additional interrupt CCA_ED_DONE is issued after finishing the CCA measurement and register bit CCA_DONE is set to 1.

8.5.5 Measurement Time

The response time for a manually initiated CCA measurement depends on the receiver state.

In RX_ON state the CCA measurement is done over eight symbol periods and the result is accessible 140 µs after the request (see above).

In BUSY_RX state the CCA measurement duration depends on the CCA Mode and the CCA request relative to the reception of an SHR. The end of the CCA measurement is indicated by an IRQ_4 (CCA_ED_DONE). The variation of a CCA measurement period in BUSY_RX state is described in Table 8-9 on page 95.

Table 8-9. CCA Measurement Period and Access in BUSY_RX state

CCA Mode	Request within ED measurement ⁽¹⁾	Request after ED measurement				
<u>1</u>	Energy above threshold.					
	CCA result is available after finishing automated ED measurement period.	CCA result is immediately available after request.				
2	Carrier sense only.					
	CCA result is immediately available after request.					



Table 8-9. CCA Measurement Period and Access in BUSY_RX state

3	Carrier sense with Energy above threshold (AND).					
	CCA result is available after finishing automated ED measurement period.	CCA result is immediately available after request.				
0	Carrier sense with Energy above threshold (OR).					
	CCA result is available after finishing automated ED measurement period	CCA result is immediately available after request.				

Note:
1. After receiving the SHR an automated ED measurement is started with a length of 8 symbol periods (PSDU rate 250 kb/s), refer to Section 8.4 "Energy Detection (ED)" on page 91. This automated ED measurement must be finished to provide a result for the CCA measurement. Only one automated ED measurement per frame is performed.

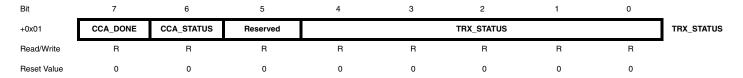
It is recommended to perform CCA measurements in RX_ON state only. To avoid switching accidentally to BUSY_RX state the SHR detection can be disabled by setting register bit RX_PDT_DIS (register 0x15, RX_SYN), refer to Section 9.1 "Receiver (RX)" on page 101. The receiver remains in RX_ON state to perform a CCA measurement until the register bit RX_PDT_DIS is set back to continue the frame reception. In this case the CCA measurement duration is 8 symbol periods.



8.5.6 Register Description

Register 0x01 (TRX_STATUS):

Two register bits of register 0x01 (TRX_STATUS) signal the status of the CCA measurement.



• Bit 7 - CCA DONE

This register indicates if a CCA request is completed. This is also indicated by an interrupt IRQ_4 (CCA_ED_DONE). Note, register bit CCA_DONE is cleared in response to a CCA_REQUEST.

Table 8-10. CCA Algorithm Status

Register Bit	Value	State Description
CCA_DONE	<u>0</u>	CCA calculation not finished
	1	CCA calculation finished

• Bit 6 - CCA_STATUS

After a CCA request is completed the result of the CCA measurement is available in register bit CCA_STATUS. Note, register bit CCA_STATUS is cleared in response to a CCA_REQUEST.

Table 8-11. CCA Status Result

Register Bit Value		State Description		
CCA_STATUS	<u>0</u>	Channel indicated as busy		
	1	Channel indicated as idle		

· Bit 5 - Reserved

• Bit [4:0] - TRX STATUS

Refer to Section 7.1.5 "Register Description" on page 44 and Section 7.2.7 "Register Description - Control Registers" on page 68.

Register 0x08 (PHY_CC_CCA):

This register is provided to initiate and control a CCA measurement.



• Bit 7 - CCA REQUEST

A manual CCA measurement is initiated with setting CCA_REQUEST = 1. The end of the CCA measurement is indicated by interrupt IRQ_4 (CCA_ED_DONE). Register bits CCA_DONE and



CCA_STATUS (register 0x01, TRX_STATUS) are updated after a CCA_REQUEST. The register bit is automatically cleared after requesting a CCA measurement with CCA_REQUEST = 1.

Bit [6:5] - CCA_MODE

The CCA mode can be selected using register bits CCA_MODE.

Table 8-12. CCA Status Result

Register Bit	Value	State Description
CCA_MODE	0	Mode 3a, Carrier sense OR energy above threshold
	1	Mode 1, Energy above threshold
	2	Mode 2, Carrier sense only
	3	Mode 3b, Carrier sense AND energy above threshold

Note that IEEE 802.15.4-2006 CCA Mode 3 defines the logical combination of CCA Mode 1 and 2 with the logical operators AND or OR. This can be selected with:

CCA_MODE = 0 for logical operation OR, and
 CCA_MODE = 3 for logical operation AND.

• Bit [4:0] - CHANNEL

Refer to Section 9.7 "Frequency Synthesizer (PLL)" on page 121.

Register 0x09 (CCA_THRES):

This register sets the ED threshold level for CCA.

Bit	7	6	5	4	3	2	1	0	_
+0x09	Reserved				CCA_ED_THRES				CCA_THRES
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	_
Reset Value	1	1	0	0	0	1	1	1	

• Bit [7:5] - Reserved

• Bit [4:0] - CCA ED THRES

The CCA Mode 1 request indicates a busy channel if the measured received power is above RSSI_BASE_VAL + 2 * CCA_ED_THRES [dBm]. CCA Modes 0 and 3 are logical related to this result.



8.6 Link Quality Indication (LQI)

According to IEEE 802.15.4, the LQI measurement is a characterization of the strength and/or quality of a received packet. The measurement may be implemented using receiver ED, a signal-to-noise ratio estimation, or a combination of these methods. The use of the LQI result by the network or application layers is not specified in this standard. LQI values shall be an integer ranging from 0x00 to 0xFF. The minimum and maximum LQI values (0x00 and 0xFF) should be associated with the lowest and highest quality compliant signals, respectively, and LQI values in between should be uniformly distributed between these two limits.

8.6.1 Overview

The LQI measurement of the AT86RF231 is implemented as a measure of the link quality which can be described with the packet error rate (PER) for this link. An LQI value can be associated with an expected packet error rate. The PER is the ratio of erroneous received frames to the total number of received frames. A PER of zero indicates no frame error, whereas at a PER of one no frame was received correctly.

The radio transceiver uses correlation results of multiple symbols within a frame to determine the LQI value. This is done for each received frame. The minimum frame length for a valid LQI value is two octets PSDU. LQI values are integers ranging from 0 to 255.

As an example, Figure 8-6 on page 99 shows the conditional packet error when receiving a certain LQI value.

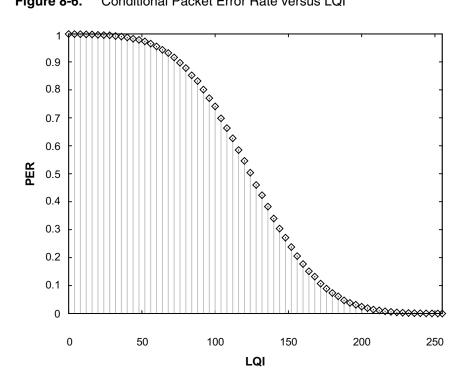


Figure 8-6. Conditional Packet Error Rate versus LQI

The values are taken from received frames of PSDU length of 20 octets on transmission channels with reasonable low multipath delay spreads. If the transmission channel characteristic has higher multipath delay spread than assumed in the example, the PER is slightly higher for a cer-



tain LQI value. Since the packet error rate is a statistical value, the PER shown in Section 8-6 "Conditional Packet Error Rate versus LQI" on page 99 is based on a huge number of transactions. A reliable estimation of the packet error rate cannot be based on a single or a small number of LQI values.

8.6.2 Request an LQI Measurement

The LQI byte can be obtained after a frame has been received by the radio transceiver. One additional byte is automatically attached to the received frame containing the LQI value. This information can also be read via Frame Buffer read access, see Section 6.2.2 "Frame Buffer Access Mode" on page 20. The LQI byte can be read after IRQ 3 (TRX END) interrupt.

8.6.3 Data Interpretation

According to IEEE 802.15.4 a low LQI value is associated with low signal strength and/or high signal distortions. Signal distortions are mainly caused by interference signals and/or multipath propagation. High LQI values indicate a sufficient high signal power and low signal distortions.

Note, the received signal power as indicated by received signal strength indication (RSSI) value or energy detection (ED) value of the AT86RF231 do not characterize the signal quality and the ability to decode a signal.

As an example, a received signal with an input power of about 6 dB above the receiver sensitivity likely results in a LQI value close to 255 for radio channels with very low signal distortions. For higher signal power the LQI value becomes independent of the actual signal strength. This is because the packet error rate for these scenarios tends towards zero and further increased signal strength, i.e. increasing the transmission power does not decrease the error rate any further. In this case RSSI or ED can be used to evaluate the signal strength and the link margin.

ZigBee networks often require the identification of the "best" routing between two nodes. Both, the LQI and the RSSI/ED can be used for this, dependent on the optimization criteria. If a low packet error rate (corresponding to high throughput) is the optimization criteria then the LQI value should be taken into consideration. If a low transmission power or the link margin is the optimization criteria then the RSSI/ED value is also helpful.

Combinations of LQI, RSSI and ED are possible for routing decisions. As a rule of thumb RSSI and ED values are useful to differentiate between links with high LQI values. Transmission links with low LQI values should be discarded for routing decisions even if the RSSI/ED values are high. This is because RSSI/ED does not say anything about the possibility to decode a signal. It is only an information about the received signal strength whereas the source can be an interferer.

